

Fall 2022 (111-1)

控制系統  
Control Systems

Unit 5B  
Guidelines for Determining a Root Locus

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# Formal Definition of Root Locus

- **Definition I:**

$$1 + K \frac{b(s)}{a(s)} = 0$$
- The **root locus** is the set of values of  $s$  for which  $1 + KL(s) = 0$  is satisfied as the real parameter  $K$  varies from  $0$  to  $+\infty$ 

$$L(s) = \frac{b(s)}{a(s)}$$
- Typically,  $1 + KL(s) = 0$  is the **characteristic equation** of the system, and in this case **the roots** on the locus are the **closed-loop poles** of that system.
 
$$\Rightarrow L(s) = -\frac{1}{K}$$
- **Definition II:**
- The **root locus** of  $L(s)$  is the set of points in the  $s$ -plane where the **phase** of  $L(s)$  is  $180^\circ$ .
- To test whether a point in the  $s$ -plane is on the locus, we define the **angle** to the test point from a **zero** as  $\psi$  and the **angle** to test point from a **pole** as  $\phi$  as follows:

$$\sum \psi_i - \sum \phi_i = 180^\circ + 360^\circ (l - 1)$$

# Formal Definition of Root Locus

- $K$  is real and **positive**,  
 the **phase** of  $L(s)$  is  $180^\circ$ ,  
 the **positive locus** or  $180^\circ$  locus

$$\Rightarrow L(s) = -\frac{1}{K}$$

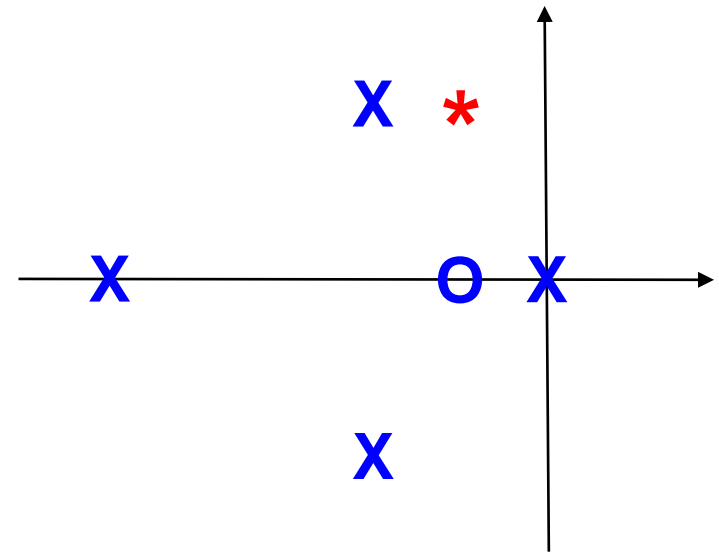
- $K$  is real and **negative**,  
 the **phase** of  $L(s)$  is  $0^\circ$ ,  
 the **negative locus** or  $0^\circ$  locus

- Illustrative Example:

$$L(s) = \frac{s + 1}{s(s + 5)[(s + 2)^2 + 4]}$$

$$s_0 = -1 + 2j$$

$$\angle L(s_0) = 180^\circ + 360^\circ (l - 1)$$



● Illustrative Example:  $L(s) = \frac{s + 1}{s(s + 5)[(s + 2)^2 + 4]}$

$$s_0 = -1 + 2j$$

$$\angle L(s_0) = 180^\circ + 360^\circ (l - 1)$$

$$= \sum \psi_i - \sum \phi_i$$

$$= \angle(s_0 + 1)$$

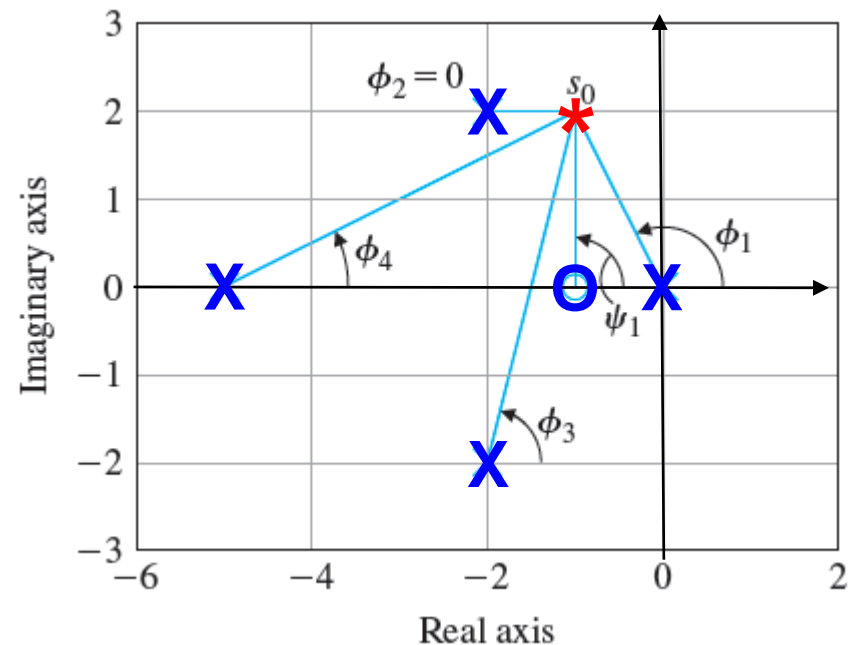
$$- \angle(s_0) - \angle(s_0 + 5)$$

$$- \angle[(s_0 + 2)^2 + 4]$$

$$= 90^\circ - 116.6^\circ - 0^\circ - 76^\circ - 26.6^\circ$$

$$= -129.2^\circ \neq 180^\circ$$

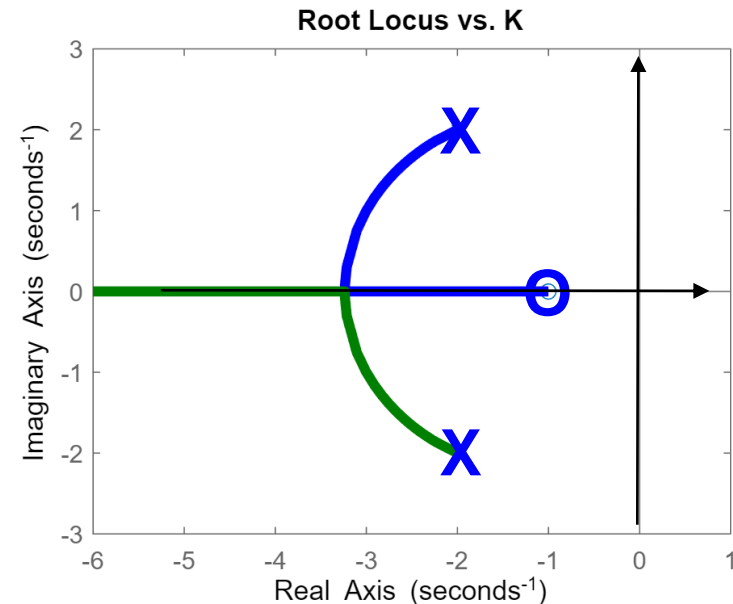
⇒  $s_0$  is not on the root locus



$$L(s) = \frac{s + 1}{s^2 + 4s + 8}$$

$$1 + K \frac{s + 1}{s^2 + 4s + 8} = 0$$

$$(s^2 + 4s + 8) + K(s + 1) = 0$$



● Rule 1:

- The  $n$  branches of the locus start at the poles of  $L(s)$  and
- $m$  of these branches end on the zeros of  $L(s)$ .

●  $a(s) + K b(s) = 0$ ,

● If  $K = 0$ , then  $a(s) = 0$ , whose roots are the poles.

● When  $K \rightarrow \infty$ , then  $b(s) = 0$  ( $m$  zeros) or  $s \rightarrow \infty$ . (the rest  $n-m$ )

$$L(s) = \frac{s + 1}{s^2 + 4s + 8}$$

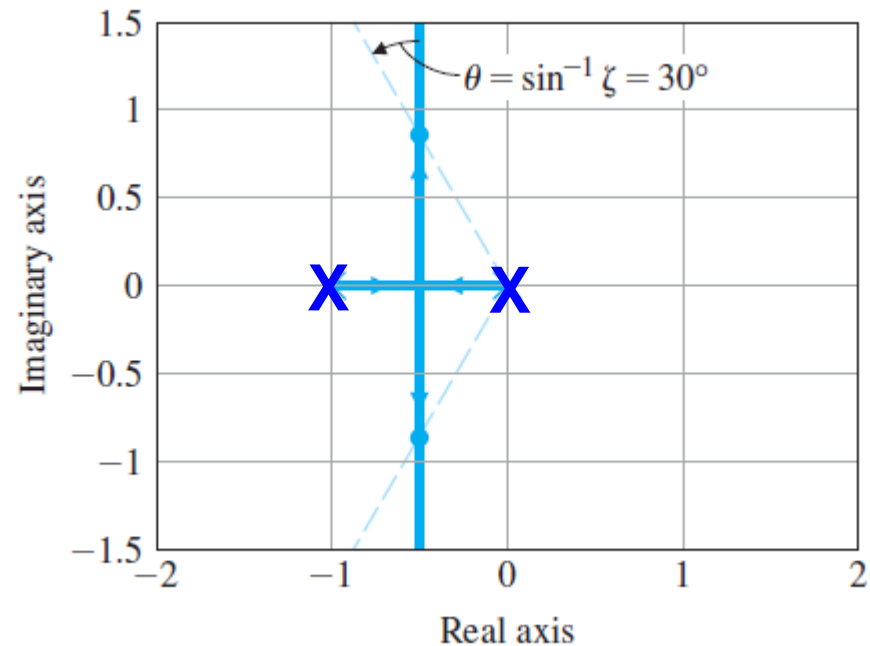
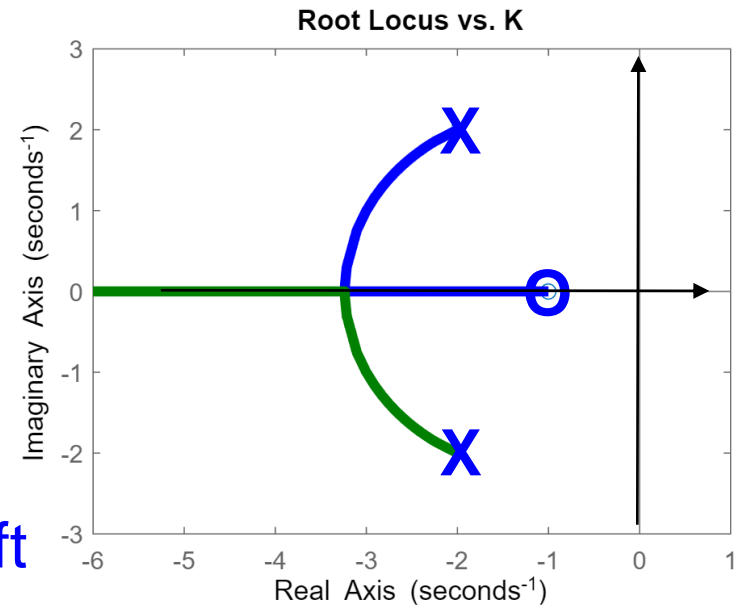
$$1 + K \frac{s + 1}{s^2 + 4s + 8} = 0$$

$$(s^2 + 4s + 8) + K(s + 1) = 0$$

● Rule 2:

- The loci are on the real axis to the left of an odd number of poles and zeros.

$$L(s) = \frac{1}{s(s + 1)}$$



● Rule 3:

● For large  $s$  and  $K$ ,

$n-m$  branches of the loci are asymptotic to lines at angles  $\phi$  radiating out

from the point  $s = \alpha$  on the real axis,

$$\phi_l = \frac{180^\circ + 360^\circ (l - 1)}{n - m}$$

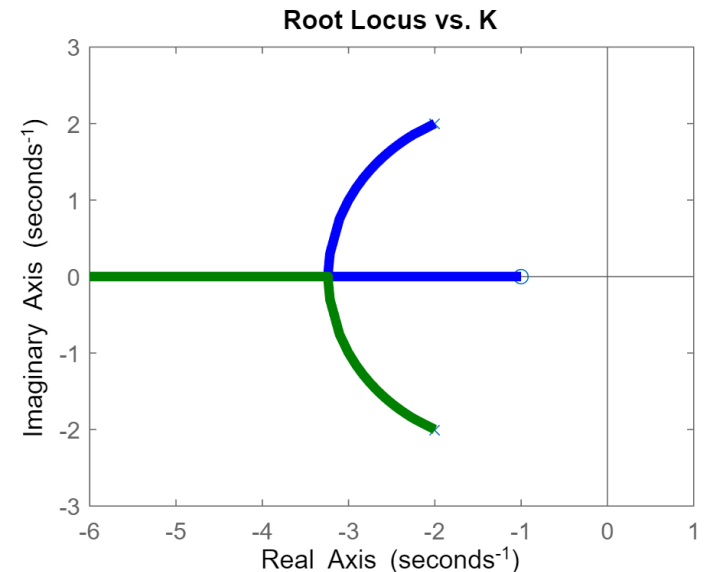
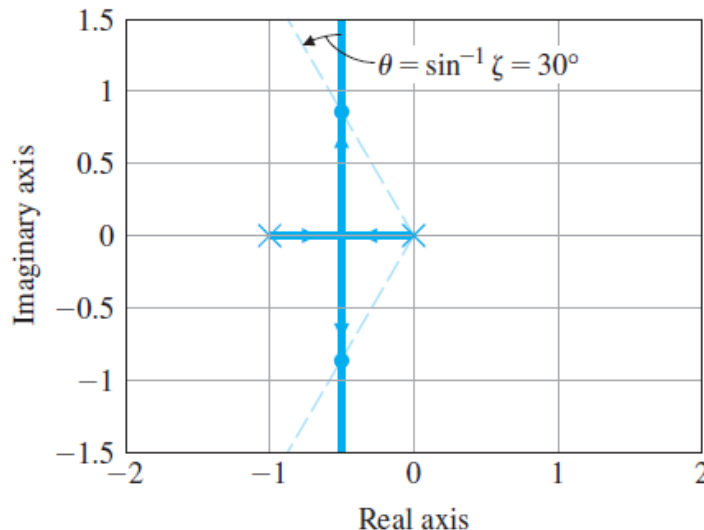
$$\alpha = \frac{\sum p_i - \sum z_i}{n - m}$$

$$L(s) = \frac{s + 1}{s^2 + 4s + 8}$$

$$L(s) = \frac{1}{s(s + 1)}$$

where

$$l = 1, 2, \dots, n - m$$



● Rule 3:

● As  $K \rightarrow \infty$ ,  $L(s) = -\frac{1}{K} \Rightarrow L(s) = 0$

$$L(s) = \frac{s + 1}{s^2 + 4s + 8}$$

1)  $m$  roots will be found to approach the zeros of  $L(s)$

2)  $s \rightarrow \infty$  because  $n \geq m$  that is,

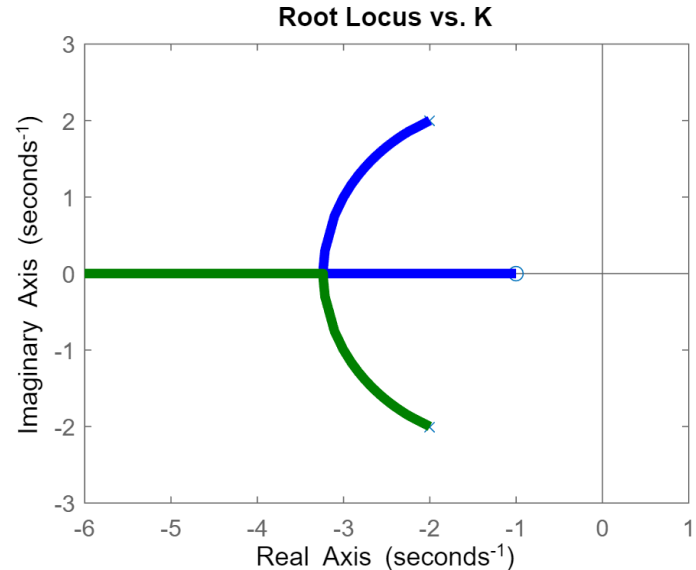
$n - m$  roots approach  $s \rightarrow \infty$

$$\Rightarrow 1 + K \frac{b(s)}{a(s)} = 0$$

$$\Rightarrow 1 + K \frac{s^m + b_1 s^{m-1} + b_2 s^{m-2} + \dots + b_{m-1} s + b_m}{s^n + a_1 s^{n-1} + a_2 s^{n-2} + \dots + a_{n-1} s + a_n} = 0$$

● Can be approximated by

$$\Rightarrow 1 + K \frac{1}{(s - \alpha)^{n-m}} = 0$$





● Rule 3:

$$\Rightarrow 1 + K \frac{1}{(s - \alpha)^{n-m}} = 0$$

● The search point:  $s_0 = R e^{j\phi}$   $l = 1, 2, \dots, (n - m)$

$$(n - m) \phi_l = 180^\circ + 360^\circ (l - 1)$$

$$\Rightarrow \phi_l = \frac{180^\circ + 360^\circ (l - 1)}{(n - m)}$$

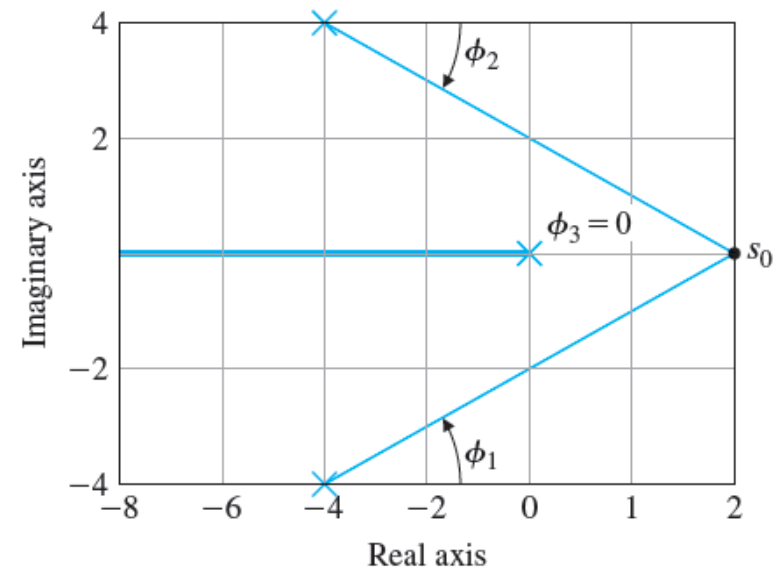
● For this example:

$$L(s) = \frac{1}{s [(s + 4)^2 + 16]}$$

$$(n - m) = 3$$

$$\phi_{1,2,3} = 60^\circ, 180^\circ, 300^\circ,$$

or  $\pm 60^\circ, 180^\circ$



● Rule 3:

$$L(s) = \frac{1}{s [(s + 4)^2 + 16]}$$

● Determine asymptotic lines:

$$= \frac{1}{s^3 + 8s^2 + 32s + 0}$$

$$a(s) = s^n + a_1s^{n-1} + a_2s^{n-2} + \dots + a_{n-1}s + a_n$$

$$= (s - p_1)(s - p_2) \dots (s - p_{n-1})(s - p_n)$$

$$\Rightarrow a_1 = -p_1 - p_2 \dots - p_{n-1} - p_n = -\sum p_i$$

$$b(s) = s^m + b_1s^{m-1} + b_2s^{m-2} + \dots + b_{m-1}s + b_m$$

$$= (s - z_1)(s - z_2) \dots (s - z_{m-1})(s - z_m)$$

$$\Rightarrow b_1 = -z_1 - z_2 \dots - z_{n-1} - z_n = -\sum z_i$$

$$= \frac{s + 1}{(s + 2 + 2j)(s + 2 - 2j)}$$

$$L(s) = \frac{s + 1}{s^2 + 4s + 8}$$

- Rule 3:

$$L(s) = \frac{s + 1}{s [(s + 4)^2 + 16]}$$

- Determine asymptotic lines:

$$\Rightarrow a(s) + K b(s) = 0$$

$$= \frac{s + 1}{s^3 + 8s^2 + 32s + 0}$$

$$(s^3 + 8s^2 + 32s) + K(s + 1) = 0$$

$$\Rightarrow s^n + a_1 s^{n-1} + a_2 s^{n-2} + \dots + a_{n-1} s + a_n$$

$$+ K (s^m + b_1 s^{m-1} + b_2 s^{m-2} + \dots + b_{m-1} s + b_m) = 0$$

$$= (s - r_1) (s - r_2) \dots (s - r_{n-1}) (s - r_n) = 0$$

- If  $m < n - 1$ :

$$\Rightarrow a_1 = -r_1 - r_2 \dots - r_{n-1} - r_n = -\sum r_i$$

- And this term is independent of  $K$

- The open-loop sum and closed-loop sum are the same

and are equal to  $-a_1$

$$\Rightarrow -\sum r_i = -\sum p_i$$

● Rule 3:

$$L(s) = \frac{1}{s [(s + 4)^2 + 16]}$$

● For large values of  $K$ :

- $m$  of the roots  $r_i$  approach the zeros  $z_i$
- $n - m$  of the roots  $r_i$  approach the branches of the asymptotic system

$$\frac{1}{(s - \alpha)^{n-m}} \quad \text{whose poles add up to } (n - m) \alpha$$

$$\Rightarrow -\sum r_i = -(n - m)\alpha - \sum z_i = -\sum p_i$$

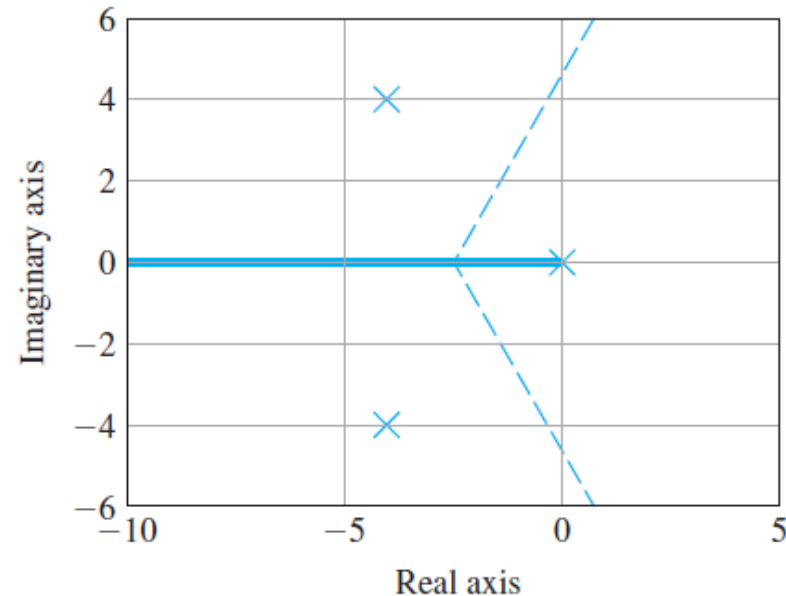
$$\Rightarrow \alpha = \frac{\sum p_i - \sum z_i}{(n - m)}$$

$$2.67 \times \sqrt{3} = 4.62$$

● For this example:

$$\begin{aligned} \Rightarrow \alpha &= \frac{-4 - 4 + 0}{3 - 0} \\ &= -\frac{8}{3} = -2.67 \end{aligned}$$

$$\phi_{1,2,3} = \pm 60^\circ, 180^\circ$$



# Rules for Determining a Positive (180°) Root Locus

● Rule 3:

● For large values of  $K$ :

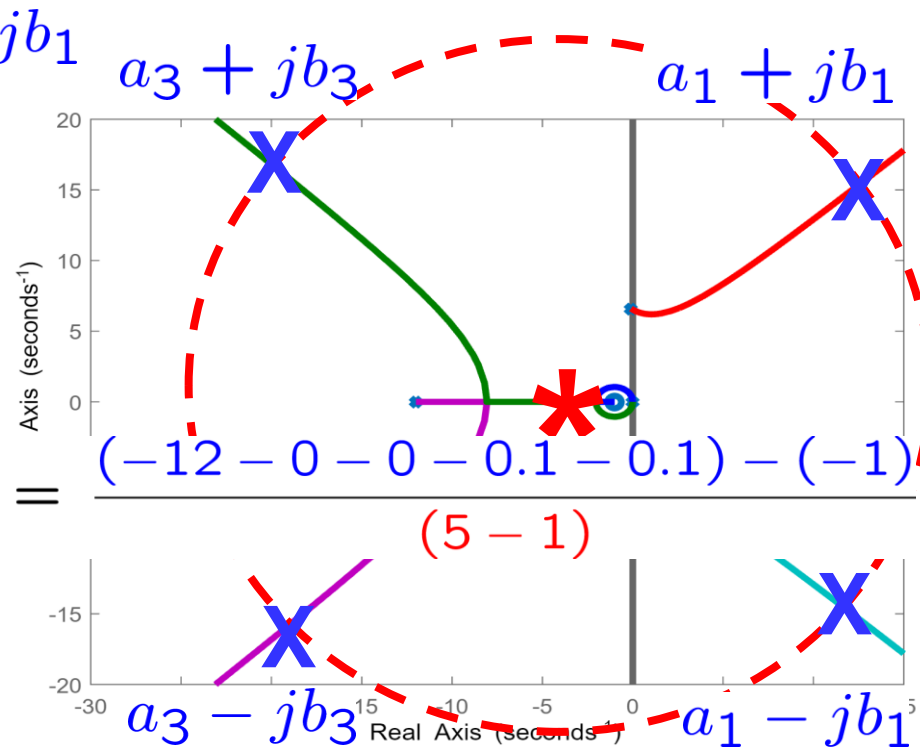
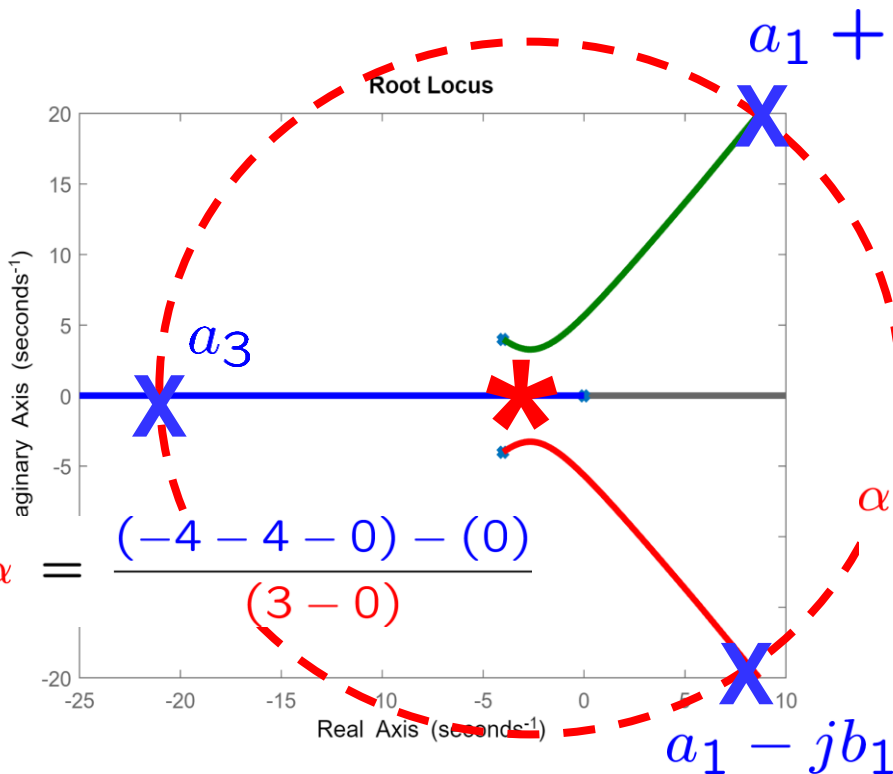
●  $m$  of the roots  $r_i$  approach the zeros  $z_i$

●  $n - m$  of the roots  $r_i$  approach the branches of the asymptotic system

$$\Rightarrow \alpha = \frac{\sum p_i - \sum z_i}{(n - m)}$$

$$L(s) = \frac{1}{s [(s + 4)^2 + 16]}$$

$$\frac{(s + 1)}{(s + 12)} \frac{1}{s^2 [(s + 0.1)^2 + 6.6^2]}$$



- Rule 4: 
$$\sum \psi_i - \sum \phi_i = 180^\circ + 360^\circ (l - 1)$$

- The **angle of departure** of a branch of the locus from a **single pole** is given by

$$\phi_{dep} = \sum \psi_i - \sum_{i \neq dep} \phi_i - 180^\circ - 360^\circ (l - 1)$$

$\sum \phi_i$  the sum of the angles to **the remaining poles**

$\sum \psi_i$  the sum of the angles to **all the zeros**

- The **angle of departure** of a branch of the locus from **repeated poles** with **multiplicity  $q$**  is given by

$$q \phi_{l,dep} = \sum \psi_i - \sum_{i \neq l, dep} \phi_i - 180^\circ - 360^\circ (l - 1)$$

$$l = 1, 2, \dots, q$$

- Rule 4: 
$$\sum \psi_i - \sum \phi_i = 180^\circ + 360^\circ (l - 1)$$
- The **angle of arrival** of a branch at a zero with **multiplicity  $q$**  is given by

$$q \psi_{l,arr} = \sum \phi_i - \sum_{i \neq l, arr} \psi_i + 180^\circ + 360^\circ(l-1)$$

$l = 1, 2, \dots, q$

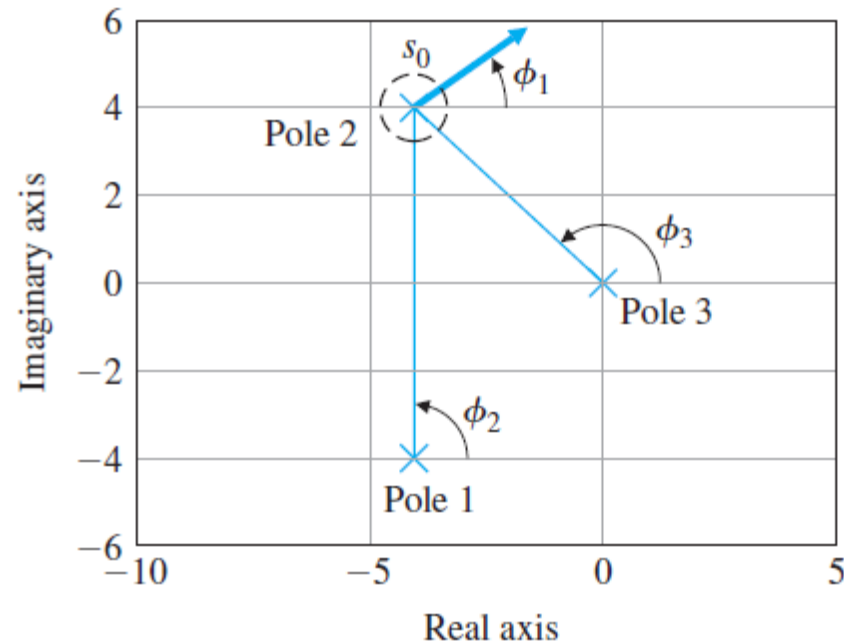
$\sum \phi_i$  the sum of the angles to **all the poles**

$\sum \psi_i$  the sum of the angles to **the remaining zeros**

- For this example:

$$\sum \psi_i - \sum \phi_i = 180^\circ + 360^\circ (l - 1)$$

$$\begin{aligned} \phi_1 &= -90^\circ - 135^\circ - 180^\circ \\ &= -405^\circ \\ &= -45^\circ \end{aligned}$$



- Rule 5:
- The locus can have **multiple roots** at points on the locus and the branches will **approach** a point of  **$q$  roots** at angles **separated** by 
$$\frac{180^\circ - 360^\circ(l - 1)}{q}$$
- And will **depart** at angles with **same separation**.

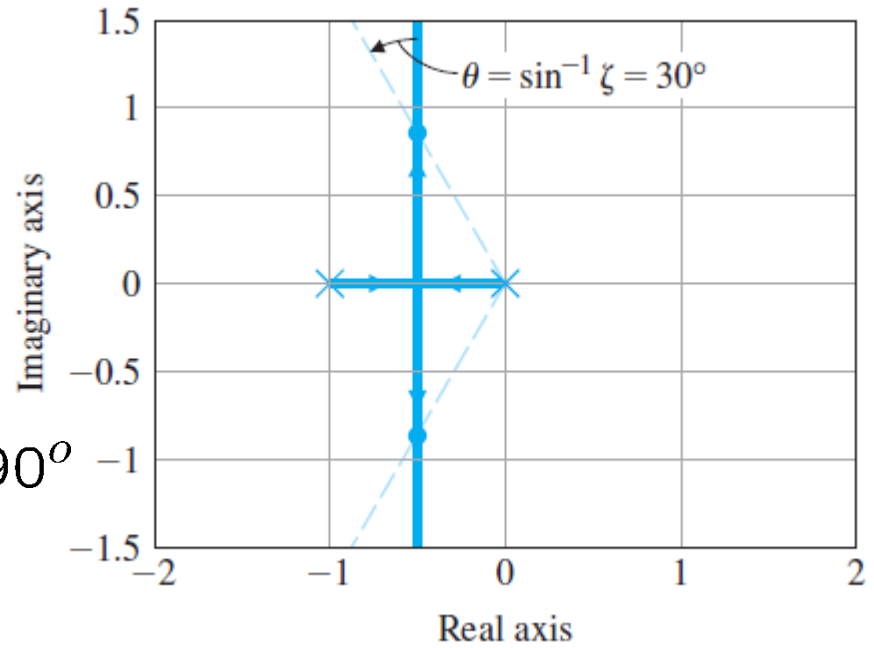
● For example:

$$L(s) = \frac{b(s)}{a(s)} = \frac{1}{s(s + 1)}$$

$$K = \frac{1}{4} \Rightarrow s_{1,2} = -\frac{1}{2}$$

$$\text{sep.} = \frac{180^\circ - 360^\circ(l - 1)}{2} = 90^\circ$$

$$\Rightarrow 0^\circ, 180^\circ \Rightarrow +90^\circ, -90^\circ$$





- Rule 5:

- Continuation Locus:

$$L(s) = \frac{b(s)}{a(s)} = \frac{1}{s(s+1)}$$

$$K_1 = \frac{1}{4} \Rightarrow K = K_1 + K_2 = \frac{1}{4} + K_2$$

$$\Rightarrow s^2 + s + \frac{1}{4} + K_2 = 0$$

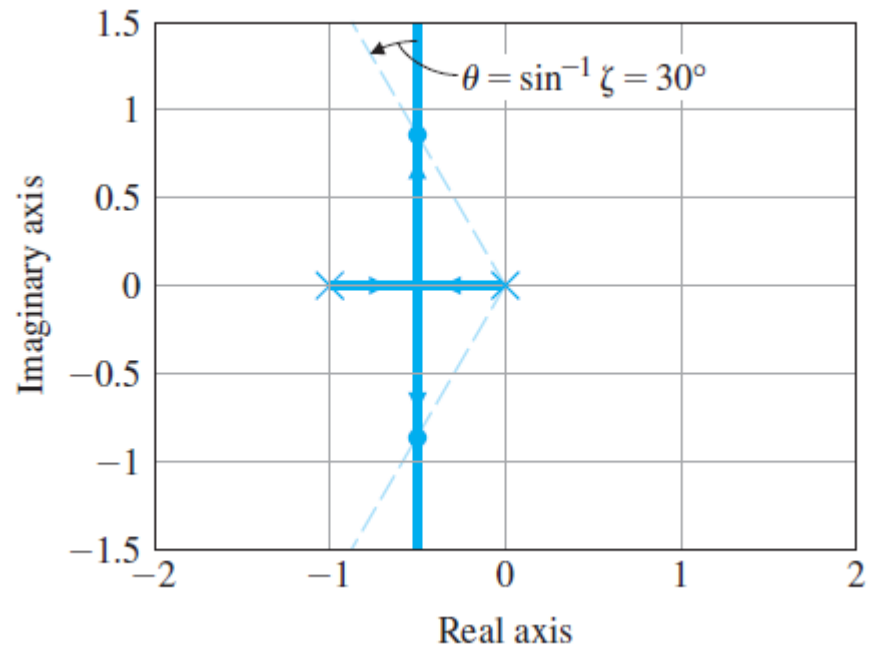
$$\Rightarrow \left(s + \frac{1}{2}\right)^2 + K_2 = 0$$

$$K_2 = 0 \Rightarrow s_{1,2} = -\frac{1}{2}$$

$$2 \phi_{dep} = -180^\circ - 360^\circ(l-1)$$

$$\phi_{dep} = \pm 90^\circ$$

$$\phi_{arr} = 0^\circ, 180^\circ$$



- The third-order example:

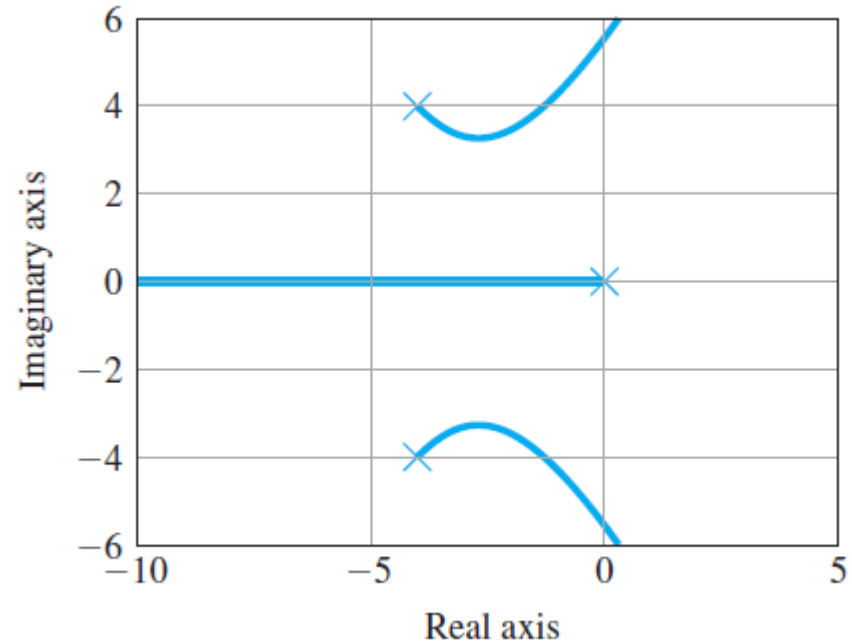
$$L(s) = \frac{1}{s(s^2 + 8s + 32)}$$

```
s = tf('s')
```

```
sysL = (1)/(s*(s^2+8*s+32));
```

```
sysL = 1/(s*(s+4)^2+16);
```

```
rlocus(sysL);
```



- Rule 1:

- The  $n$  branches of the locus start at the poles of  $L(s)$  and
- $m$  of these branches end on the zeros of  $L(s)$ .

- Rule 2:

- The loci are on the real axis to the left of an odd number of poles and zeros.

- Rule 3:

- For large  $s$  and  $K$ ,  
 $n-m$  branches of the loci are asymptotic to lines at angles  $\phi$  radiating out from the point  $s = \alpha$  on the real axis, where

$$\phi_l = \frac{180^\circ + 360^\circ (l - 1)}{n - m} \quad \alpha = \frac{\sum p_i - \sum z_i}{n - m}$$

$l = 1, 2, \dots, n - m$

- Rule 4: 
$$\sum \psi_i - \sum \phi_i = 180^\circ + 360^\circ (l - 1)$$

- The **angle of departure** of a branch of the locus from **repeated poles** with **multiplicity  $q$**  is given by

$$q \phi_{l,dep} = \sum \psi_i - \sum_{i \neq l, dep} \phi_i - 180^\circ - 360^\circ(l-1)$$

$l = 1, 2, \dots, q$

- The **angle of arrival** of a branch at a zero with **multiplicity  $q$**  is given by

$$q \psi_{l,arr} = \sum \phi_i - \sum_{i \neq l, arr} \psi_i + 180^\circ + 360^\circ(l-1)$$

- Rule 5:

- The locus can have **multiple roots** at points on the locus and the branches will **approach** a point of  **$q$  roots** at angles **separated** by

$$\frac{180^\circ - 360^\circ(l - 1)}{q}$$

- And will **depart** at angles with same **separation**.

- The **positive root locus** is a plot of all possible locations for roots to the equation  $1 + K L(s) = 0$  for some **real positive value** of **K**.
- The purpose of **design** is to select **a particular value** of **K** that will meet the **specifications** for **static and dynamic** response.

# Selecting the Parameter Value

$$L(s) = \frac{1}{s [(s + 4)^2 + 16]}$$

$$L(s_0) = \frac{1}{s_0 (s_0 - s_2) (s_0 - s_3)}$$

$$K = \frac{1}{|L(s_0)|}$$

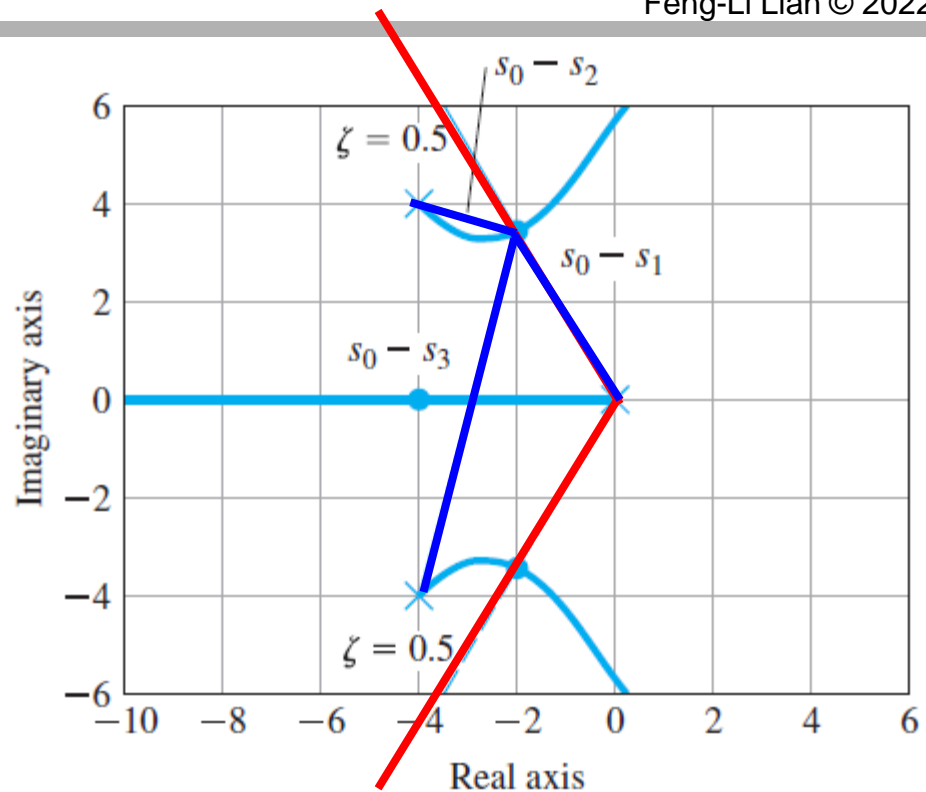
$$= |s_0| |s_0 - s_2| |s_0 - s_3|$$

$$\approx 4.0 \times 2.1 \times 7.7$$

$$\approx 65$$

$$\Rightarrow 1 + K L(s) = 0$$

$$\Rightarrow L(s) = -\frac{1}{K}$$



```
s = tf('s')
```

```
sysL = (1)/(s*(s^2+8*s+32));
```

```
sysL = (1)/(s*((s+4)^2+16));
```

```
rlocus(sysL);
```

```
[K, p] = rlocfind(sysL);
```

- Compute the **error constant** of the control system
- For example,  
the **steady-state error** in tracking a ramp input  
is giving by **the velocity constant**:

$$K_v = \lim_{s \rightarrow 0} s K L(s)$$

$$K_v = \lim_{s \rightarrow 0} s L(s)$$

$$= \lim_{s \rightarrow 0} s K \frac{1}{s [(s + 4)^2 + 16]}$$

$$= \frac{K}{32}$$

$$= \frac{65}{32} \approx 2 \text{ sec}^{-1}$$