Fall 2022 (111-1)

控制系統 Control Systems

Unit 3F Stability

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Stability

Stability:

- An LTI system is said to be stable
- if all the roots of the transfer function denominator polynomial
- have negative real parts

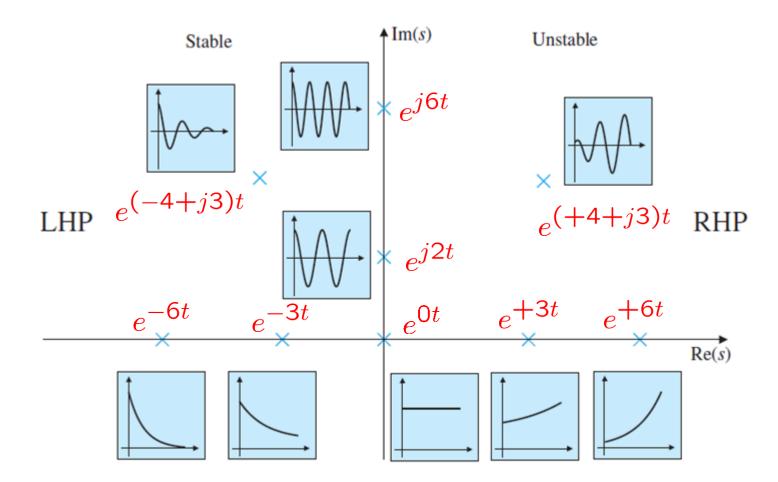
(that is, they are all in the left-hand s-plane)

and is unstable otherwise.

- Stable System:
 - A system is stable
 - if its initial conditions decay to zero
 - and is unstable if they diverge.

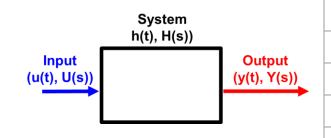
Stability

Time functions associated with points in the s-plane (LHP, left half-plane; RHP, right half-plane)



Bounded Input-Bounded Output Stability (BIBO Stable)

- A system is said to have BIBO stability
- if every bounded input results in a bounded output (regardless of what goes on inside the system).

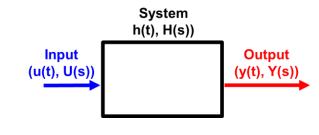


If the system has

input u(t), output y(t), and impulse response h(t), then

$$y(t) = \int_{-\infty}^{\infty} h(t) u(t-\tau) d\tau$$

$$y(t) = \int_{-\infty}^{\infty} h(t) \ u(t-\tau) \ d\tau$$



- If input u(t) is bounded $|u(\cdot)| \leq M < \infty$
- And output y(t) is bounded by

$$|\mathbf{y}| = \left| \int_{-\infty}^{\infty} h \, \mathbf{u} \, d\tau \right| \qquad \leq \int_{-\infty}^{\infty} |h| \, |\mathbf{u}| \, d\tau \qquad \leq M \int_{-\infty}^{\infty} |h| \, d\tau$$

That is, output y(t) is bounded if

 $\int_{-\infty}^{\infty} |h| d\tau \quad \text{ is bounded.}$

Bounded Input-Bounded Output Stability

y(t)

suppose

Output

(y(t), Y(s))

System h(t), H(s))

is bounded,

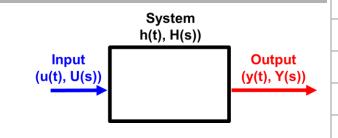
$$y(t) = \int_{-\infty}^{\infty} h(t) u(t - \tau) d\tau$$

$$u(t) = \int_{-\infty}^{\infty} h(t) u(t - \tau) d\tau$$
• On the other hand,
suppose the integral is not bounded and the input
• e.g., $u(t-\tau) = +1$ if $h(\tau) > 0$
and $u(t-\tau) = -1$ if $h(\tau) < 0$
$$|y| = \int_{-\infty}^{\infty} |h(\tau)| d\tau$$

The output y(t) is not bounded.

Bounded Input-Bounded Output Stability

$$y(t) = \int_{-\infty}^{\infty} h(t) u(t-\tau) d\tau$$



Mathematical Definition of BIBO Stability

- The system with impulse response h(t) is BIBO stable
- if and only if
 - the integral

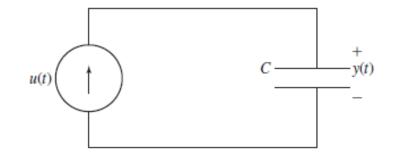
$$\int_{-\infty}^{\infty} |h(au)| \ d au \ < \ {oldsymbol \infty}$$

Bounded Input-Bounded Output Stability



Example 3.31: BIBO Stability for a Capacitor

$$\Rightarrow h(\tau) = 1(t)$$



• Capacitor driven by current source

$$\int_{-\infty}^{\infty} |h(au)| \, d au \ = \ \int_{-\infty}^{\infty} \, d au \ o \ \infty$$

Consider the LTI whose transfer function:

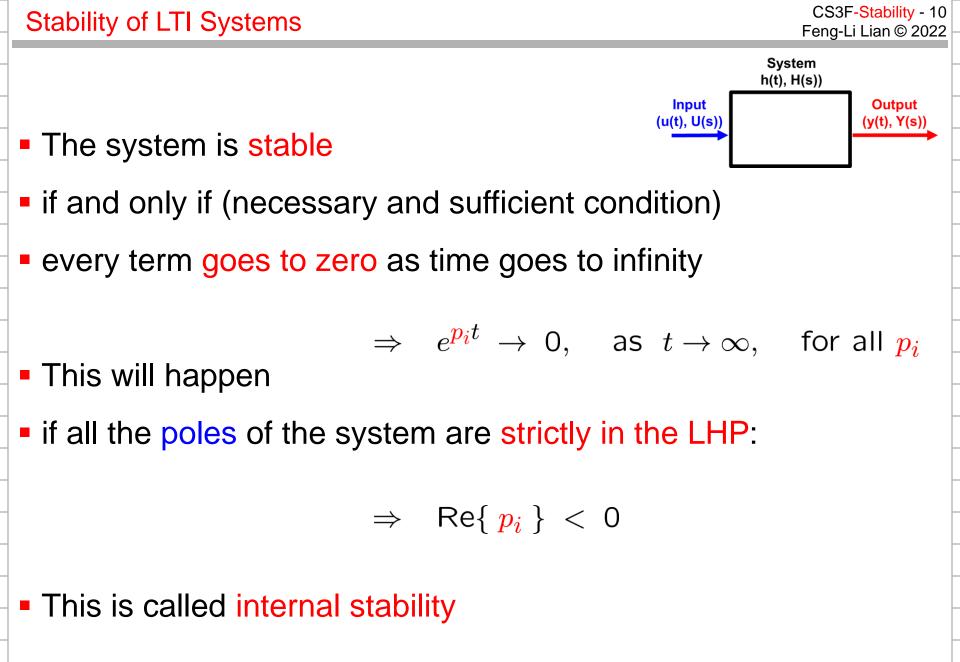
$$T(s) = \frac{Y(s)}{R(s)} = \frac{b_0 s^m + b_1 s^{m-1} + \dots + b_m}{s^n + a_1 s^{n-1} + \dots + a_n} = \frac{b(s)}{a(s)}$$

$$= K \frac{\prod_{i=1}^{m} (s - z_i)}{\prod_{i=1}^{n} (s - p_i)}, \quad m \le n$$

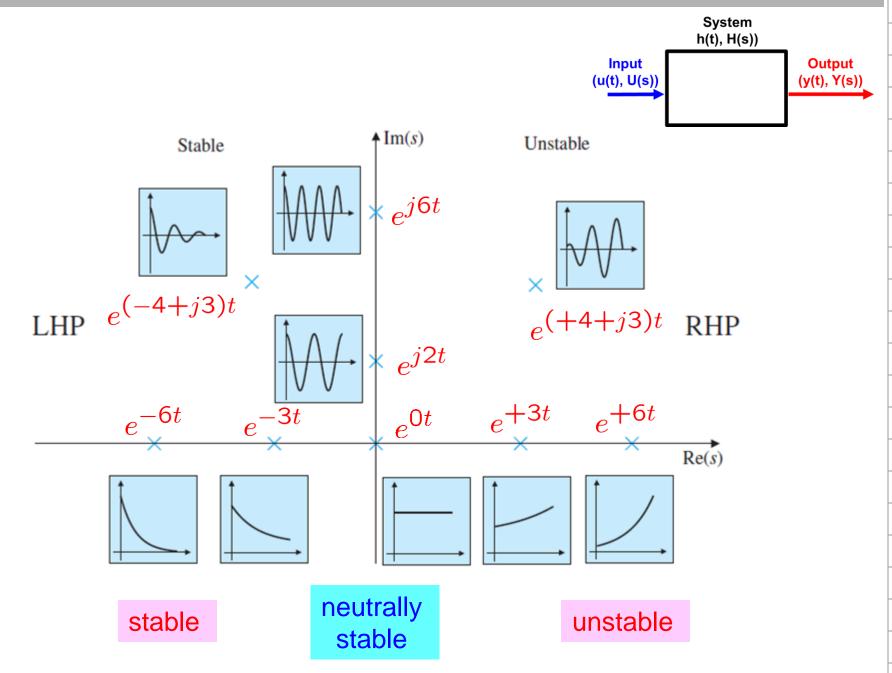
$$\Rightarrow y(t) = \sum_{i=1}^{n} K_i e^{p_i t}$$

p_i are the roots of a(s), denominator polynomial

K_i depend on the initial condition and zero locations



Stability of LTI Systems



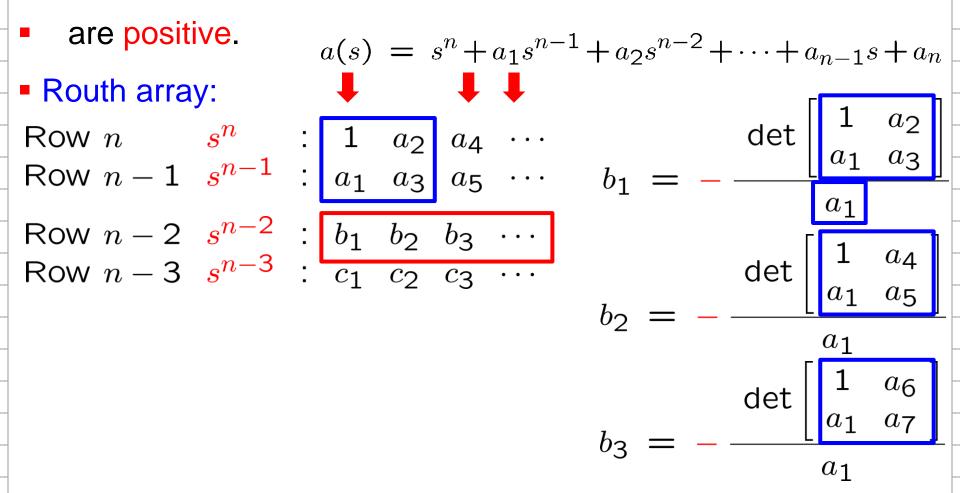
Consider the characteristic equation of an nth-order system:

$$a(s) = s^{n} + a_{1}s^{n-1} + a_{2}s^{n-2} + \dots + a_{n-1}s + a_{n}$$

A necessary (but not sufficient) condition for stability of the system is that
all of the roots have negative real parts
which in turn requires that all the { a_i } be positive.

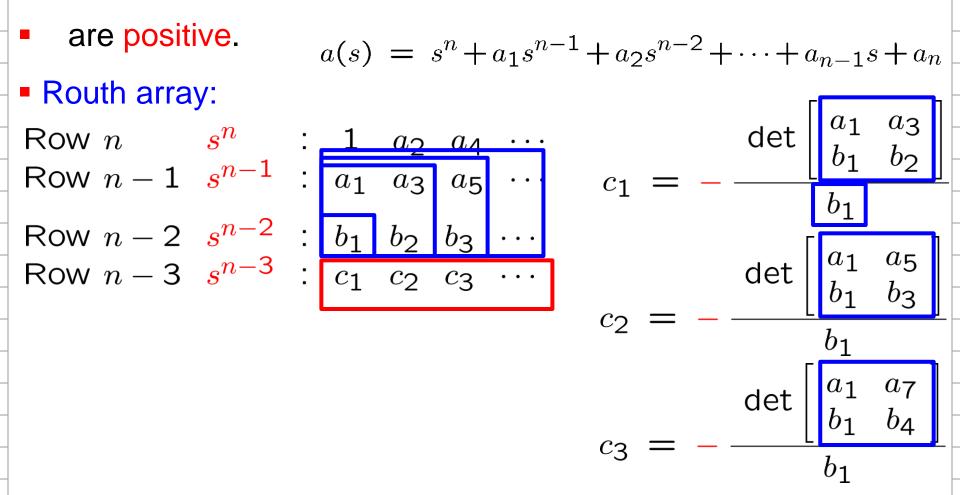
 Equivalent test were independently proposed by Routh in 1874 and Hurwitz in 1895. Routh showed that

- a system is stable if and only if
- all the elements in the first column of the Routh array



Routh showed that

- a system is stable if and only if
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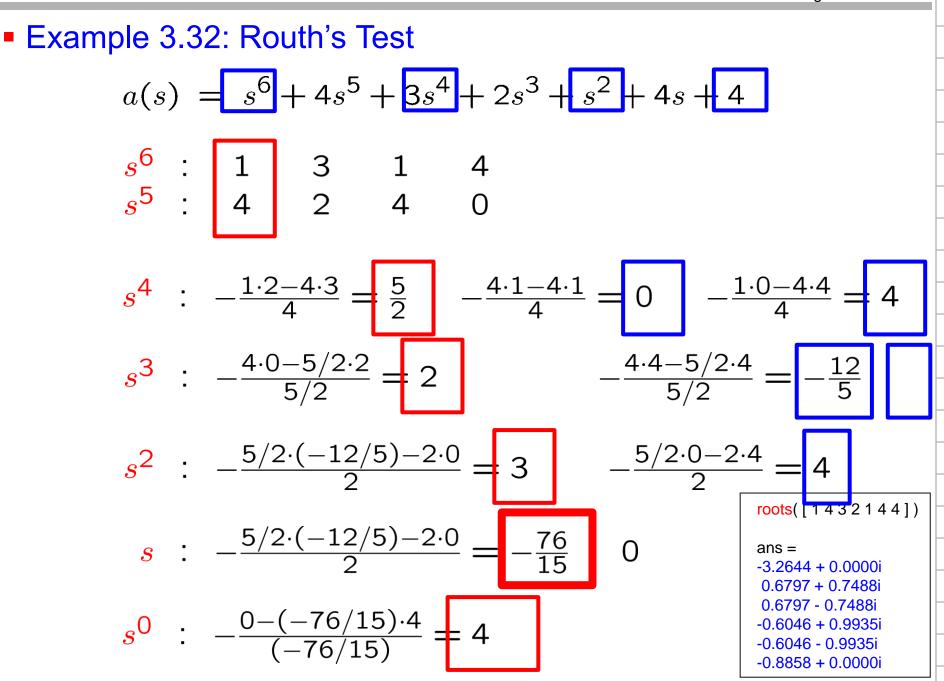
Routh showed that

- a system is stable if and only if
- all the elements in the first column of the Routh array

are positive.
a(s) = sⁿ + a₁sⁿ⁻¹ + a₂sⁿ⁻² + ··· + a_{n-1}s + a_n
Routh array:

Row n	s^n		1	a_2	a_{4}	
Row $n-1$	s^{n-1}	:	a_1	a_3	a_5	
Row $n-2$ Row $n-3$	s^{n-2} s^{n-3}	:	$b_1 \\ c_1$	b_2 c_2	bз сз	
Row :	: :					
Row 2	<i>s</i> ²	:	* *	*		
Row 1	\boldsymbol{S}	:	*			
Row 0	s^0	:	*			

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- Example 3.33: Stability versus Parameter Range
- A feedback system for testing stability

$$r \xrightarrow{+} \Sigma \xrightarrow{K} \xrightarrow{s+1} \xrightarrow{s(s-1)(s+6)} \xrightarrow{\circ} y$$

• The characteristic equation for the system:

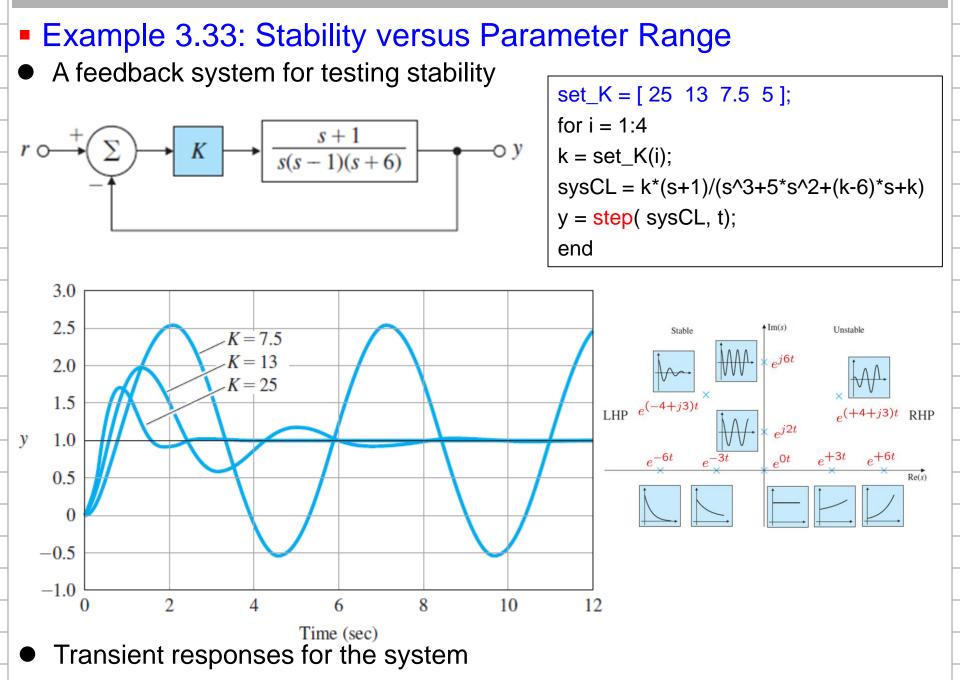
$$1 + K \frac{s+1}{s(s-1)(s+6)} = 0 \quad s^{3} + 5 s^{2} + (K-6) s + K = 0$$

$$s^{3} : \begin{bmatrix} 1 & K-6 \\ s^{2} : 5 & K \end{bmatrix}$$

$$s : \begin{bmatrix} (4K-30)/5 \\ K \end{bmatrix} \Rightarrow \frac{(4K-30)}{5} > 0 \Rightarrow K > 7.5$$

$$\Rightarrow K > 0$$

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- Example 3.34: Stability versus Two Parameter Ranges
- System with proportional-integral (PI) control

$$R \circ \xrightarrow{+} \Sigma \longrightarrow K + \frac{K_I}{s} \longrightarrow 1 \longrightarrow (s+1)(s+2) \longrightarrow Y$$

• The characteristic equation for the system:

$$1 + (K + \frac{K_{I}}{s}) \frac{1}{(s+1)(s+2)} = 0$$

$$s^{3} : \begin{bmatrix} 1 \\ 3 \end{bmatrix} 2 + K$$

$$s^{2} : \begin{bmatrix} 3 \\ K_{I} \end{bmatrix} K_{I}$$

$$k = K > \frac{1}{3}K_{I} - 2$$

$$K_{I} > 0$$

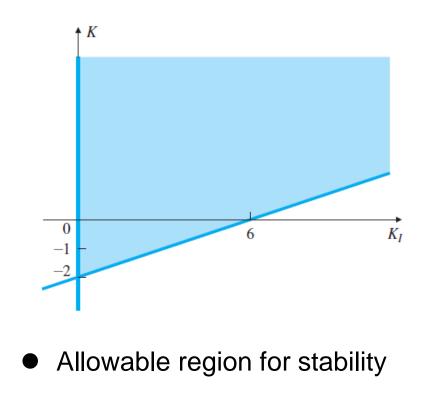
Example 3.34: Stability versus Two Parameter Ranges

• System with proportional-integral (PI) control

$$R \circ \xrightarrow{+} \Sigma \xrightarrow{-} K + \frac{K_l}{s} \xrightarrow{-} (s+1)(s+2) \xrightarrow{-} Y$$

$$\Rightarrow K > \frac{1}{3}K_I - 2$$

 $\Rightarrow K_I > 0$



- Example 3.34: Stability versus Two Parameter Ranges
 - System with proportional-integral (PI) control

