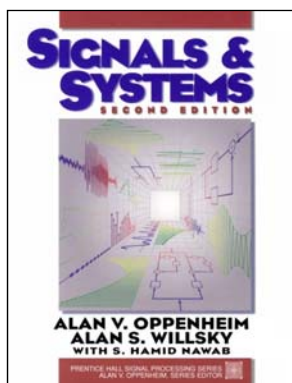


Spring 2015

信號與系統 Signals and Systems

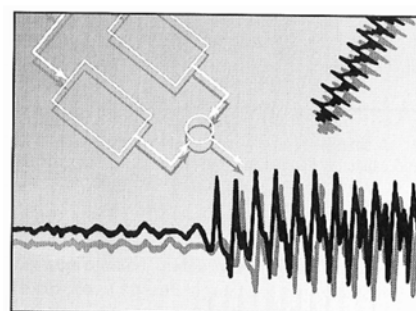
Chapter SS-1 Signals and Systems



Feng-Li Lian

NTU-EE

Feb15 – Jun15



Figures and images used in these lecture notes are adopted from
“Signals & Systems” by Alan V. Oppenheim and Alan S. Willsky, 1997

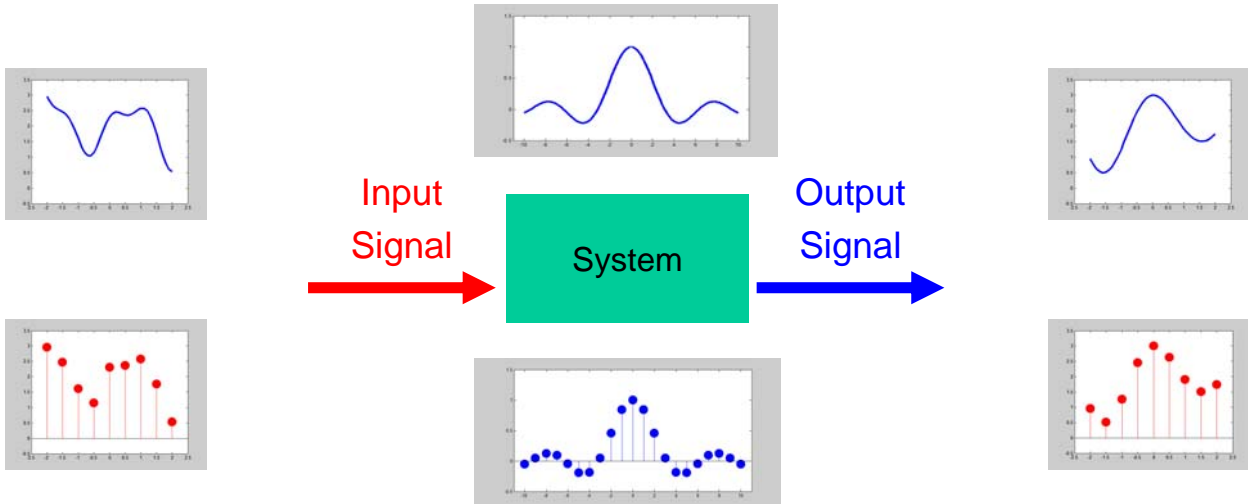
Outline

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NTUEE-SS1-SS-2

- Introduction
- Continuous-Time & Discrete-Time Signals
- Transformations of the Independent Variable
- Exponential & Sinusoidal Signals
- The Unit Impulse & Unit Step Functions
- Continuous-Time & Discrete-Time Systems
- Basic System Properties

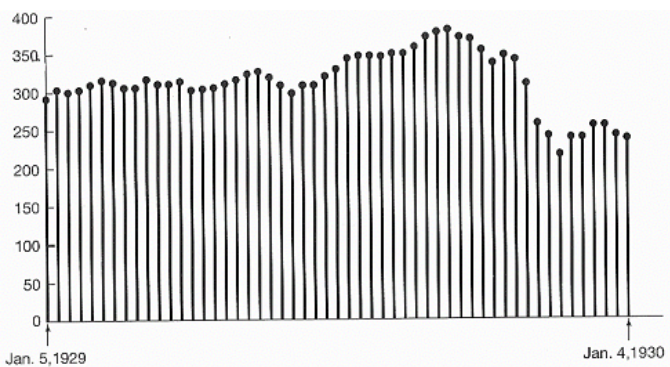
▪ Signals & Systems:

- Is about using **mathematical techniques** to help **describe** and **analyze systems** which process **signals**
 - **Signals** are variables that carry **information**
 - **Systems** process **input signals** to produce **output signals**

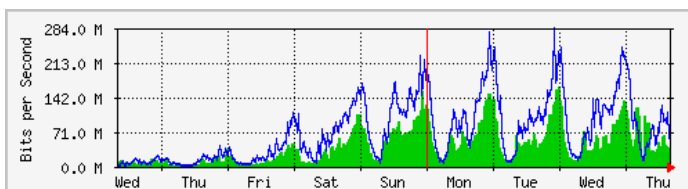


▪ Discrete-Time Signals:

- The weekly Dow-Jones stock market index

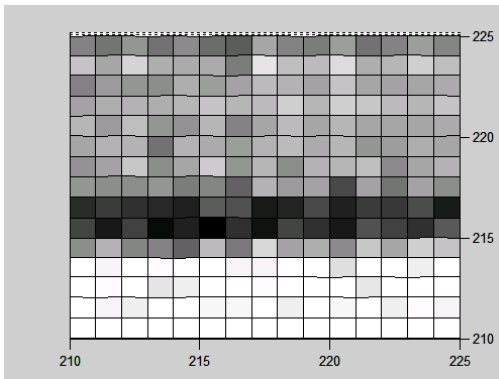
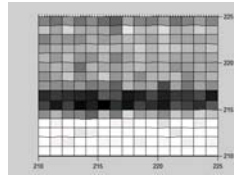
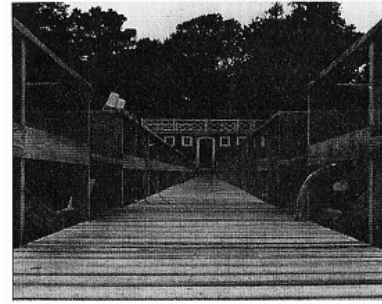


- Boy-One Dormitory, 2/13-2/21, 2013, every 30 min



▪ **Discrete-Time Signals:**

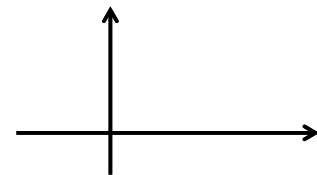
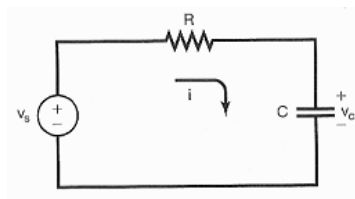
- A monochromatic picture



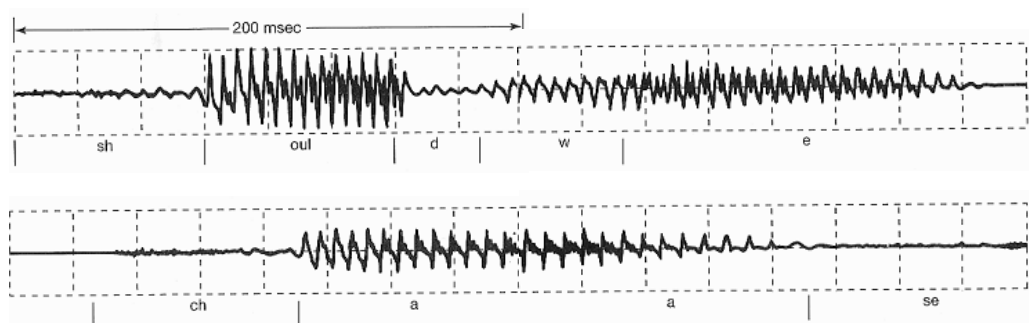
| | | | | | | | | | | | | | |
|-----|-----|-----|-----|-----|-----|-----|-----|-----|-----|-----|-----|-----|-----|
| 117 | 151 | 112 | 136 | 110 | 95 | 166 | 131 | 125 | 159 | 114 | 130 | 159 | 133 |
| 164 | 209 | 173 | 171 | 168 | 126 | 225 | 189 | 184 | 219 | 172 | 180 | 206 | 188 |
| 154 | 151 | 143 | 174 | 156 | 160 | 186 | 179 | 163 | 193 | 167 | 161 | 167 | 168 |
| 173 | 179 | 172 | 182 | 195 | 181 | 187 | 205 | 183 | 205 | 196 | 186 | 180 | 193 |
| 155 | 188 | 149 | 145 | 181 | 131 | 166 | 185 | 166 | 181 | 164 | 164 | 166 | 171 |
| 176 | 176 | 113 | 178 | 169 | 159 | 177 | 184 | 171 | 182 | 148 | 159 | 174 | 164 |
| 162 | 198 | 141 | 164 | 202 | 149 | 181 | 142 | 176 | 182 | 189 | 136 | 165 | 176 |
| 143 | 143 | 148 | 127 | 132 | 97 | 177 | 152 | 160 | 74 | 163 | 119 | 162 | 143 |
| 62 | 51 | 40 | 32 | 95 | 82 | 28 | 39 | 75 | 35 | 60 | 58 | 77 | 28 |
| 26 | 67 | 14 | 34 | 3 | 49 | 22 | 69 | 48 | 26 | 81 | 67 | 49 | 91 |
| 176 | 134 | 129 | 97 | 185 | 120 | 212 | 160 | 173 | 139 | 197 | 166 | 207 | 192 |
| 243 | 255 | 250 | 255 | 254 | 242 | 242 | 255 | 254 | 222 | 255 | 253 | 239 | 252 |
| 248 | 255 | 226 | 238 | 255 | 255 | 249 | 255 | 255 | 255 | 230 | 252 | 255 | 255 |
| 243 | 238 | 255 | 253 | 245 | 247 | 255 | 238 | 255 | 244 | 255 | 237 | 249 | 241 |

▪ **Continuous-Time Signals:**

- Source voltage & capacity voltage in a simple RC circuit

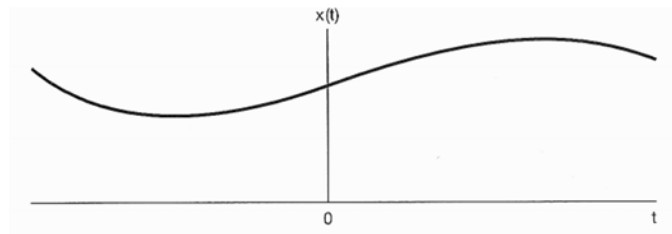


- Recording of a speech signal

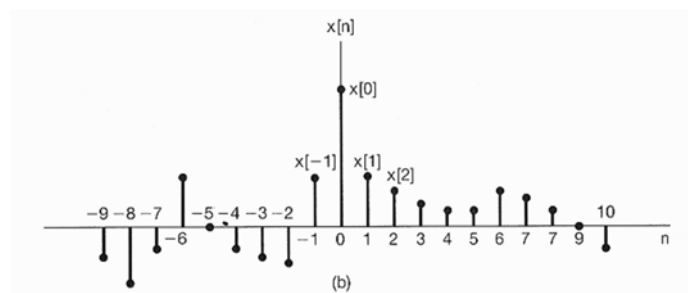


▪ Graphical Representations of Signals:

- Continuous-time signals $x(t)$ or $x_c(t)$



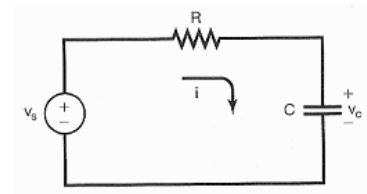
- Discrete-time signals $x[n]$ or $x_d[n]$



▪ Energy & Power of a resistor:

- Instantaneous power

$$p(t) = v(t)i(t) = \frac{1}{R}v^2(t)$$



- Total energy over a finite time interval

$$\int_{t_1}^{t_2} p(t) dt = \int_{t_1}^{t_2} \frac{1}{R} v^2(t) dt$$

- Average power over a finite time interval

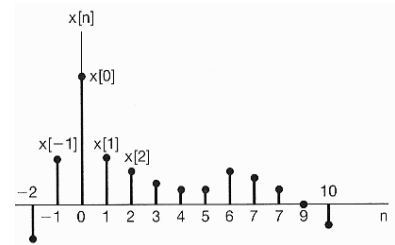
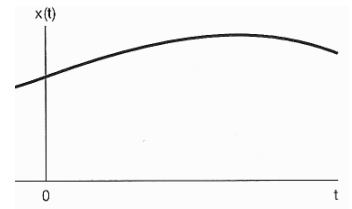
$$\frac{1}{t_2 - t_1} \int_{t_1}^{t_2} p(t) dt = \frac{1}{t_2 - t_1} \int_{t_1}^{t_2} \frac{1}{R} v^2(t) dt$$

▪ **Signal Energy & Power:**

- Total energy over a finite time interval

$$E \triangleq \int_{t_1}^{t_2} |x(t)|^2 dt \quad \text{continuous-time}$$

$$E \triangleq \sum_{n=n_1}^{n_2} |x[n]|^2 \quad \text{discrete-time}$$



- Time-averaged power over a finite time interval

$$P \triangleq \frac{1}{t_2 - t_1} \int_{t_1}^{t_2} |x(t)|^2 dt \quad \text{continuous-time}$$

$$P \triangleq \frac{1}{n_2 - n_1 + 1} \sum_{n=n_1}^{n_2} |x[n]|^2 \quad \text{discrete-time}$$

▪ **Signal Energy & Power:**

- Total energy over an infinite time interval

$$E_\infty \triangleq \lim_{T \rightarrow \infty} \int_{-T}^T |x(t)|^2 dt = \int_{-\infty}^{+\infty} |x(t)|^2 dt$$

$$E_\infty \triangleq \lim_{N \rightarrow \infty} \sum_{n=-N}^{+N} |x[n]|^2 = \sum_{n=-\infty}^{+\infty} |x[n]|^2$$

- Time-averaged power over an infinite time interval

$$P_\infty \triangleq \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt$$

$$P_\infty \triangleq \lim_{N \rightarrow \infty} \frac{1}{2N + 1} \sum_{n=-N}^{+N} |x[n]|^2$$

Three Classes of Signals:

$$E_{\infty} = \lim_{T \rightarrow \infty} \int_{-T}^T |x(t)|^2 dt$$

$$P_{\infty} = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt$$

- Finite total energy & zero average power

$$0 \leq E_{\infty} < \infty \Rightarrow P_{\infty} = \lim_{T \rightarrow \infty} \frac{E_{\infty}}{2T} = 0$$

- Finite average power & infinite total energy

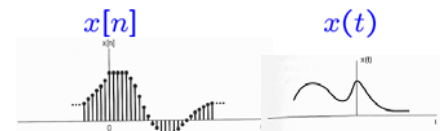
$$0 \leq P_{\infty} < \infty \Rightarrow E_{\infty} = \infty \text{ (if } P_{\infty} > 0 \text{)}$$

- Infinite average power & infinite total energy

$$P_{\infty} = \infty \ \& \ E_{\infty} = \infty$$

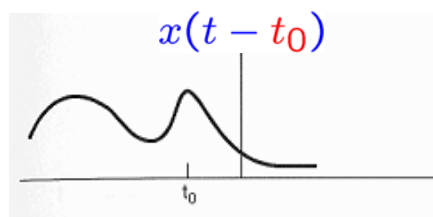
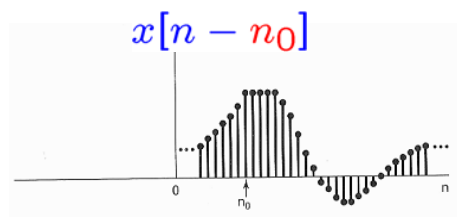
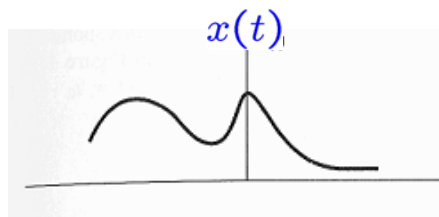
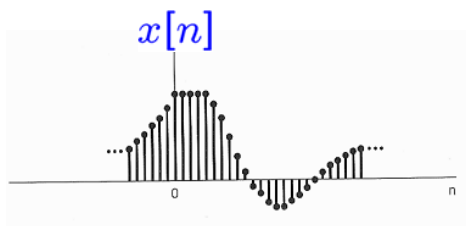
Outline

- Introduction
- Continuous-Time & Discrete-Time Signals
- Transformations of the Independent Variable
 - Time Shift
 - Time Reversal
 - Time Scaling
 - Periodic Signals
 - Even & Odd Signals
- Exponential & Sinusoidal Signals
- The Unit Impulse & Unit Step Functions
- Continuous-Time & Discrete-Time Systems
- Basic System Properties



▪ Time Shift:

$$\begin{cases} n_0, t_0 > 0 : & \text{delay} \\ n_0, t_0 < 0 : & \text{advance} \end{cases}$$



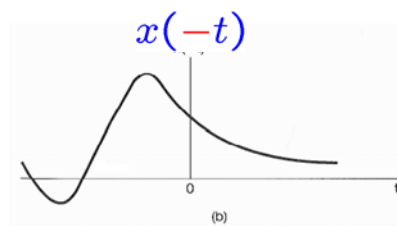
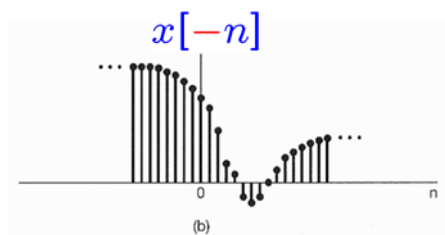
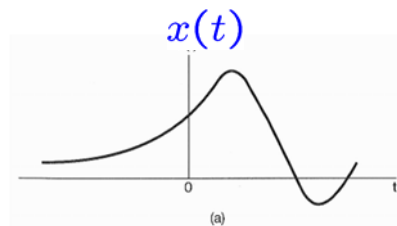
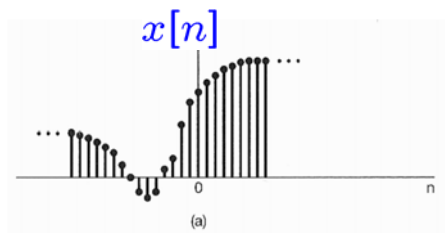
$n_0 > 0$

$t_0 < 0$

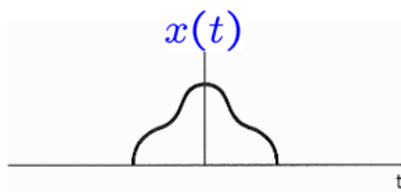
$x[n - 8]$

$x(t + 5)$

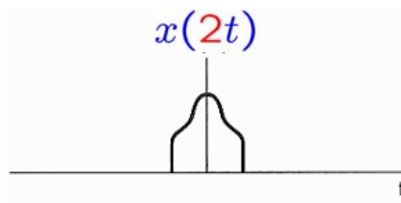
▪ Time Reversal:



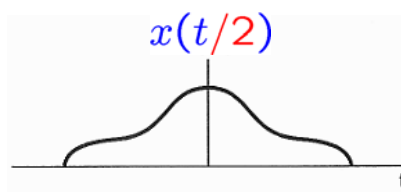
▪ Time Scaling:



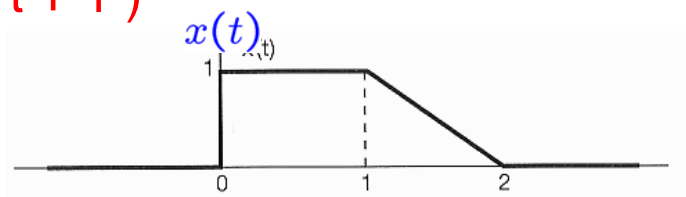
$t \rightarrow 2t$



$t \rightarrow t/2$



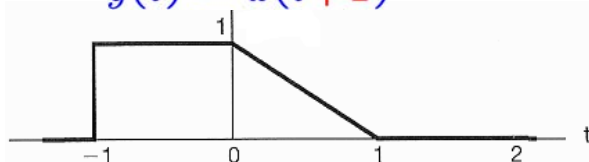
▪ $x(t) \rightarrow x(-t+1)$



$t \rightarrow t+1$



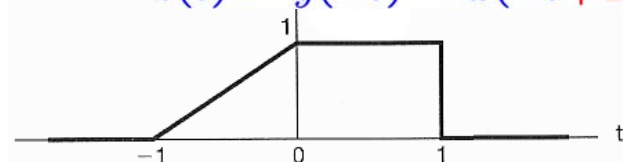
$y(t) = x(t+1)$



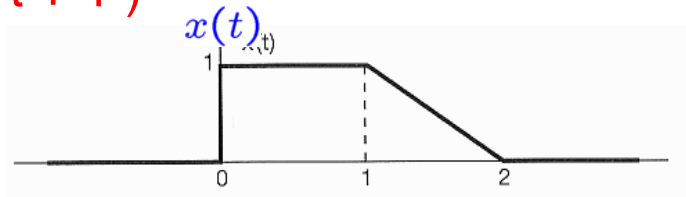
$t \rightarrow -t$



$z(t) = y(-t) = x(-t+1)$



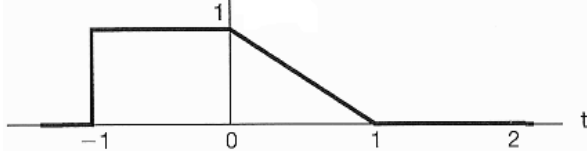
$x(t) \rightarrow x(-t+1)$



$t \rightarrow t+1$



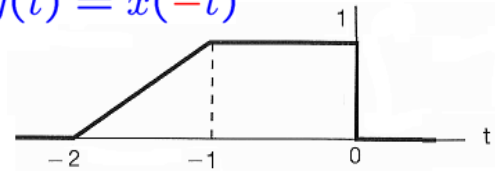
$y(t) = x(t+1)$



$t \rightarrow -t$



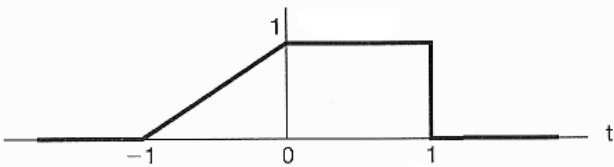
$y(t) = x(-t)$



$t \rightarrow -t$



$z(t) = y(-t) = x(-t+1)$

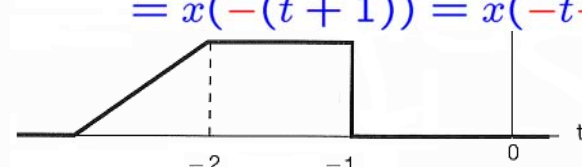


$t \rightarrow t+1$

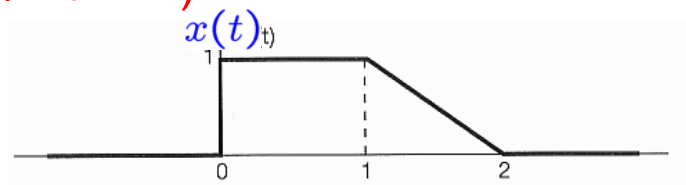


$z(t) = y(t+1)$

$= x(-(t+1)) = x(-t-1)$



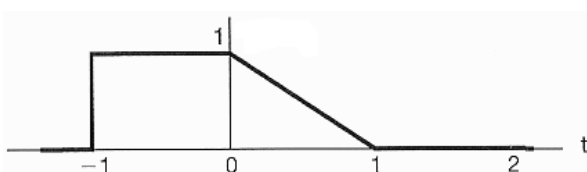
$x(t) \rightarrow x(3/2 t + 1)$



$t \rightarrow t+1$



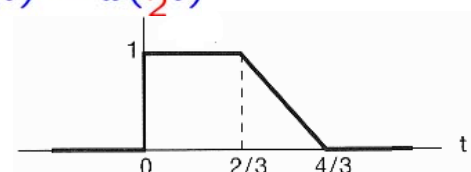
$y(t) = x(t+1)$



$t \rightarrow \frac{3}{2}t$



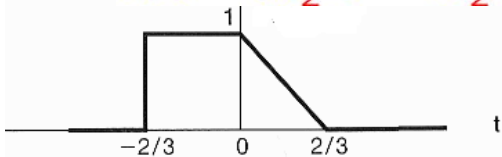
$y(t) = x(\frac{3}{2}t)$



$t \rightarrow \frac{3}{2}t$



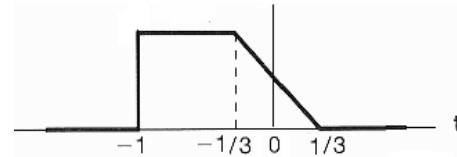
$z(t) = y(\frac{3}{2}t) = x(\frac{3}{2}t+1)$



$t \rightarrow t+1$



$z(t) = y(t+1) = x(\frac{3}{2}t+\frac{3}{2})$

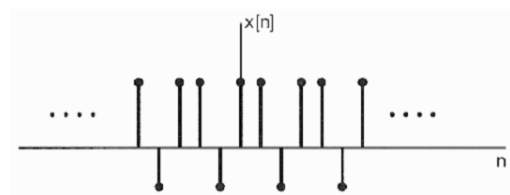
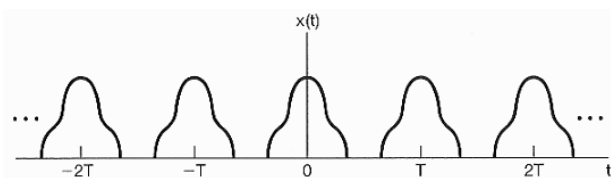


$$\blacksquare x(t) \rightarrow x(at - b)$$

- $|a| < 1$: linearly stretched
- $|a| > 1$: linearly compressed
- $a < 0$: time reversal
- $b > 0$: delayed time shift
- $b < 0$: advanced time shift

 \blacksquare Problems:

- P1.21 for CT
- P1.22 for DT

 \blacksquare CT & DT Periodic Signals:


$$N = 3$$

$$x(t) = x(t + T) \quad \text{for } T > 0 \text{ and all values of } t$$

$$x[n] = x[n + N] \quad \text{for } N > 0 \text{ and all values of } n$$

▪ Periodic Signals:

$$x(t) = x(t + T) \quad \text{for } T > 0 \text{ and all values of } t$$

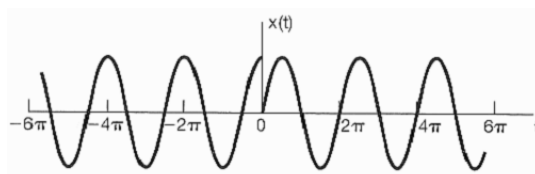
$$x[n] = x[n + N] \quad \text{for } N > 0 \text{ and all values of } n$$

- A periodic signal is **unchanged** by a **time shift** of **T** or **N**
- They are also **periodic** with period
 - 2T, 3T, 4T, ...
 - 2N, 3N, 4N, ...
- **T** or **N** is called the **fundamental period**
denoted as **T₀** or **N₀**

▪ Periodic signal ?

$$x(t) = x(t + T) \quad \forall t, T > 0$$

$$x(t) = \begin{cases} \cos(t), & \text{if } t < 0 \\ \sin(t), & \text{if } t \geq 0 \end{cases}$$



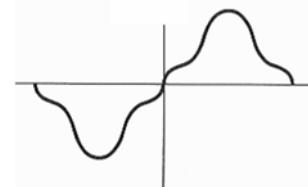
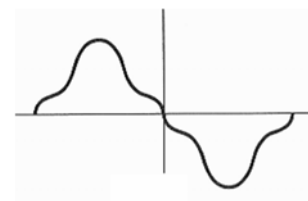
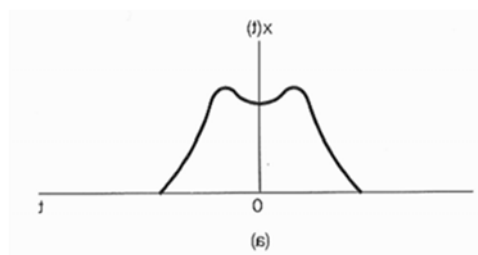
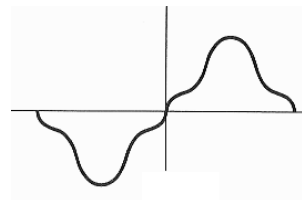
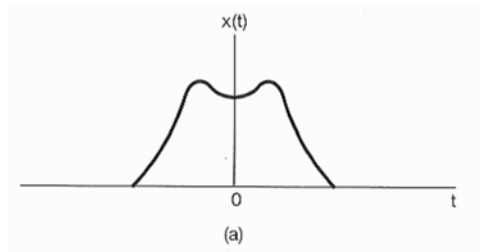
▪ Problems:

- P1.25 for CT
- P1.26 for DT

▪ Even & odd signals:

A signal is **even** if $x(-t) = x(t)$ or $x[-n] = x[n]$

A signal is **odd** if $x(-t) = -x(t)$ or $x[-n] = -x[n]$



▪ Even-odd decomposition of a signal:

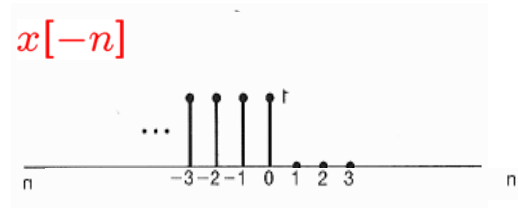
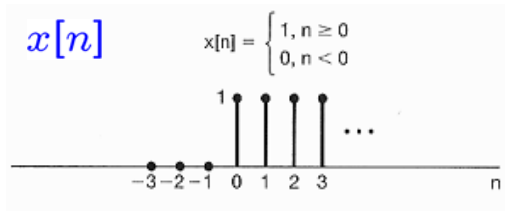
- Any signal can be broken into a **sum** of one **even** signal and one **odd** signal

$$\mathcal{E}v\{x(t)\} = \frac{1}{2} [x(t) + x(-t)] = \frac{1}{2} [x(-t) + x(t)]$$

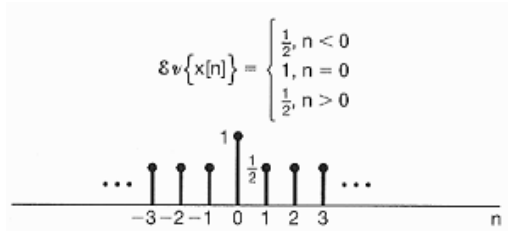
$$\mathcal{O}d\{x(t)\} = \frac{1}{2} [x(t) - x(-t)] = -\frac{1}{2} [x(-t) - x(t)]$$

$$\Rightarrow x(t) = \mathcal{E}v\{x(t)\} + \mathcal{O}d\{x(t)\}$$

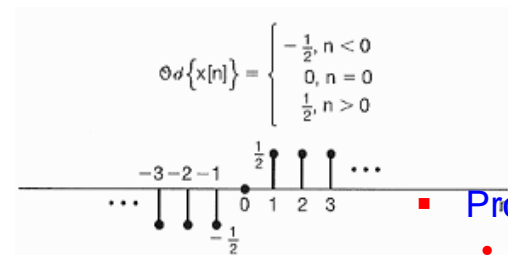
▪ Even-odd decomposition of a DT signal:



$$\mathcal{E}v \{x[n]\} = \frac{1}{2} [x[n] + x[-n]]$$



$$\mathcal{O}d \{x[n]\} = \frac{1}{2} [x[n] - x[-n]]$$



▪ **Problems:**

- P1.23 for CT
- P1.24 for DT

▪ Uniqueness of even-odd decomposition:

Assume that $x(t) = \mathcal{E}v_1(t) + \mathcal{O}d_1(t)$

and $x(t) = \mathcal{E}v_2(t) + \mathcal{O}d_2(t)$

So, $\mathcal{E}v_1(t) + \mathcal{O}d_1(t) = \mathcal{E}v_2(t) + \mathcal{O}d_2(t)$

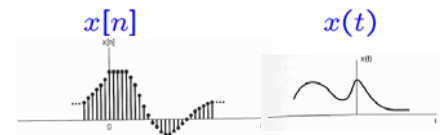
and $\mathcal{E}v_1(-t) + \mathcal{O}d_1(-t) = \mathcal{E}v_2(-t) + \mathcal{O}d_2(-t)$

Because $\begin{cases} \mathcal{E}v_1(-t) = \mathcal{E}v_1(t) \\ \mathcal{E}v_2(-t) = \mathcal{E}v_2(t) \end{cases}$ and $\begin{cases} \mathcal{O}d_1(-t) = -\mathcal{O}d_1(t) \\ \mathcal{O}d_2(-t) = -\mathcal{O}d_2(t) \end{cases}$

Then, $\mathcal{E}v_1(t) - \mathcal{O}d_1(t) = \mathcal{E}v_2(t) - \mathcal{O}d_2(t)$

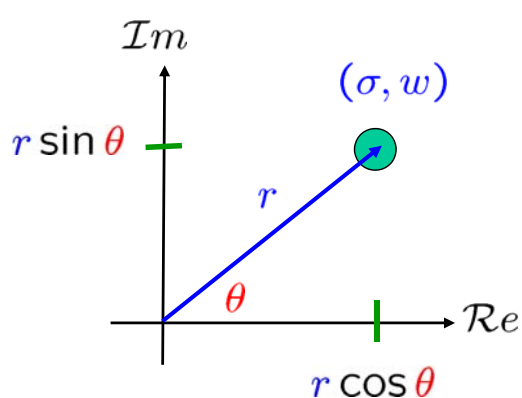
$\Rightarrow 2\mathcal{E}v_1(t) = 2\mathcal{E}v_2(t)$ or, $\mathcal{E}v_1(t) = \mathcal{E}v_2(t)$

$\Rightarrow 2\mathcal{O}d_1(t) = 2\mathcal{O}d_2(t)$ or, $\mathcal{O}d_1(t) = \mathcal{O}d_2(t)$



- Introduction
- Continuous-Time & Discrete-Time Signals
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 - Time Shift $x[n - n_0]$ $x(t - t_0)$ $x(-t) = x(t), x[-n] = x[n]$
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 - Periodic Signals $x(t) = x(t + T)$ $\mathcal{O}d\{x[n]\} = \frac{1}{2}[x[n] - x[-n]]$
 - Even & Odd Signals $x[n] = x[n + N]$
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▪ Magnitude & Phase Representation:



$$j = \sqrt{-1}$$

$$\sigma + jw \Rightarrow \begin{cases} r = \sqrt{\sigma^2 + w^2} \\ \tan(\theta) = \frac{w}{\sigma} \end{cases}$$

$$\Rightarrow \sigma + jw = r e^{j\theta}$$

▪ Euler's relation:

$$e^{j\theta} = \cos \theta + j \sin \theta$$

$$\Rightarrow \sigma + jw = r (\cos \theta + j \sin \theta)$$

$$= (r \cos \theta) + j(r \sin \theta)$$

▪ CT Complex Exponential Signals:

$$x(t) = Ce^{at}$$

- where **C** & **a** are, in general, complex numbers

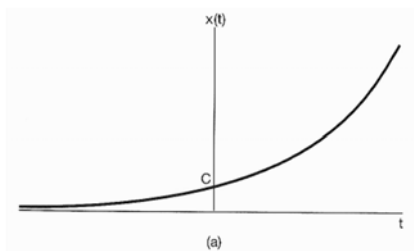
$$a = \sigma + j\omega$$

$$C = |C| e^{j\theta}$$

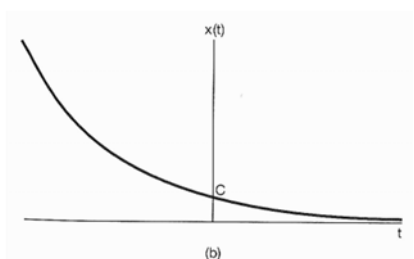
▪ Real exponential signals:

- If **C** & **a** are real

$$x(t) = Ce^{at}$$



$$a > 0$$



$$a < 0$$

▪ Periodic complex exponential signals: $e^{j\theta} = \cos \theta + j \sin \theta$

- If a is purely imaginary

$$a = \sigma + j\omega$$

$$x(t) = e^{j\omega_0 t}$$

- It is periodic

– Because let $T_0 = \frac{2\pi}{|\omega_0|}$

– Then $e^{j\omega_0 T_0} = e^{j\omega_0 \frac{2\pi}{\omega_0}} = \cos(\quad) + j \sin(\quad) =$

– Hence $e^{j\omega_0(t+T_0)} = e^{j\omega_0 t} e^{j\omega_0 T_0} = e^{j\omega_0 t}$

$$x(t + T_0) = x(t)$$

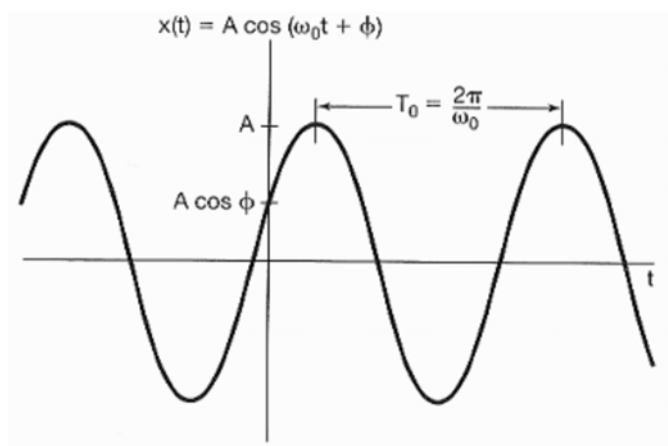
▪ Periodic sinusoidal signals:

$$\omega_0 = 2\pi f_0$$

$$x(t) = A \cos(\omega_0 t + \phi)$$

$$T_0 = \frac{2\pi}{\omega_0}$$

$$T_0 = \frac{1}{f_0}$$



$$T_0 : (\text{sec})$$

$$\omega_0 : (\text{rad/sec})$$

$$f_0 : (1/\text{sec} = \text{Hz})$$

▪ Period & Frequency:

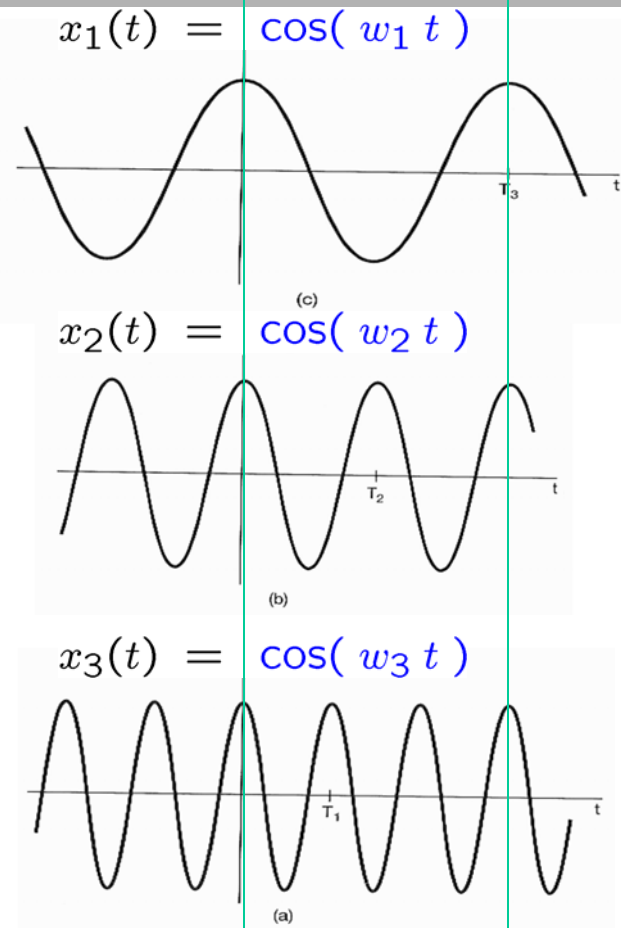
$$T_0 = \frac{2\pi}{\omega_0}$$

$$\omega_0 = 2\pi f_0$$

$$T_0 = \frac{1}{f_0}$$

$$\omega_1 \quad \omega_2 \quad \omega_3$$

$$T_1 \quad T_2 \quad T_3$$



▪ Euler's relation:

$$e^{j\theta} = \cos \theta + j \sin \theta$$

$$\cos(\theta) = \text{Re} \{ e^{j\theta} \}$$

$$\sin(\theta) = \text{Im} \{ e^{j\theta} \}$$

$$e^{j(-\theta)} = \cos(-\theta) + j \sin(-\theta)$$

$$\Rightarrow \cos(\theta) = \frac{e^{j\theta} + e^{-j\theta}}{2}$$

$$= \cos(\theta) - j \sin(\theta)$$

$$\Rightarrow \sin(\theta) = \frac{e^{j\theta} - e^{-j\theta}}{2j}$$

$$\Rightarrow A \cos(\omega_0 t + \phi) = \frac{A}{2} e^{j(\phi + \omega_0 t)} + \frac{A}{2} e^{-j(\phi + \omega_0 t)}$$

$$= \frac{A}{2} e^{j\phi} e^{j\omega_0 t} + \frac{A}{2} e^{-j\phi} e^{-j\omega_0 t}$$

▪ Total energy & average power:

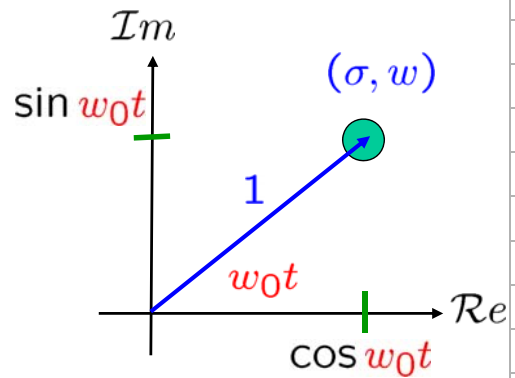
$$E_{\text{period}} = \int_0^{T_0} |e^{j\omega_0 t}|^2 dt$$

$$= \int_0^{T_0} 1 \cdot dt = T_0$$

$$P_{\text{period}} = \frac{1}{T_0} E_{\text{period}} = 1$$

$$E_{\infty} = \infty$$

$$P_{\infty} = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |e^{j\omega_0 t}|^2 dt = 1$$



▪ Problem:
• P1.3

▪ Harmonically related periodic exponentials

$$e^{j0\omega_0 t}, \quad e^{j1\omega_0 t}, \quad e^{j2\omega_0 t}, \quad e^{j3\omega_0 t}, \quad \dots,$$

$$e^{j(-1)\omega_0 t}, \quad e^{j(-2)\omega_0 t}, \quad e^{j(-3)\omega_0 t}, \quad \dots$$

$$\phi_k(t) = e^{j k \omega_0 t}, \quad k = 0, \pm 1, \pm 2, \dots$$

- For $k = 0$, $\phi_k(t)$ is constant
- For $k \neq 0$, $\phi_k(t)$ is periodic with

fundamental frequency $|k|\omega_0$ and
fundamental period $\frac{T_0}{|k|}$

▪ General complex exponential signals:

$$C e^{at} = (|C| e^{j\theta}) (e^{(r+j\omega_0)t})$$

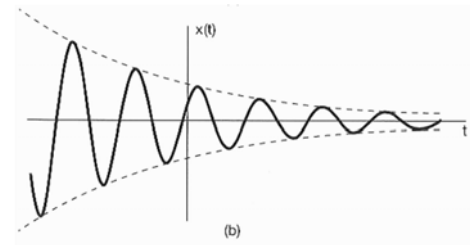
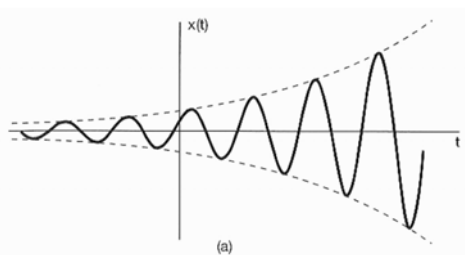
$$\sigma + j\omega = r e^{j\theta}$$

$$= (|C| e^{j\theta}) (e^{rt} e^{j\omega_0 t})$$

$$e^{j\theta} = \cos \theta + j \sin \theta$$

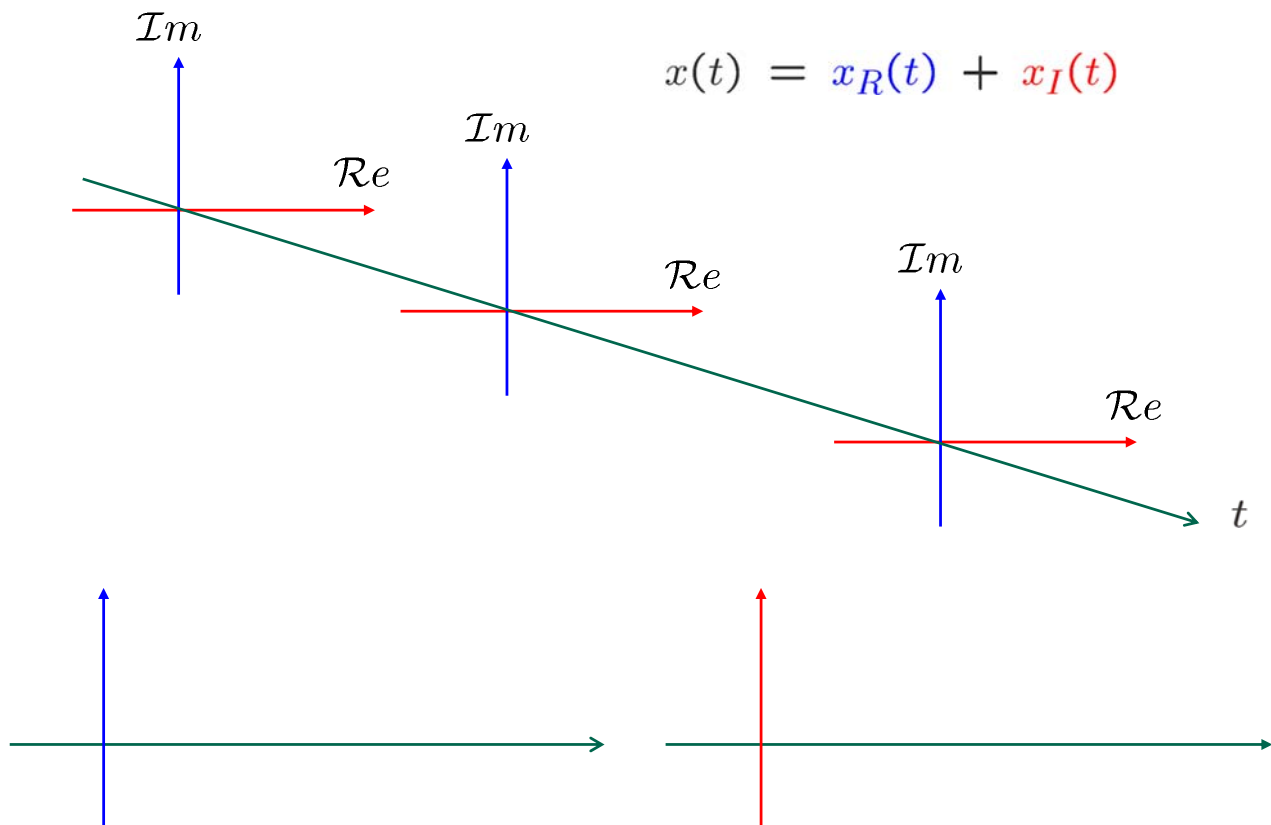
$$= |C| e^{rt} e^{j(\omega_0 t + \theta)}$$

$$= |C| e^{rt} \cos(\omega_0 t + \theta) + j |C| e^{rt} \sin(\omega_0 t + \theta)$$



▪ General complex exponential signals:

$$x(t) = x_R(t) + x_I(t)$$



▪ DT complex exponential signal or sequence:

$$x(t) = Ce^{at}$$

$$x[n] = Ce^{bn}$$

$$= C(e^b)^n \quad \text{with } a = e^b$$

$$x[n] = Ca^n$$

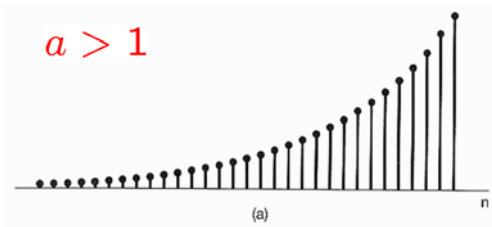
- where C & a are, in general, complex numbers

▪ Real exponential signals:

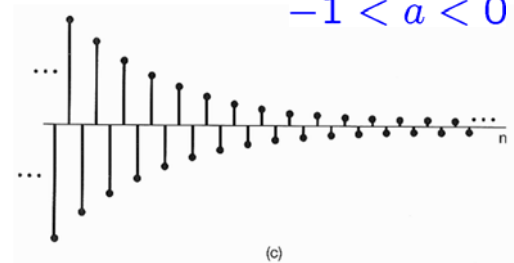
- If C & a are real

$$x[n] = Ca^n$$

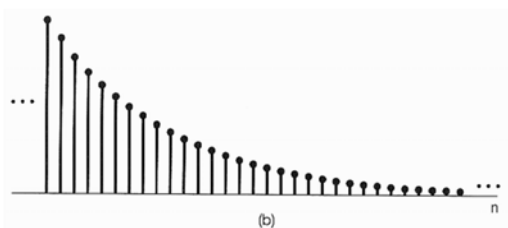
$$a > 1$$



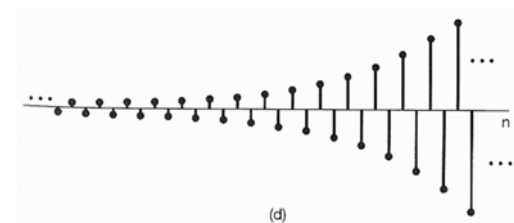
$$-1 < a < 0$$



$$0 < a < 1$$



$$a < -1$$



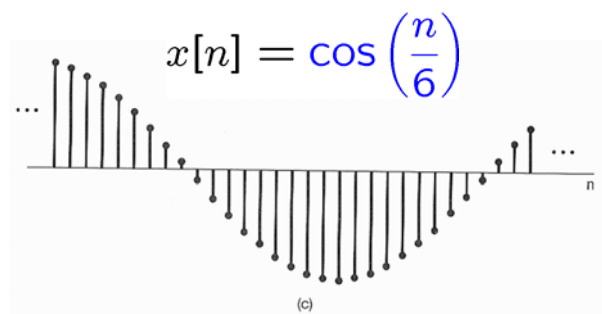
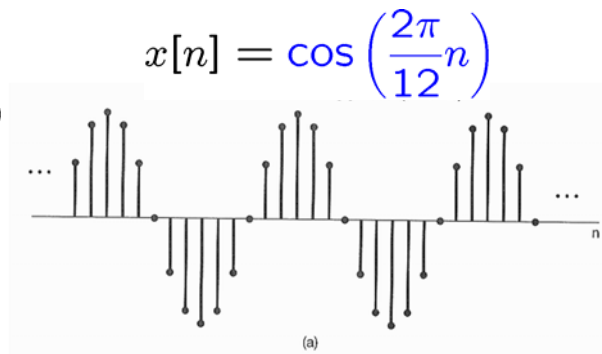
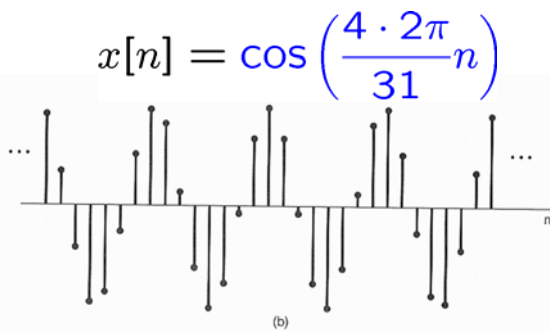
▪ DT Complex Exponential & Sinusoidal Signals

- If b is purely imaginary (or $|a| = 1$) $e^{j\theta} = \cos \theta + j \sin \theta$

$$x[n] = e^{jw_0n}$$

$$= \cos(w_0n) + j \sin(w_0n)$$

$$x[n] = A \cos(w_0n + \phi)$$



▪ Euler's relation:

$$e^{jw_0n} = \cos w_0n + j \sin w_0n$$

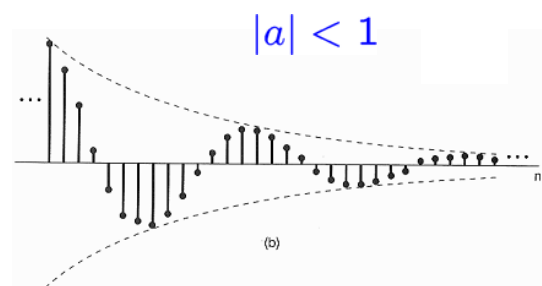
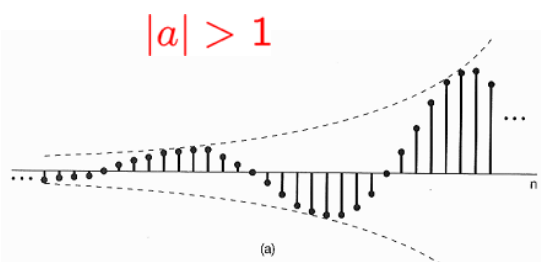
- And,

$$A \cos(w_0n + \phi) = \frac{A}{2} e^{j\phi} e^{jw_0n} + \frac{A}{2} e^{-j\phi} e^{-jw_0n}$$

General complex exponential signals:

$$Ca^n = (|C|e^{j\theta})(|a|e^{j\omega_0})^n$$

$$= |C||a|^n \cos(\omega_0 n + \theta) + j|C||a|^n \sin(\omega_0 n + \theta)$$



Periodicity properties of DT complex exponentials:

$$e^{j\omega_0 n} \quad e^{j2\pi n} = \cos 2\pi n + j \sin 2\pi n$$

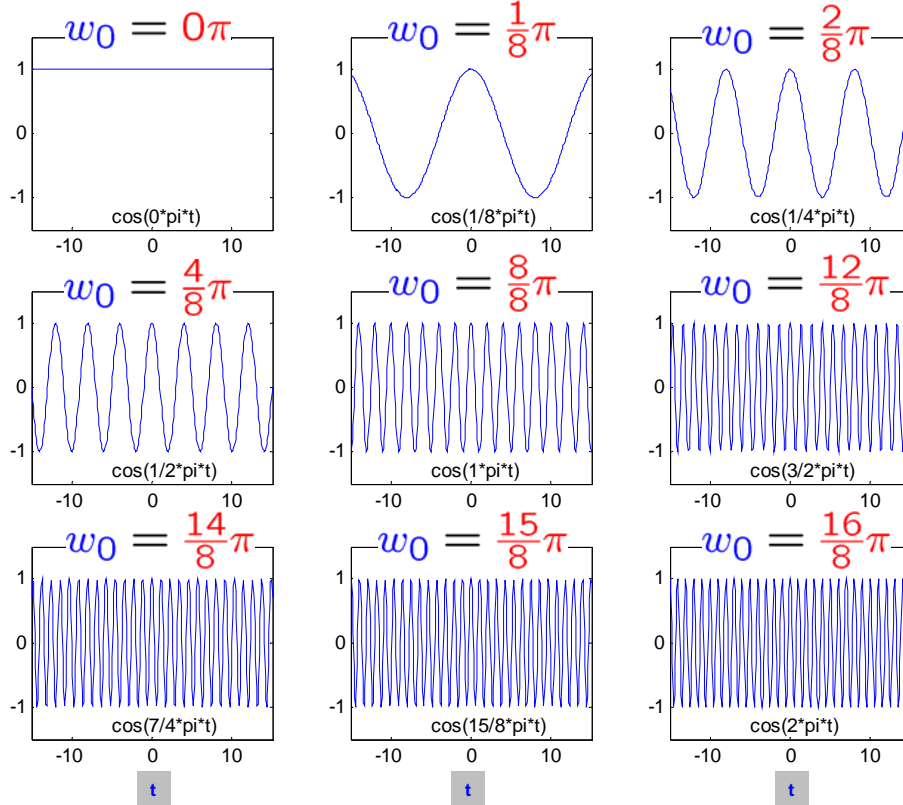
$$e^{j(\omega_0 + 2\pi)n} = e^{j2\pi n} e^{j\omega_0 n} = e^{j\omega_0 n}$$

- The signal with frequency ω_0 is identical to the signals with frequencies $\omega_0 \pm 2\pi, \omega_0 \pm 4\pi, \omega_0 \pm 6\pi, \dots$
- Only need to consider a frequency interval of length 2π
 - Usually use $0 \leq \omega_0 < 2\pi$ or $-\pi \leq \omega_0 < \pi$,
- The **low** frequencies are located at $\omega_0 = 0, \pm 2\pi, \dots$
 The **high** frequencies are located at $\omega_0 = \pm\pi, \pm 3\pi, \dots$

$$e^{j(0)n} = 1 \quad \text{and} \quad e^{j(\pi)n} = (e^{j(\pi)})^n = (-1)^n$$

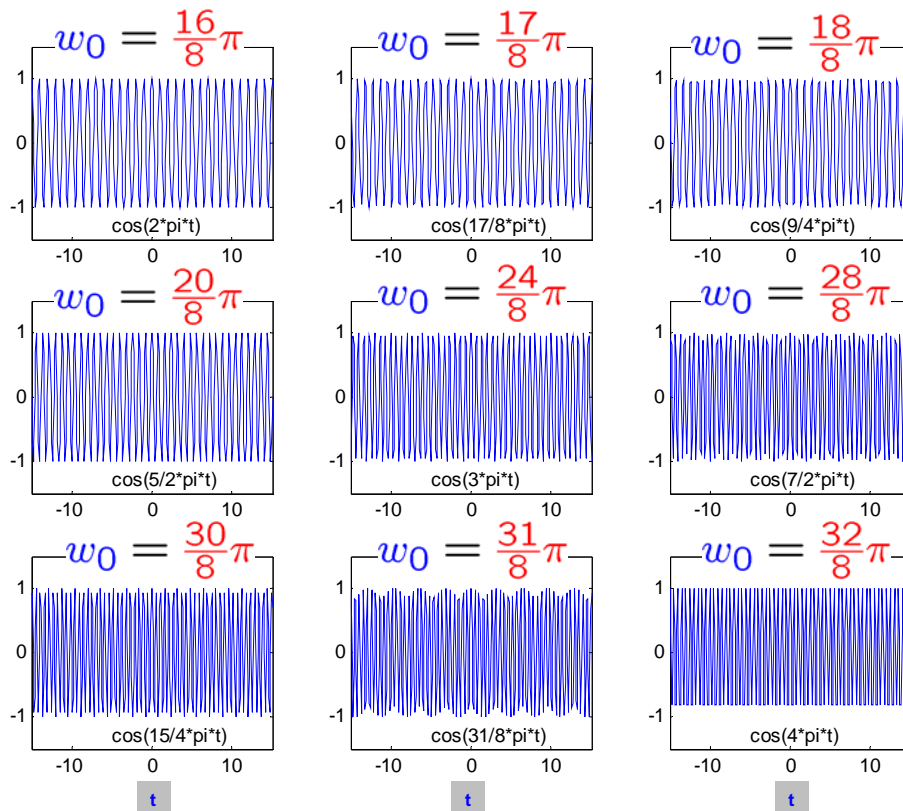
CT exponential signals

$\cos(w_0 t)$

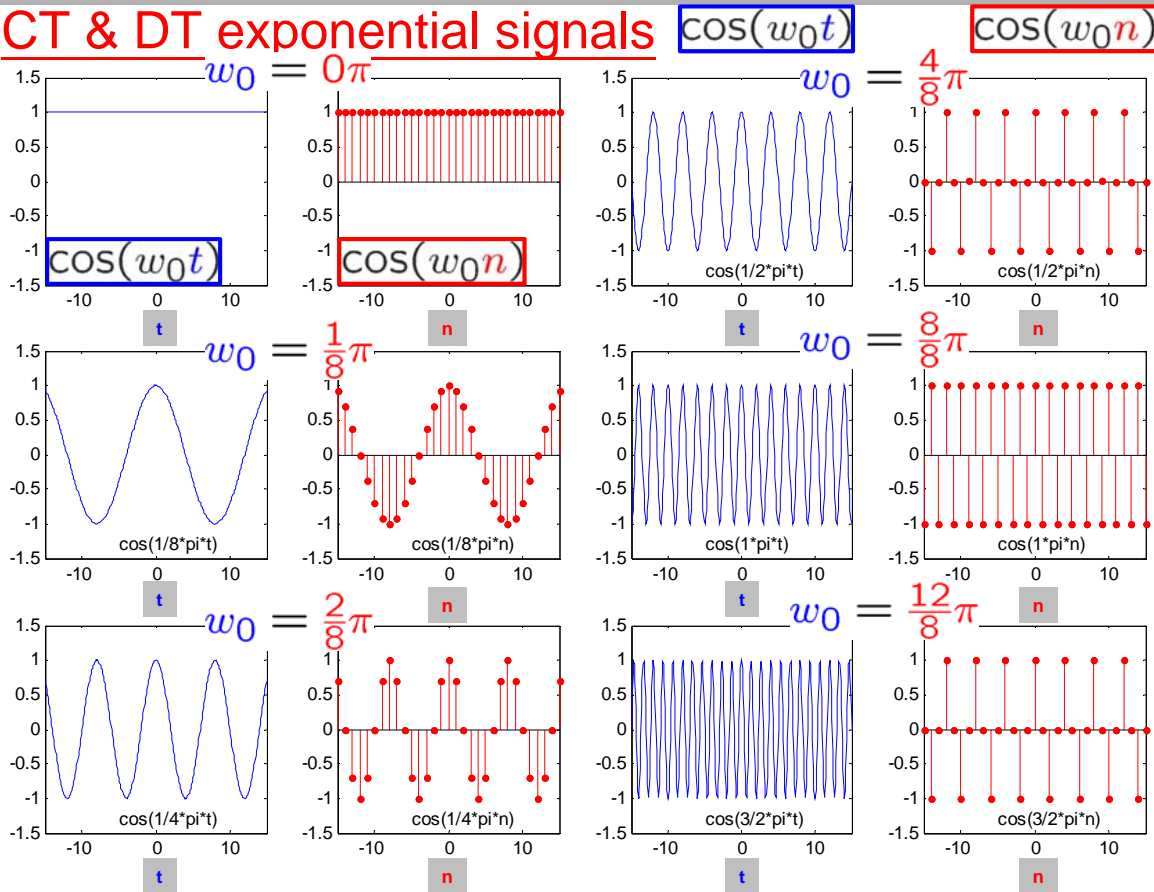


CT exponential signals

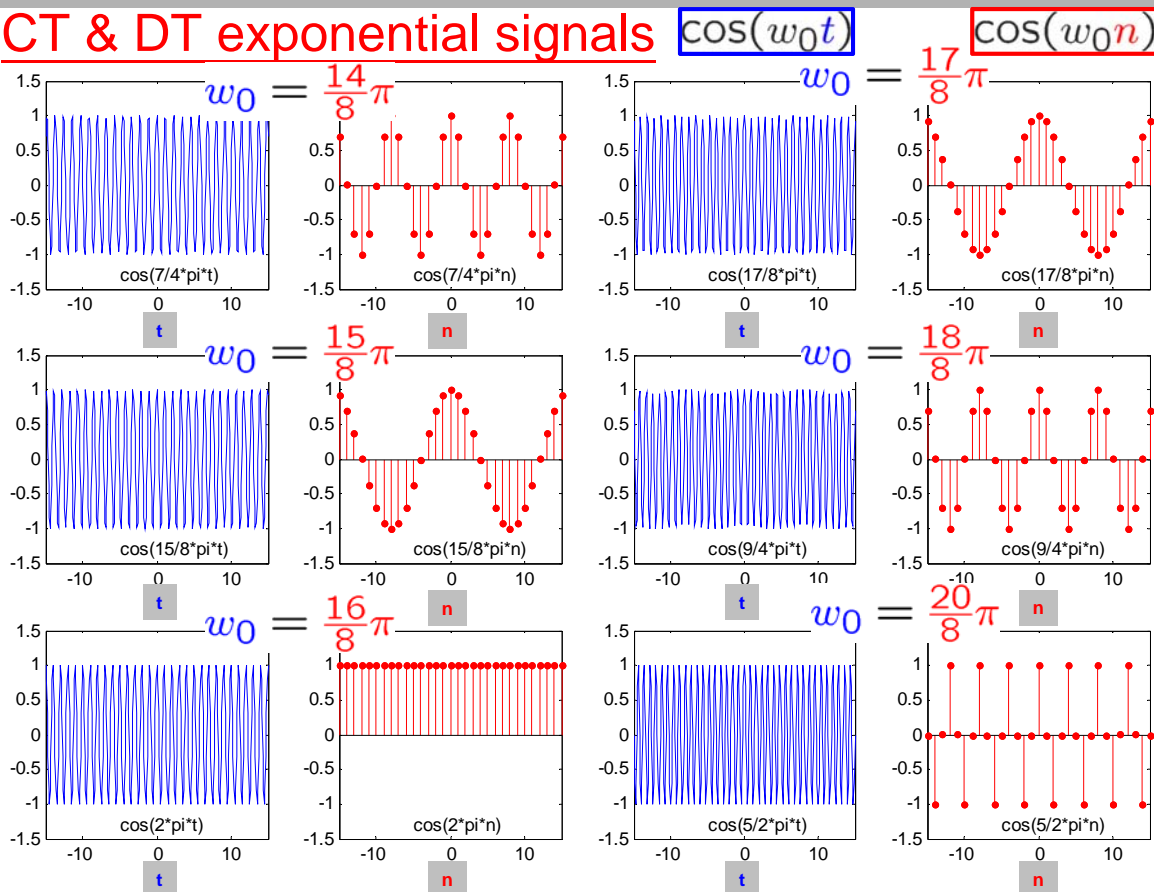
$\cos(w_0 t)$



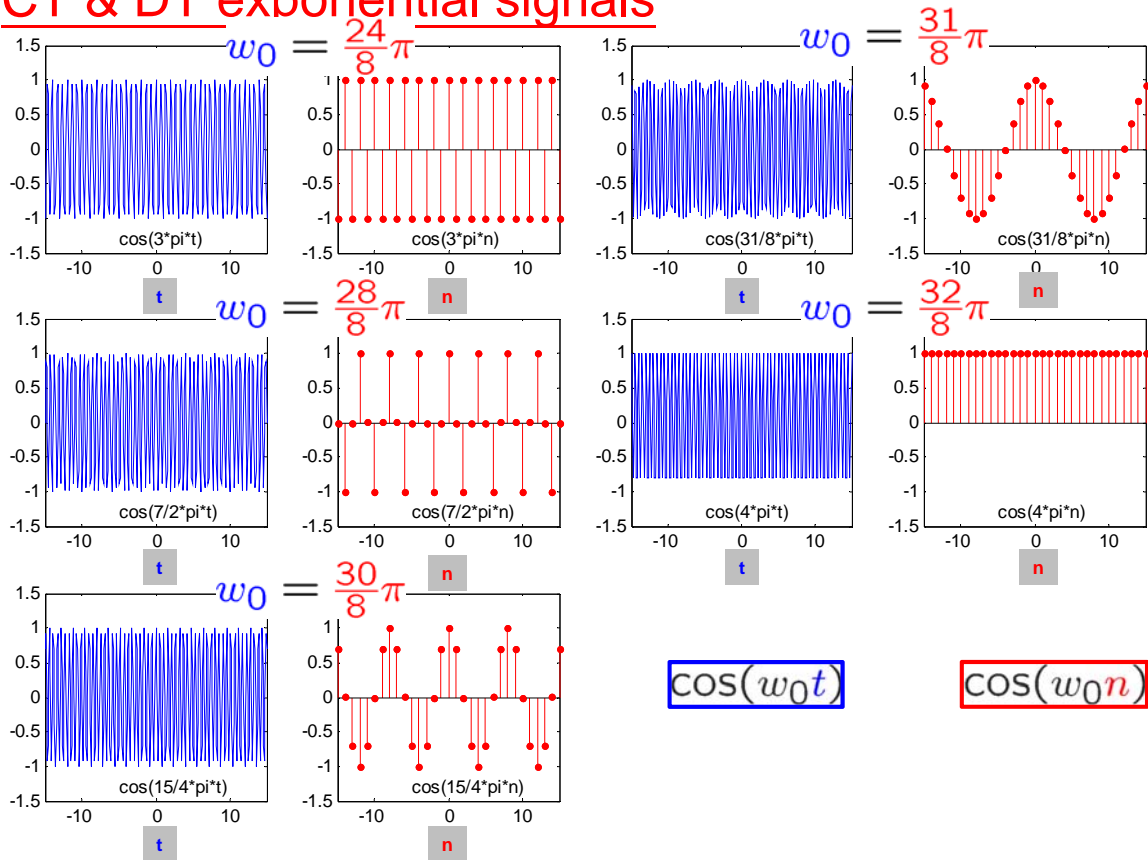
CT & DT exponential signals



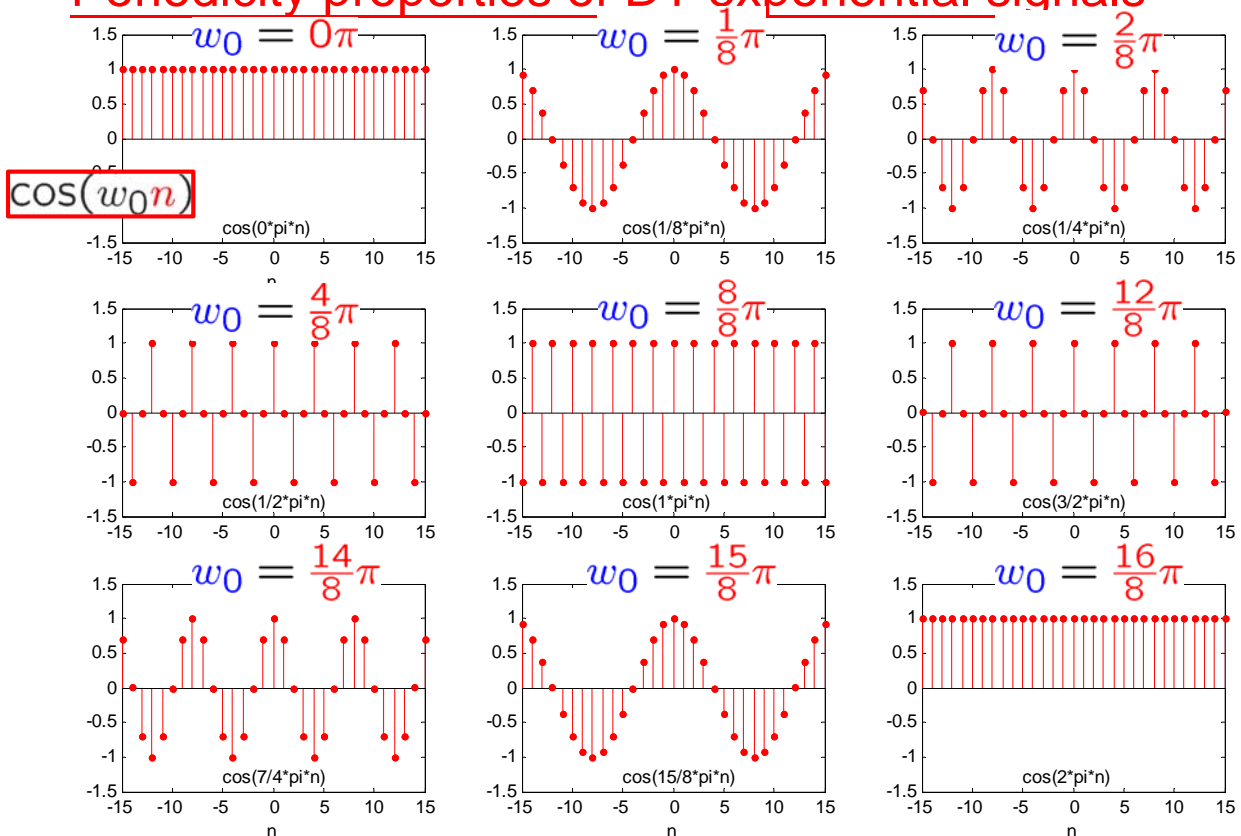
CT & DT exponential signals



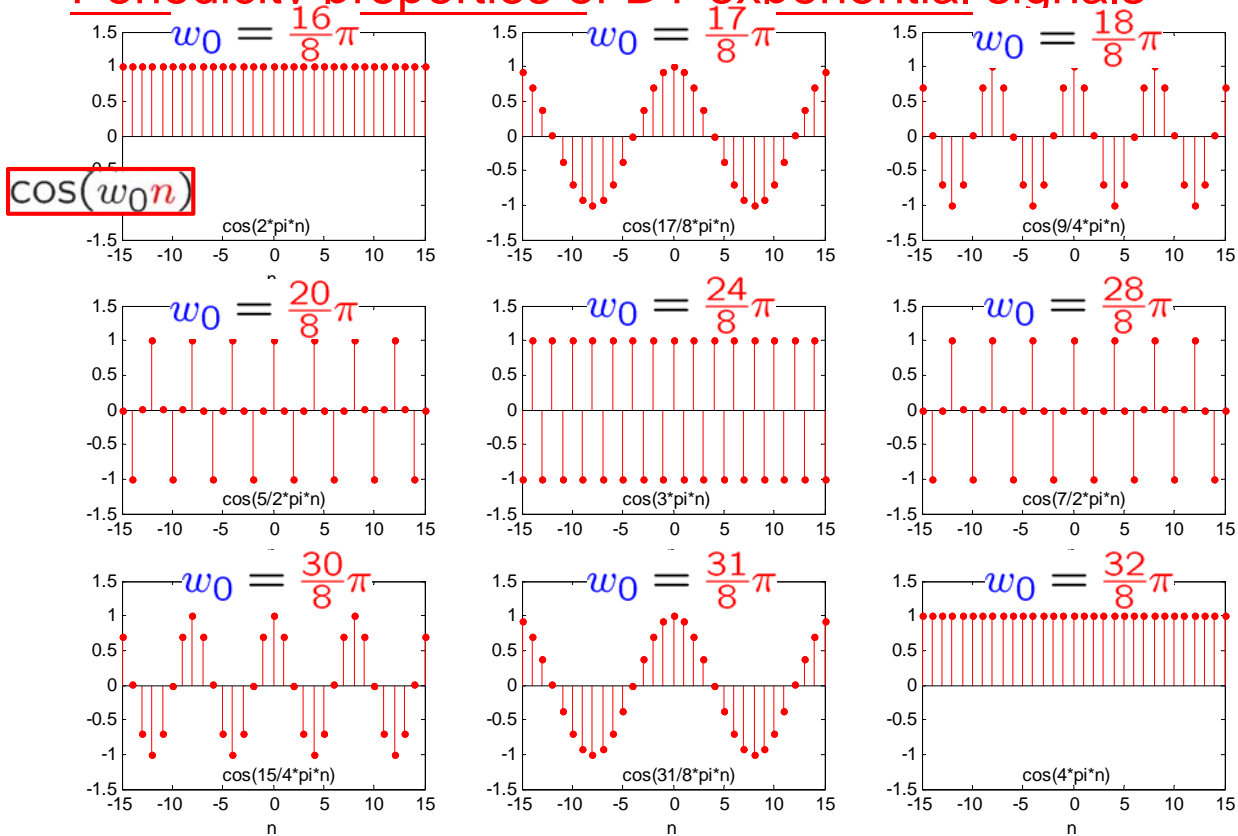
CT & DT exponential signals



Periodicity properties of DT exponential signals



Periodicity properties of DT exponential signals



Periodicity properties of DT exponential signals

Periodicity of $N > 0$

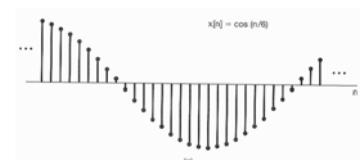
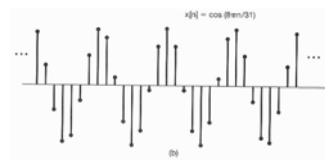
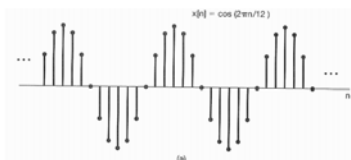
Problem:
• P1.35

$$e^{j\omega_0 n} \quad e^{j\omega_0(n+N)} = e^{j\omega_0 n} e^{j\omega_0 N} = e^{j\omega_0 n} \quad \text{or} \quad e^{j\omega_0 N} = 1$$

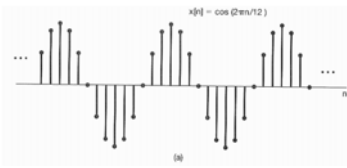
That is, $\omega_0 N = 2\pi m$ or $\frac{\omega_0}{2\pi} = \frac{m}{N}$

Hence, $e^{j\omega_0 n}$ is periodic if $\frac{\omega_0}{2\pi}$ is a rational number

$$x[n] = \cos\left(\frac{2\pi}{12}n\right) \quad x[n] = \cos\left(\frac{4 \cdot 2\pi}{31}n\right) \quad x[n] = \cos\left(\frac{1}{6}n\right)$$

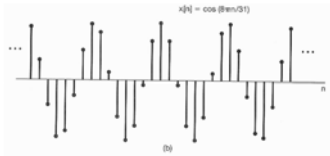


▪ Periodicity properties of DT exponential signals



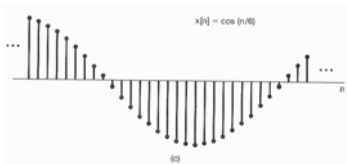
$$x(t) = \cos\left(\frac{2\pi}{12}t\right) \quad T = 12?$$

$$x[n] = \cos\left(\frac{2\pi}{12}n\right) \quad N = 12?$$



$$x(t) = \cos\left(\frac{4 \cdot 2\pi}{31}t\right) \quad T = \frac{31}{4}?$$

$$x[n] = \cos\left(\frac{4 \cdot 2\pi}{31}n\right) \quad N = \frac{31}{4}?$$



$$x(t) = \cos\left(\frac{1}{6}t\right) \quad T = 12\pi?$$

$$x[n] = \cos\left(\frac{1}{6}n\right) \quad N = 12\pi?$$

▪ Harmonically related periodic exponentials

$$\phi_k[n] = e^{jk(w_0)n}, \quad = e^{jk\left(\frac{2\pi}{N}\right)n}, \quad k = 0, \pm 1, \pm 2, \dots$$

$$\phi_{k+N}[n] = e^{j(k+N)\left(\frac{2\pi}{N}\right)n}$$

$$= e^{jk\left(\frac{2\pi}{N}\right)n} e^{jN\left(\frac{2\pi}{N}\right)n} = \phi_k[n]$$

- Only **N distinct** periodic exponentials in the set

$$\phi_0[n] = 1, \quad \phi_1[n] = e^{j\left(\frac{2\pi}{N}\right)n}, \quad \phi_2[n] = e^{j\left(2\frac{2\pi}{N}\right)n},$$

$$\dots, \quad \phi_{N-1}[n] = e^{j(N-1)\frac{2\pi}{N}n}$$

$$\phi_N[n] = e^{j(N)\frac{2\pi}{N}n} = e^{j2\pi n} = 1 = \phi_0[n], \quad ; \quad \phi_{N+1}[n] = \phi_1[n], \dots$$

■ Comparison of CT & DT signals:

TABLE 1.1 Comparison of the signals $e^{j\omega_0 t}$ and $e^{j\omega_0 n}$.

| CT | $e^{j\omega_0 t}$ | DT | $e^{j\omega_0 n}$ |
|---|-------------------|--|-------------------|
| Distinct signals for distinct values of ω_0 | | Identical signals for values of ω_0 separated by multiples of 2π | |
| Periodic for any choice of ω_0 | | Periodic only if $\omega_0 = 2\pi m/N$ for some integers $N > 0$ and m . | |
| Fundamental frequency ω_0 | | Fundamental frequency* ω_0/m | |
| Fundamental period $\omega_0 = 0$: undefined $\omega_0 \neq 0$: $\frac{2\pi}{\omega_0}$ | | Fundamental period* $\omega_0 = 0$: undefined $\omega_0 \neq 0$: $m\left(\frac{2\pi}{\omega_0}\right)$ | |

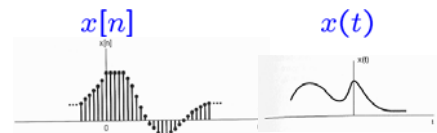
*Assumes that m and N do not have any factors in common.

DT 1, $e^{j\left(\frac{2\pi}{N}n\right)}$, $e^{j\left(2\frac{2\pi}{N}n\right)}$, ..., $e^{j(N-1)\frac{2\pi}{N}n}$

CT 1, $e^{j1\omega_0 t}$, $e^{j2\omega_0 t}$, $e^{j3\omega_0 t}$, ...,
 $e^{j(-1)\omega_0 t}$, $e^{j(-2)\omega_0 t}$, $e^{j(-3)\omega_0 t}$, ...

■ Problem:
• P1.36

Outline

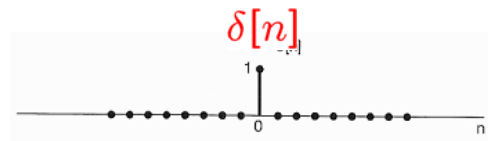


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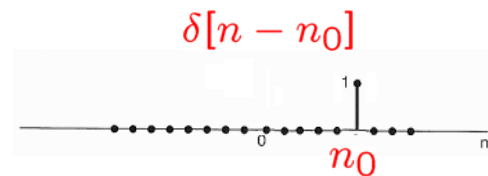
DT Unit Impulse & Unit Step Sequences

Unit impulse (or unit sample)

$$\delta[n] = \begin{cases} 1, & n = 0 \\ 0, & n \neq 0 \end{cases}$$



$$\delta[n - n_0] = \begin{cases} 1, & n = n_0 \\ 0, & n \neq n_0 \end{cases}$$

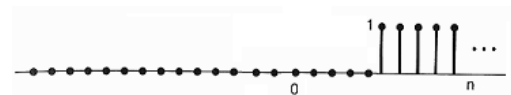


Unit step

$$u[n] = \begin{cases} 0, & n < 0 \\ 1, & n \geq 0 \end{cases}$$



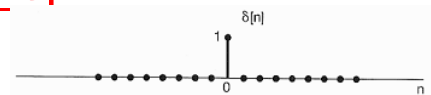
$$u[n - n_0] = \begin{cases} 0, & n < n_0 \\ 1, & n \geq n_0 \end{cases}$$



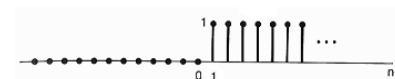
Relationship Between Impulse & Step

First difference

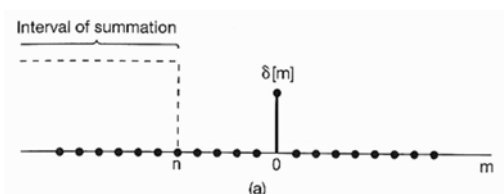
$$\delta[n] = u[n] - u[n - 1]$$



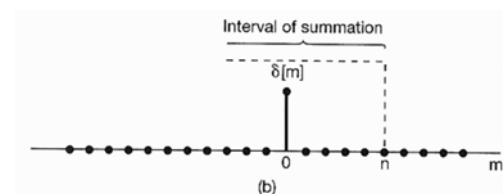
Running sum



$$u[n] = \sum_{m=-\infty}^n \delta[m] = \begin{cases} 0, & n < 0 \\ 1, & n \geq 0 \end{cases}$$



$n < 0$



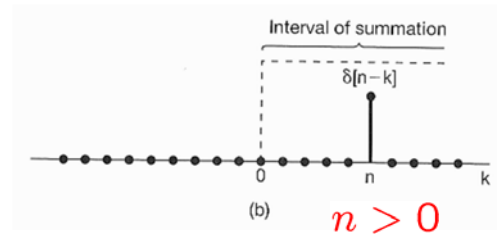
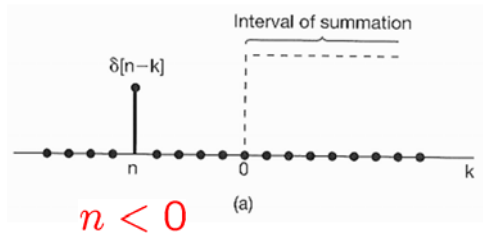
$n > 0$

▪ Relationship Between Impulse & Step

▪ Alternatively,

$$u[n] = \sum_{k=-\infty}^0 \delta[n - k], \quad \text{with } m = n - k$$

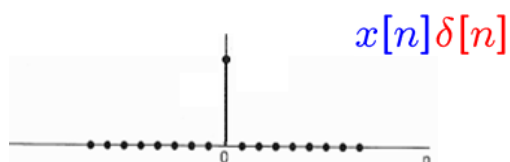
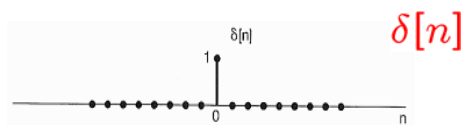
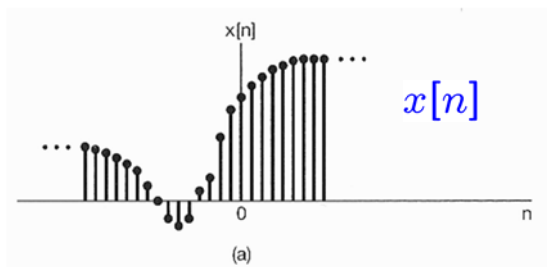
or,
$$u[n] = \sum_{k=0}^{\infty} \delta[n - k]$$



▪ Sample by Unit Impulse

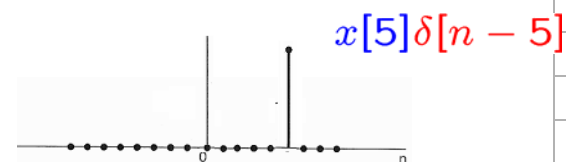
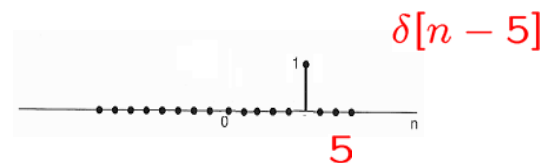
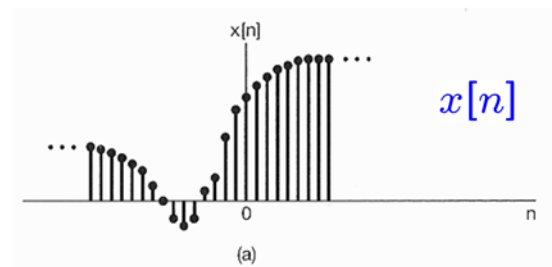
▪ For $x[n]$

$$x[n]\delta[n] = x[0]\delta[n]$$



▪ More generally,

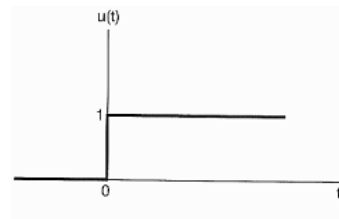
$$x[n]\delta[n - n_0] = x[n_0]\delta[n - n_0]$$



CT Unit Impulse & Unit Step Functions

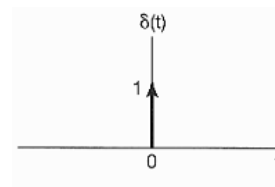
Unit step function

$$u(t) = \begin{cases} 0, & t < 0 \\ 1, & t > 0 \end{cases}$$



Unit impulse function

$$\delta(t)$$



Relationship Between Impulse & Step

Running integral

$$u(t) = \int_{-\infty}^t \delta(\tau) d\tau = \begin{cases} 0, & t < 0 \\ 1, & t > 0 \end{cases}$$

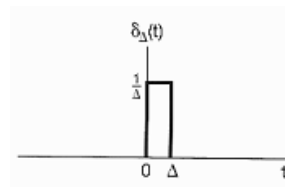
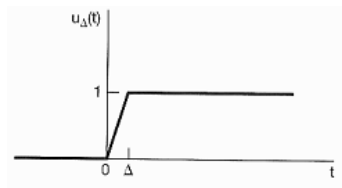
First derivative

$$\delta(t) = \frac{du(t)}{dt}$$

- But, $u(t)$ is **discontinuous** at $t = 0$, hence, **not differentiable**
- Use approximation

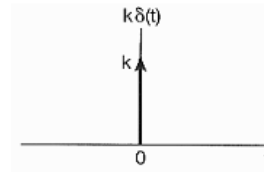
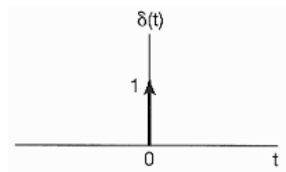
▪ Relationship Between Impulse & Step

- Use approximation



$$\delta_{\Delta}(t) = \frac{du_{\Delta}(t)}{dt}$$

$$\delta(t) = \lim_{\Delta \rightarrow 0} \delta_{\Delta}(t)$$



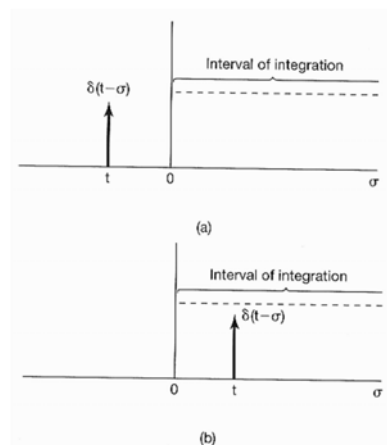
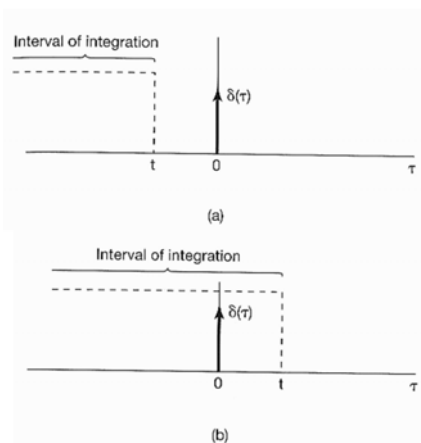
▪ Relationship Between Impulse & Step

$$u(t) = \int_{-\infty}^t \delta(\tau) d\tau = \int_{\infty}^0 \delta(t - \sigma) (-d\sigma) = \int_0^{\infty} \delta(t - \sigma) (d\sigma)$$

$$\tau = t - \sigma$$

$$d\tau = -d\sigma$$

$$= \int_0^{\infty} \delta(t - \tau) (d\tau)$$



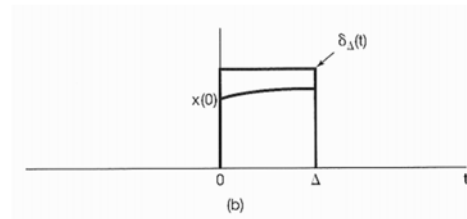
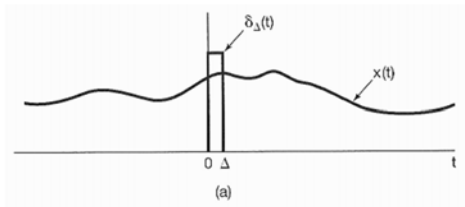
▪ Sample by Unit Impulse Function

- For $x(t)$

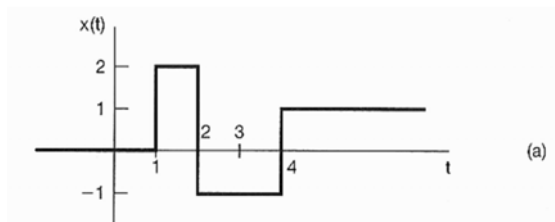
$$x(t)\delta(t) = x(0)\delta(t)$$

- More generally,

$$x(t)\delta(t - t_0) = x(t_0)\delta(t - t_0)$$



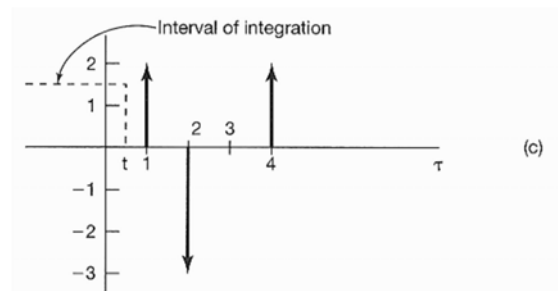
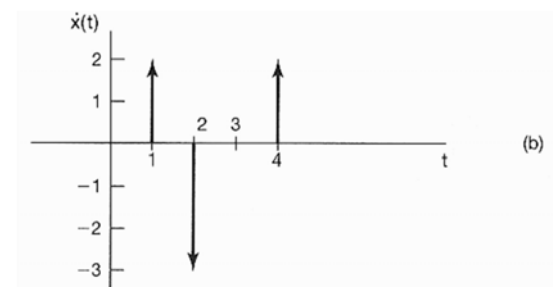
▪ Example 1.7:



$$\delta(t) = \frac{du(t)}{dt}$$

$$u(t) = \int_{-\infty}^t \delta(\tau) d\tau$$

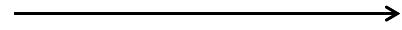
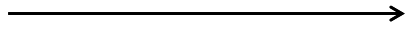
$$x(t) = \int_0^t \dot{x}(\tau) d\tau$$



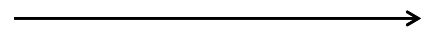
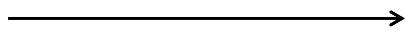
▪ Exponential, Sinusoidal, Impulse, Step functions

time

frequency



$$e^{j\omega_0 t} = \cos \omega_0 t + j \sin \omega_0 t$$



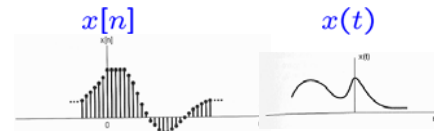
▪ Summation or Integration of Three Functions

$$f(\quad) = \int g(\quad) h(\quad) d$$

$$= \int g(\quad) h(\quad) d$$

$$f[\quad] = \sum g[\quad] h[\quad]$$

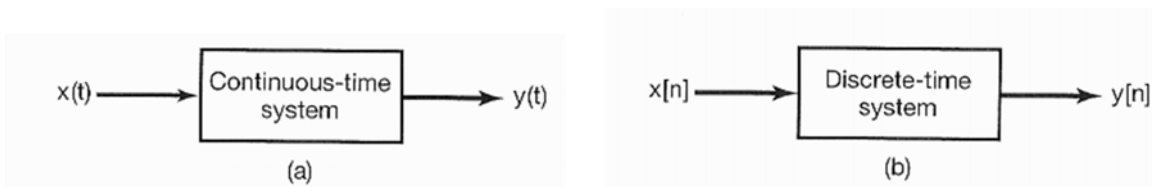
$$= \sum g[\quad] h[\quad]$$



- Introduction
- Continuous-Time & Discrete-Time Signals
- Transformations of the Independent Variable
 - Time Shift $x[n - n_0]$ $x(t - t_0)$ $x(-t) = x(t), x[-n] = x[n]$
 - Time Reversal $x[-n]$ $x(-t)$ $x(-t) = -x(t), x[-n] = -x[n]$
 - Time Scaling $x[an]$ $x(at)$ $\mathcal{E}v\{x[n]\} = \frac{1}{2}[x[n] + x[-n]]$
 - Periodic Signals $x(t) = x(t + T)$ $\mathcal{O}d\{x[n]\} = \frac{1}{2}[x[n] - x[-n]]$
 - Even & Odd Signals $x[n] = x[n + N]$ $\phi_k(t) = e^{jk\omega_0 t}, k = 0, \pm 1, \dots$
- Exponential & Sinusoidal Signals $\phi_k[n] = e^{jk\omega_0 n}, k = 0, \dots, N - 1$
- The Unit Impulse & Unit Step Functions $\delta[n], u[n]$
 $\delta(t), u(t)$
- Continuous-Time & Discrete-Time Systems
- Basic System Properties

▪ Physical Systems & Mathematical Descriptions

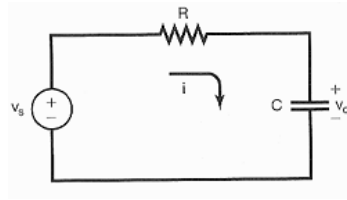
- Examples of **physical systems** are signal processing, communications, electromechanical motors, automotive vehicles, chemical-processing plants
- A **system** can be viewed as a **process** in which **input signals** are **transformed** by the system or **cause** the system to **respond** in some way, resulting in **other signals** or **outputs**



▪ Simple examples of CT systems

▪ RC circuit

Input signal: $v_s(t)$

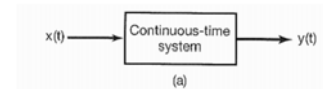


Output signal: $v_c(t)$

$$i(t) = \frac{v_s(t) - v_c(t)}{R} \quad i(t) = C \frac{dv_c(t)}{dt}$$

$$\Rightarrow \frac{dv_c(t)}{dt} + \frac{1}{RC}v_c(t) = \frac{1}{RC}v_s(t)$$

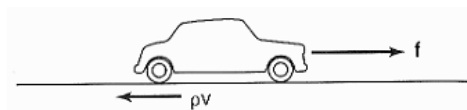
$$\Rightarrow \frac{dy(t)}{dt} + ay(t) = bx(t)$$



$x(t) \rightarrow y(t)$

▪ Simple examples of CT systems

▪ Automobile



Input signal: $f(t)$

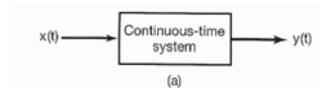
Output signal: $v(t)$

$$f(t) - \rho v(t) = m \frac{dv(t)}{dt}$$

$$\frac{dv(t)}{dt} = \frac{1}{m}[f(t) - \rho v(t)]$$

$$\Rightarrow \frac{dv(t)}{dt} + \frac{\rho}{m}v(t) = \frac{1}{m}f(t)$$

$$\Rightarrow \frac{dy(t)}{dt} + ay(t) = bx(t)$$



$x(t) \rightarrow y(t)$

Simple examples of DT systems

- Balance in a bank account

$$y[n] = 1.01y[n-1] + x[n]$$

$$\text{or, } y[n] - 1.01y[n-1] = x[n]$$

$$\Rightarrow y[n] + ay[n-1] = bx[n]$$



$$x[n] \rightarrow y[n]$$

Simple examples of DT systems

- Digital simulation of differential equation

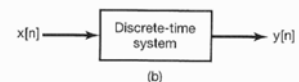
$$\frac{dv(t)}{dt} \approx \frac{v(n\Delta) - v((n-1)\Delta)}{\Delta} = \frac{v[n] - v[n-1]}{\Delta},$$

$$t = n\Delta$$

$$\frac{dv(t)}{dt} + \frac{\rho}{m}v(t) = \frac{1}{m}f(t)$$

$$\Rightarrow v[n] - \frac{m}{m + \rho\Delta}v[n-1] = \frac{\Delta}{m + \rho\Delta}f[n]$$

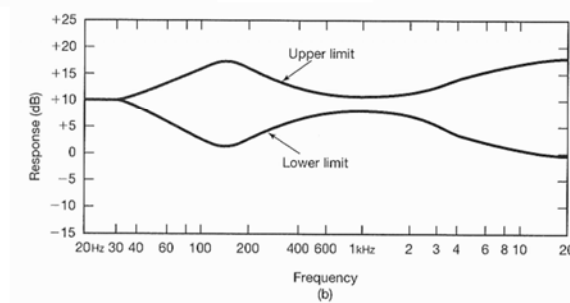
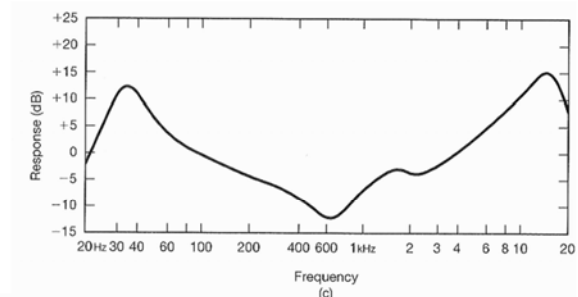
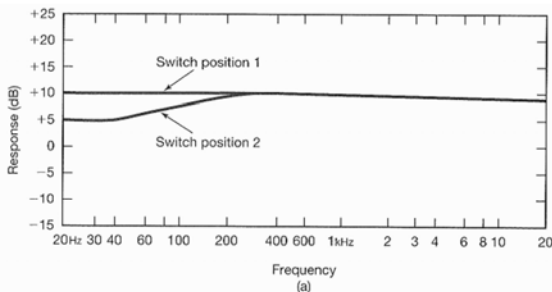
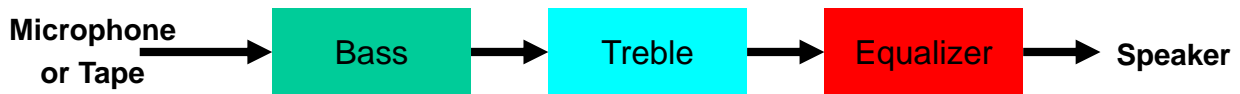
$$\Rightarrow y[n] + ay[n-1] = bx[n]$$



$$x[n] \rightarrow y[n]$$

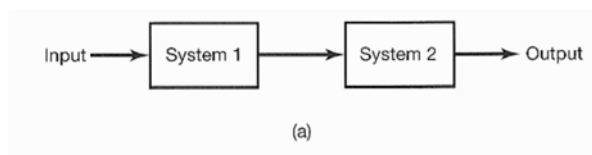
▪ Interconnections of Systems:

- Audio System:



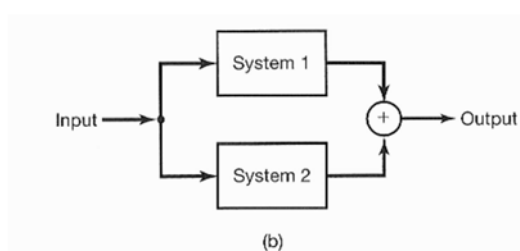
▪ Interconnections of Systems

- Series or cascade interconnection of 2 systems



> e.x. radio receiver + amplifier

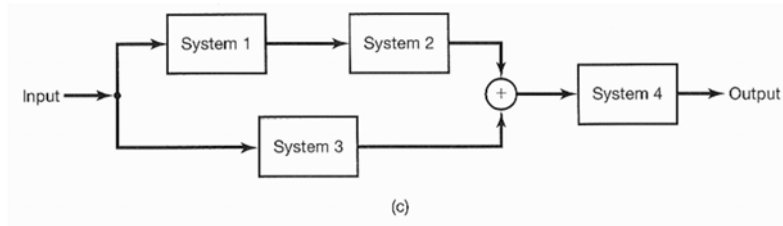
- Parallel interconnection of 2 systems



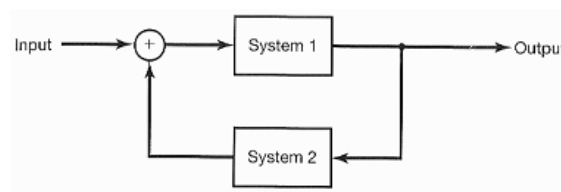
> e.x. audio system with several microphones or speakers

▪ **Interconnections of Systems**

▪ **Series-parallel interconnection**



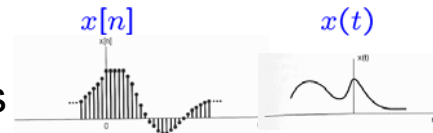
▪ **Feedback interconnection**



> e.x. cruise control, electrical circuit

Chapter 1: Signals and Systems

- Introduction
- Continuous-Time & Discrete-Time Signals
- Transformations of the Independent Variable



- Time Shift $x[n - n_0]$ $x(t - t_0)$ $x(-t) = x(t), x[-n] = x[n]$
- Time Reversal $x[-n]$ $x(-t)$ $x(-t) = -x(t), x[-n] = -x[n]$
- Time Scaling $x[an]$ $x(at)$
- Periodic Signals $x(t) = x(t + T)$
- Even & Odd Signals $x[n] = x[n + N]$

$$Ev\{x[n]\} = \frac{1}{2} [x[n] + x[-n]]$$

$$Od\{x[n]\} = \frac{1}{2} [x[n] - x[-n]]$$

$$\phi_k(t) = e^{jk\omega_0 t}, k = 0, \pm 1, \dots$$

$$\phi_k[n] = e^{jk\omega_0 n}, k = 0, \dots, N - 1$$

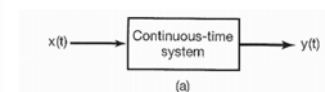
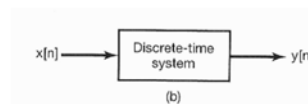
- Exponential & Sinusoidal Signals
- The Unit Impulse & Unit Step Functions
- Continuous-Time & Discrete-Time Systems

$$\delta[n], u[n]$$

$$\delta(t), u(t)$$

▪ **Basic System Properties**

- Systems with or without memory
- Invertibility & Inverse Systems
- Causality
- Stability
- Time Invariance
- Linearity



$$x[n] \rightarrow y[n]$$

$$x(t) \rightarrow y(t)$$

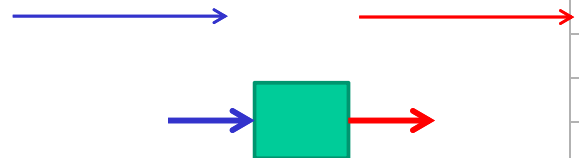
▪ Systems with or without memory

▪ Memoryless systems

- Output depends only on the input at that same time

$$y[n] = (2x[n] - x[n]^2)^2$$

$$y(t) = Rx(t) \quad (\text{resistor})$$



▪ Systems with memory

$$y[n] = \sum_{k=-\infty}^n x[k] \quad (\text{accumulator})$$

$$y[n] = x[n - 1] \quad (\text{delay})$$

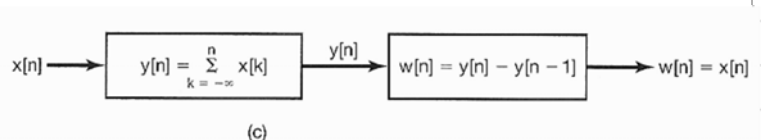
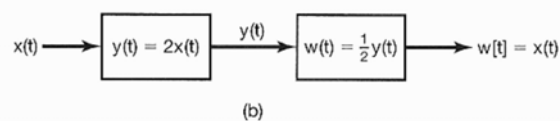
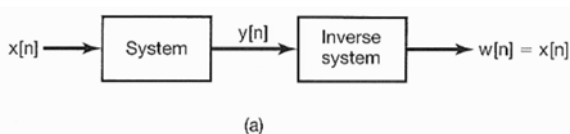
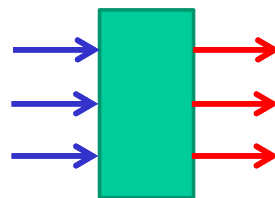
$$y(t) = \frac{1}{C} \int_{-\infty}^t x(\tau) d\tau$$



▪ Invertibility & Inverse Systems

▪ Invertible systems

- Distinct inputs lead to distinct outputs

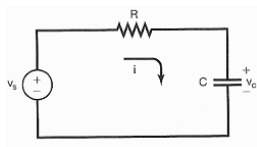


$$y(t) = x(t)^2 \text{ is not invertible}$$

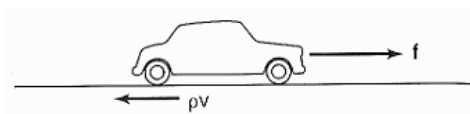
Causality

Causal systems

- Output depends only on input at present time & in the past



$$\Rightarrow \frac{dv_c(t)}{dt} + \frac{1}{RC}v_c(t) = \frac{1}{RC}v_s(t)$$



$$\Rightarrow \frac{dv(t)}{dt} + \frac{\rho}{m}v(t) = \frac{1}{m}f(t)$$

- Non-causal systems

$$y[n] = x[n] - x[n + 1]$$

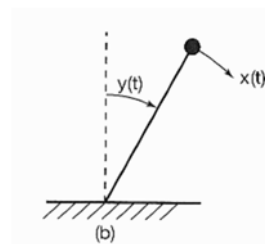
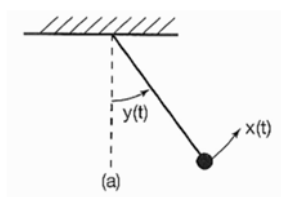
$$y(t) = x(t + 1)$$

$$y(t) = x(t) \cos(t + 1) \text{ ???}$$

Stability

Stable systems

- Small inputs lead to responses that do not diverge
- Every bounded input excites a bounded output
 - Bounded-input bounded-output stable (BIBO stable)
 - For all $|x(t)| < a$, then $|y(t)| < b$, for all t



- Balance in a bank account?

$$y[n] = 1.01y[n - 1] + x[n]$$

▪ Example 1.13: Stability

$$|x(t)| < a$$

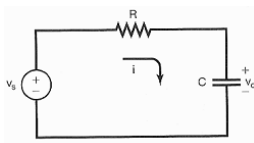
$$S_1 : y(t) = t x(t)$$

$$S_2 : y(t) = e^{x(t)}$$

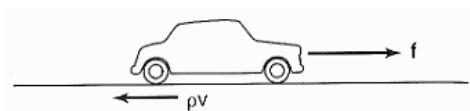
▪ Time Invariance

▪ Time-invariant systems

- Behavior & characteristics of system are **fixed over time**



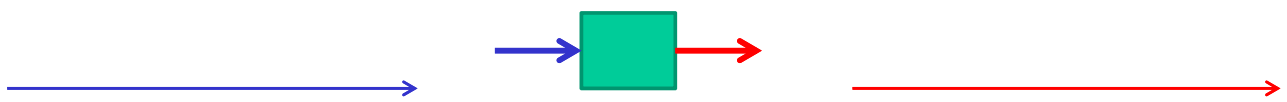
$$\Rightarrow \frac{dv_c(t)}{dt} + \frac{1}{RC}v_c(t) = \frac{1}{RC}v_s(t)$$



$$\Rightarrow \frac{dv(t)}{dt} + \frac{\rho}{m}v(t) = \frac{1}{m}f(t)$$

- A **time shift** in the **input** signal results in an **identical time shift** in the **output** signal

$$x[n] \rightarrow y[n] \iff x[n - n_0] \rightarrow y[n - n_0]$$



▪ Time Invariance

- Example of time-invariant system (Example 1.14)

$$y(t) = \sin [x(t)]$$

$$x_1(t)$$

$$y_1(t) = \sin [x_1(t)]$$

$$x_2(t) = x_1(t - t_0)$$

$$y_2(t) = \sin [x_2(t)] = \sin [x_1(t - t_0)]$$

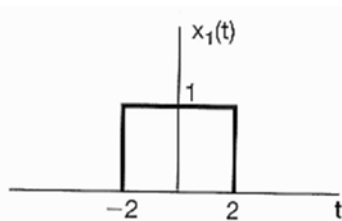
$$y_1(t - t_0) = \sin [x_1(t - t_0)]$$

$$y_2(t) = y_1(t - t_0)$$

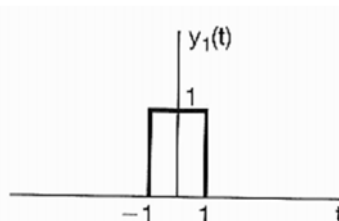
▪ Time Invariance

- Example of time-varying system (Example 1.16)

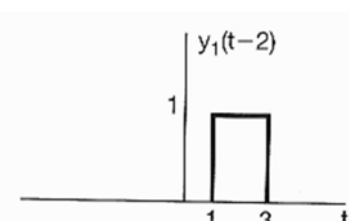
$$y(t) = x(2t)$$



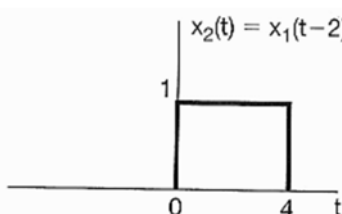
(a)



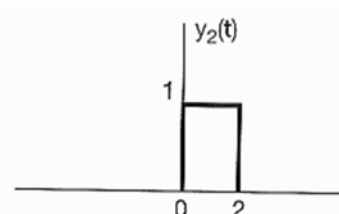
(b)



(e)



(c)



(d)

▪ Linearity

▪ Linear systems

- If an **input** consists of the **weighted sum** of several signals, then the **output** is the **superposition** of the **responses** of the system to **each** of those signals

$$x_1[n] \rightarrow y_1[n]$$

$$x_2[n] \rightarrow y_2[n]$$

IF (1) $x_1[n] + x_2[n] \rightarrow y_1[n] + y_2[n]$ (additivity)

(2) $a x_1[n] \rightarrow a y_1[n]$ (scaling or homogeneity)

a : any complex constant

THEN, the system is **linear**

▪ Linearity

▪ Linear systems

- In general, a, b : any complex constants

$$ax_1[n] + bx_2[n] \rightarrow ay_1[n] + by_2[n] \quad \text{for DT}$$

$$ax_1(t) + bx_2(t) \rightarrow ay_1(t) + by_2(t) \quad \text{for CT}$$

- OR,

$$x[n] = \sum_k a_k x_k[n] = a_1 x_1[n] + a_2 x_2[n] + \dots$$

$$\rightarrow y[n] = \sum_k a_k y_k[n] = a_1 y_1[n] + a_2 y_2[n] + \dots$$

This is known as the **superposition property**

▪ Linearity

▪ Example 1.17: $S : y(t) = tx(t)$

$$x_1(t) \rightarrow y_1(t) = tx_1(t)$$

$$x_2(t) \rightarrow y_2(t) = tx_2(t)$$

$$x_3(t) = ax_1(t) + bx_2(t)$$

$$\rightarrow y_3(t) = tx_3(t)$$

$$= t(ax_1(t) + bx_2(t)) = atx_1(t) + btx_2(t)$$

$$= ay_1(t) + by_2(t)$$

▪ Linearity

▪ Example 1.18: $S : y(t) = (x(t))^2$

$$x_1(t) \rightarrow y_1(t) = (x_1(t))^2$$

$$x_2(t) \rightarrow y_2(t) = (x_2(t))^2$$

$$x_3(t) = ax_1(t) + bx_2(t)$$

$$\rightarrow y_3(t) = (x_3(t))^2 = (ax_1(t) + bx_2(t))^2$$

$$= a^2(x_1(t))^2 + b^2(x_2(t))^2 + 2abx_1(t)x_2(t)$$

$$= a^2y_1(t) + b^2y_2(t) + 2abx_1(t)x_2(t)$$

▪ Linearity

▪ Example 1.20: $S : y[n] = 2x[n] + 3$

$$x_1[n] \rightarrow y_1[n] = 2x_1[n] + 3$$

$$x_2[n] \rightarrow y_2[n] = 2x_2[n] + 3$$

$$x_3[n] = ax_1[n] + bx_2[n]$$

$$\rightarrow y_3[n] = 2x_3[n] + 3$$

$$= 2(ax_1[n] + bx_2[n]) + 3$$

$$= a(2x_1[n] + 3) + b(2x_2[n] + 3) + 3 - 3a - 3b$$

$$= ay_1[n] + by_2[n] + 3(1 - a - b)$$

▪ Linearity

▪ Example 1.20: $S : y[n] = 2x[n] + 3$

$$x_1[n] \rightarrow y_1[n] = 2x_1[n] + 3$$

$$x_2[n] \rightarrow y_2[n] = 2x_2[n] + 3$$

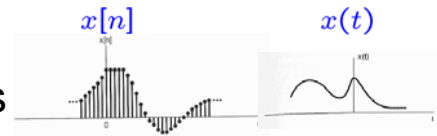
▪ However,

$$y_1[n] - y_2[n] = (2x_1[n] + 3) - (2x_2[n] + 3)$$

$$= 2[x_1[n] - x_2[n]]$$

It is a **incrementally linear system**

- Introduction
- Continuous-Time & Discrete-Time Signals
- Transformations of the Independent Variable



- Time Shift $x[n - n_0]$ $x(t - t_0)$ $x(-t) = x(t), x[-n] = x[n]$
- Time Reversal $x[-n]$ $x(-t)$ $x(-t) = -x(t), x[-n] = -x[n]$
- Time Scaling $x[an]$ $x(at)$ $\mathcal{E}v\{x[n]\} = \frac{1}{2}[x[n] + x[-n]]$
- Periodic Signals $x(t) = x(t + T)$ $\mathcal{O}d\{x[n]\} = \frac{1}{2}[x[n] - x[-n]]$
- Even & Odd Signals $x[n] = x[n + N]$

- Exponential & Sinusoidal Signals
- The Unit Impulse & Unit Step Functions
- Continuous-Time & Discrete-Time Systems
- Basic System Properties

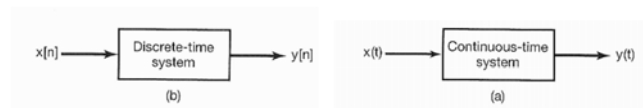
$$\phi_k(t) = e^{jk\omega_0 t}, k = 0, \pm 1, \dots$$

$$\phi_k[n] = e^{jk\omega_0 n}, k = 0, \dots, N - 1$$

$$\delta[n], u[n]$$

$$\delta(t), u(t)$$

- Systems with or without memory
- Invertibility & Inverse Systems
- Causality
- Stability
- Time Invariance
- Linearity



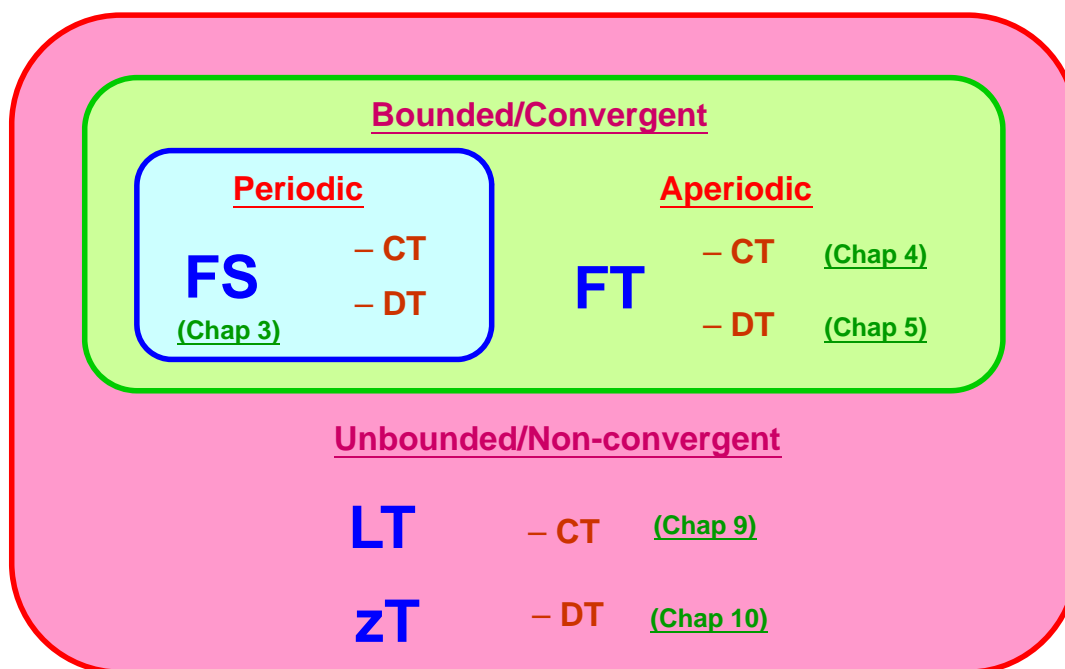
$$x[n] \rightarrow y[n]$$

$$x(t) \rightarrow y(t)$$

Flowchart

Signals & Systems (Chap 1)

LTI & Convolution (Chap 2)



Time-Frequency (Chap 6)

Communication (Chap 8)

CT-DT (Chap 7)

Control (Chap 11)

■ Problem 1.26 (Page 61)

$$x[n] = \cos\left(\frac{\pi}{8} n^2\right)$$

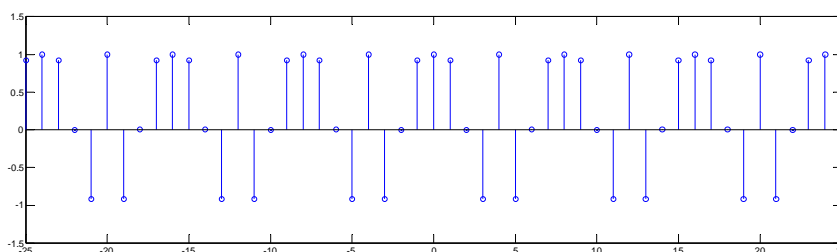
```

L = 25;
n = -L:L;

x = cos( pi/8 * (n.^2) );

figure(1)
stem( n, x, 'o' ); hold on;

axis([-L L -1.5 1.5])
    
```


 ■ Problem 1.27 (Page 62)

(a) $y(t) = x(t - 2) + x(2 - t)$ Time-Invariant?

$$x_1(t) \rightarrow y_1(t) \quad y_1(t) = x_1(t - 2) + x_1(2 - t)$$

$$x_2(t) \rightarrow y_2(t) \quad y_2(t) = x_2(t - 2) + x_2(2 - t)$$

$$x_2(t) = x_1(t - t_0)$$

$$\Rightarrow y_2(t) = x_2(t - 2) + x_2(2 - t)$$

$$= x_1((t - t_0) - 2)$$

$$+ x_1(2 - (t - t_0))$$

$$= x_1(t - t_0 - 2)$$

$$+ x_1(2 - t + t_0)$$

$$\Rightarrow y_2(t) = x_1(t - 2 - t_0)$$

$$+ x_1(2 - t - t_0)$$

$$= x_1(t - t_0 - 2)$$

$$+ x_1(2 - t - t_0)$$

▪ Problem 1.27 (Page 62)

(a) $y(t) = x(t - 2) + x(2 - t)$ Time-Invariant?

$$x_1(t) = \delta(t)$$

$$y_1(t) = \delta(t - 2) + \delta(2 - t)$$

$$x_2(t) = \delta(t - 3)$$

$$y_2(t) = \delta(t - 2 - 3) + \delta(2 - t - 3)$$

$$= \delta(t - 5) + \delta(-1 - t)$$

$$\Rightarrow y_1(t - 3) = \delta(t - 3 - 2) + \delta(2 - (t - 3))$$

$$= \delta(t - 5) + \delta(5 - t)$$