

93學年 第一學期

# 非線性系統分析 Nonlinear Systems Analysis

Lecture 1

## Course Information & Introduction

連豐力

台大電機系

Sep04 – Jan05

### Syllabus

#### Course Information:

- No: 921 U3390
- Time: Thu 1:30am-4:20pm
- Room: EE-BLH 620
- Website:  
<http://cc.ee.ntu.edu.tw/~fengli/Teaching/NonlinearSystems>

#### Instructor:

- 連豐力(Feng-Li Lian)
- Office: EE-BLH 620
- Email: [fengli@ntu.edu.tw](mailto:fengli@ntu.edu.tw)
- Phone: 02-3366-3606

#### Grading:

- Homework (30%), or (25%)
- Midterm (30%), or (25%)
- Final (40%), or (30%)
- Project (20 %)

#### Textbook:

- Nonlinear Systems  
by Hassan Khalil, 3<sup>rd</sup> Edition, 2002

#### Reference:

- Nonlinear Systems: Analysis, Stability and Control,  
by Shanker Sastry, 1999
- Nonlinear Systems Analysis  
by Vidyasagar, 2<sup>nd</sup> Edition, 1993
- Control in an Information Rich World,  
Report of the Panel on Future Directions in Control, Dynamics, and Systems.  
<http://www.cds.caltech.edu/~murray/cdspanel/report/cdspanel-15aug02.pdf>

## Course Outline

- **18 weeks (9/16/02-1/13/03)**
  - **Exam:** 2 weeks (11/25, 1/13),
  - **Holiday:** none
  - Total **16** week lectures,
  - **Homework assignments:** approximately 8-10;
    - **No late homework accepted**
- **Basic analysis**
  1. Introduction (1 week)
  2. Math + Second-order systems (3)
  3. Fundamental properties (2~3)
  4. Lyapunov stability (4~5)
- **Analysis of feedback systems**
  5. Input-output stability (1)
  6. Passivity (1)
  7. Frequency domain analysis
- **Advanced analysis**
  8. Advanced stability analysis (2)
  9. Stability of perturbed systems
  10. Perturbation theory & averaging
  11. Singular perturbations
- **Nonlinear control**
  12. Feedback control
  13. Feedback linearization
  14. Nonlinear design tools

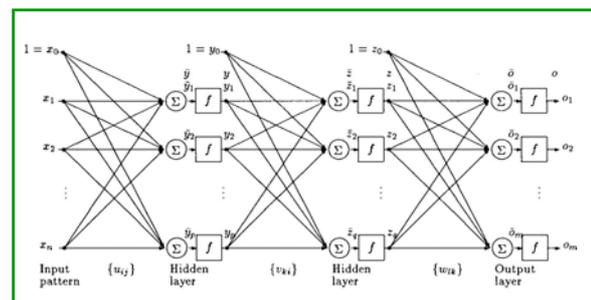
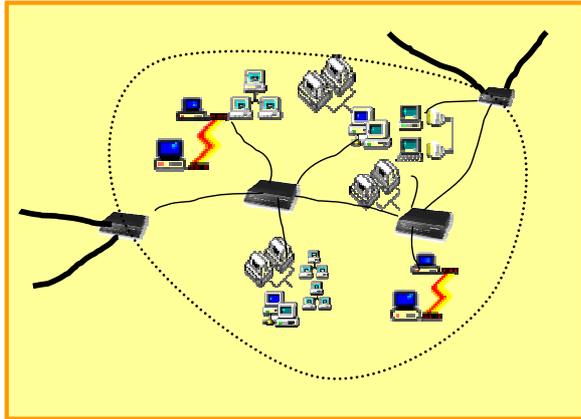
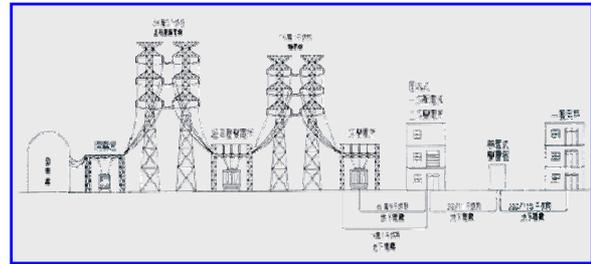
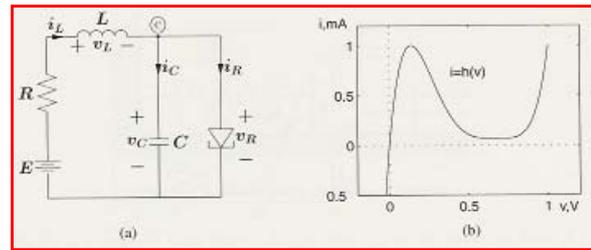
## Today's Outline

- **Engineering Examples**
  - Electrical engineering
  - Mechanical engineering
  - Chemical engineering
  - Transportation, etc.
- **Analysis & Design Methodology**
- **Nonlinear Models & Phenomena**
- **Examples of Nonlinear Systems**

## Engineering Examples - 1

### Electrical Engineering:

- Tunnel-Diode Circuit
- Electrical Power Transmission
- Internet, Communication Network
- Artificial Neural Network



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Nonlinear Systems Analysis

Ch1-5

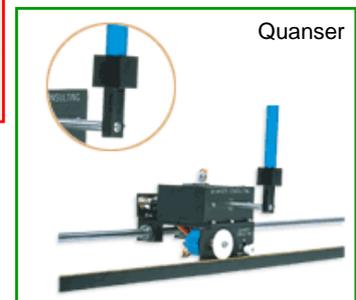
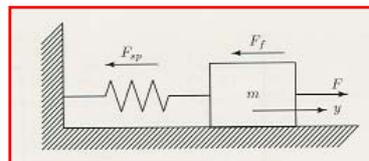
## Engineering Examples - 2

### Mechanical Engineering:

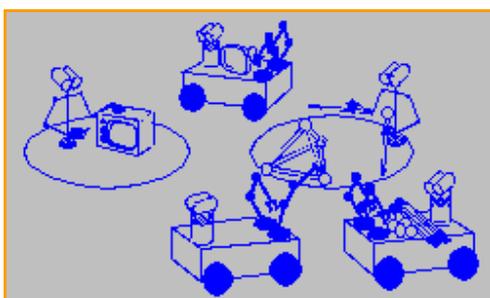
- Mass-Spring System
- (Inverted) Pendulum
- Robot Arms, Walking robots
- Vehicle Engine/Speed Control
- Fluid Dynamics
- Manufacturing/Machining Systems



SAE



Quanser



UMass  
Robotics



UMERC/  
RMS

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Nonlinear Systems Analysis

Ch1-6

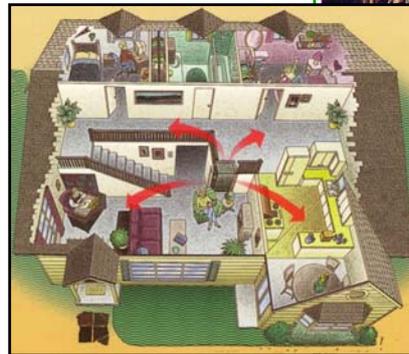
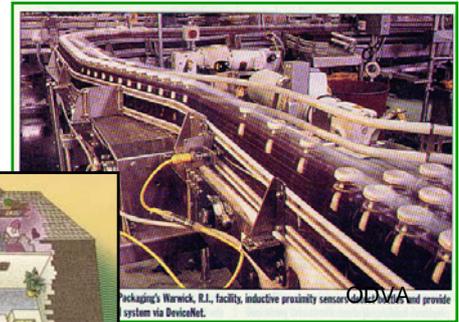
## Engineering Examples - 3

### Chemical Engineering:

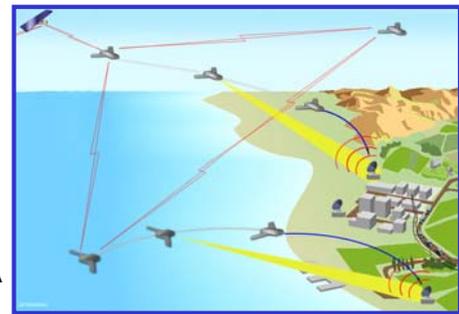
- Petroleum Engineering
- Food Processing
- Waster Material Handling
- Heating, Ventilation & Air Conditioning

### Transportation:

- Intelligent Highway System
- Air Traffics Management
- Unmanned Air/Ground Vehicle



Berkeley  
PATH



DARPA  
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P&N Distribution

ODVA

## Analysis & Design Philosophy

Problem

Modeling

Analysis

Design  
(Control)

### Mathematics, Statistics:

- Differential Equations, Linear Algebra, Probability, Stochastic Processes

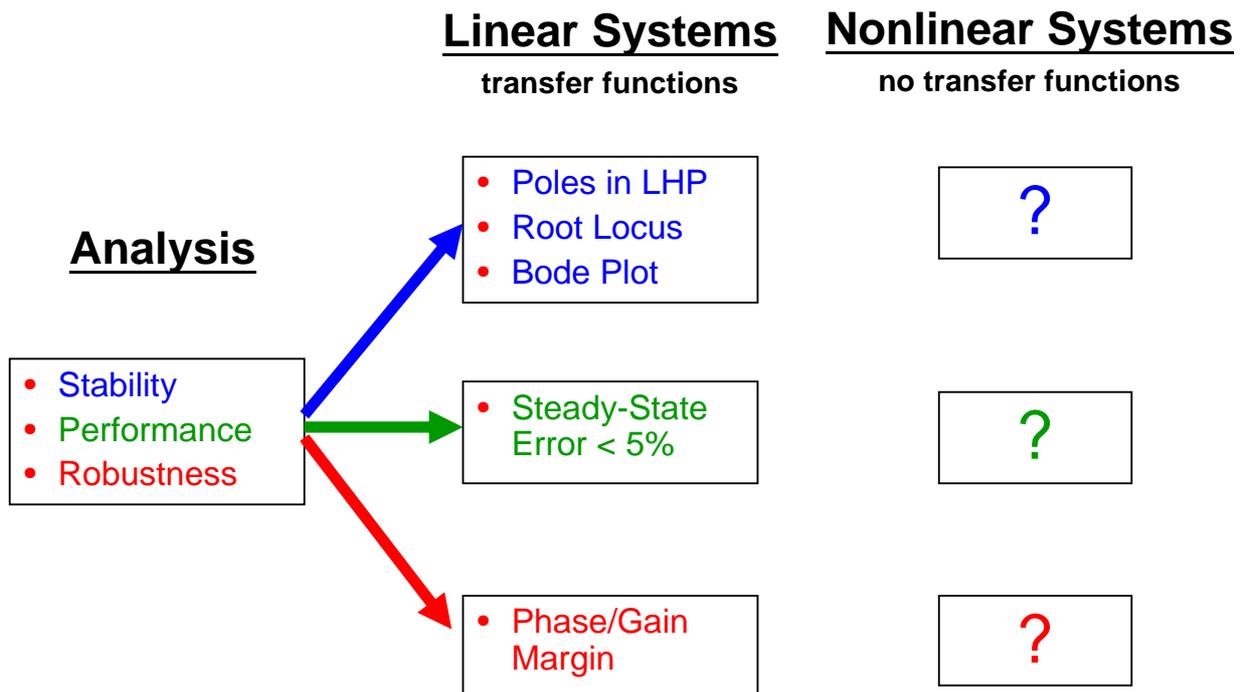
### Physics, Chemistry:

- Electronics, Electrical Circuits, Electromagnetics, Dynamics, Thermodynamics, Heat Transfer

### Linear v.s. Nonlinear

- Stability
- Performance
- Robustness

- Hardware vs Software
- Optimization vs Control
  - Optimal Control, Robust Control, Adaptive Control, Nonlinear Control



## Essentially Nonlinear Phenomena - 1

- **Finite Escape Time**
  - State goes to infinity in finite time
- **Multiple Isolated Equilibria**
  - Nonlinear systems has more than one isolated equilibrium points.
  - The state convergence depends on the initial conditions.
- **Limit Cycles**
  - Go into an oscillation of fixed magnitude and frequency, irrespective of the initial state.
- **Subharmonic, harmonic, almost-periodic oscillations**
  - Oscillation frequencies are submultiples or multiples of the input frequency.

## Essentially Nonlinear Phenomena - 2

- **Chaos**
  - Complicated steady-state behavior
  - Sometime random
- **Multiple modes of behavior**
  - Unforced systems may have more than one limit cycle
  - Forced systems with periodic excitation may exhibit harmonic, sub-harmonic, or complicated steady-state behavior, depending upon the amplitude and frequency of the input.
  - Exhibit discontinuity jump even though under smoothly changed input.

## Examples of Nonlinear Models

- **Examples:**
  - Pendulum Equation
  - Tunnel-Diode Circuit
  - Mass-Spring System
  - Negative-Resistance Oscillator
  - Artificial Neural Network
  - Adaptive Control
- **Common Nonlinearities**
  - Relay
  - Saturation
  - Dead zone
  - Quantization

## Pendulum - 1

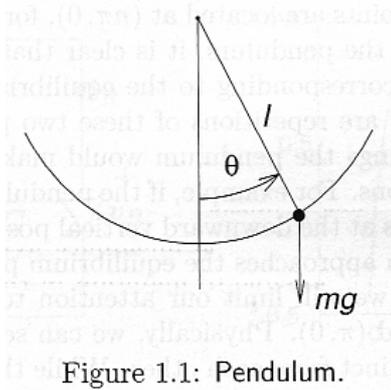


Figure 1.1: Pendulum.

Using Newton's Second Law,  
Write the equation of motion  
in the tangential direction:

$$ml\ddot{\theta} = -mg \sin \theta - k\dot{\theta}$$

State model (let  $x_1 = \theta, x_2 = \dot{\theta}$ ):

$$\begin{aligned}\dot{x}_1 &= x_2 \\ \dot{x}_2 &= -\frac{g}{l} \sin x_1 - \frac{k}{m} x_2\end{aligned}$$

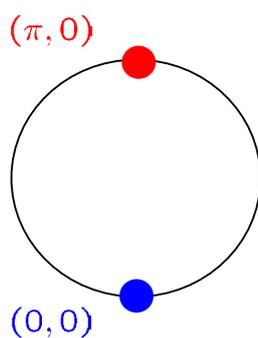
## Pendulum - 2

Equilibrium points (let  $\dot{x}_1 = \dot{x}_2 = 0$ ):

$$\begin{aligned}0 &= x_2 \\ 0 &= -\frac{g}{l} \sin x_1 - \frac{k}{m} x_2\end{aligned}$$

Equilibrium points are  $(n\pi, 0), n = 0, \pm 1, \pm 2, \dots$ ,  
or, physically,  $(0, 0)$  and  $(\pi, 0)$ .

Question? Which one is stable or unstable?



## Tunnel-Diode Circuit - 1

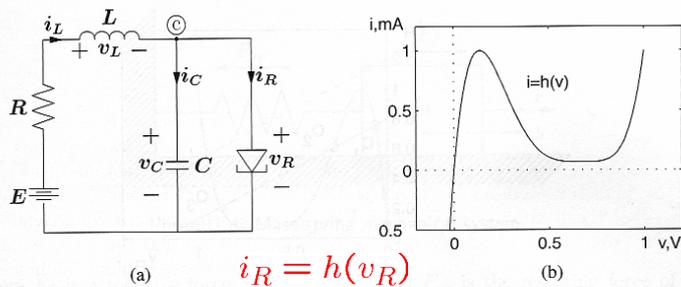


Figure 1.2: (a) Tunnel-diode circuit; (b) Tunnel-diode  $v_R$ - $i_R$  characteristic.

Kirchhoff's current/voltage law:

$$i_C + i_R - i_L = 0 \quad (\text{KCL})$$

$$v_C - E + Ri_L + v_L = 0 \quad (\text{KVL})$$

State model:

- state:  $x_1 = v_C, x_2 = i_L$ , and input:  $u = E$ ,

-  $i_C = C \frac{dv_C}{dt}, v_L = L \frac{di_L}{dt}$

$$\dot{x}_1 = \frac{1}{C}[-h(x_1) + x_2]$$

$$\dot{x}_2 = \frac{1}{L}[-x_1 - Rx_2 + u]$$

## Tunnel-Diode Circuit - 2

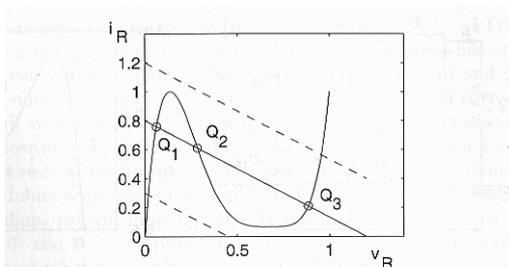


Figure 1.3: Equilibrium points of the tunnel-diode circuit.

Equilibrium points:

$$0 = -h(x_1) + x_2$$

$$0 = -x_1 - Rx_2 + u$$

That is, the roots of:

$$h(x_1) = \frac{E}{R} - \frac{1}{R}x_1$$

## Mass-Spring System - 1

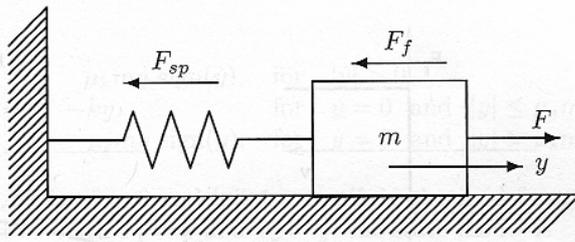


Figure 1.4: Mass-spring mechanical system.

Newton's Law of motion:

$$m\ddot{y} + F_f + F_{sp} = F$$

$F_f$  due to friction:

$$\begin{aligned} \text{Static} &= \text{limit to } \pm \mu_s mg \\ \text{Coulomb} &= \begin{cases} -\mu_k mg, & \text{for } v < 0 \\ \mu_k mg, & \text{for } v > 0 \end{cases} \\ \text{Viscous} &= h(v) \approx cv \end{aligned}$$

## Mass-Spring System - 2

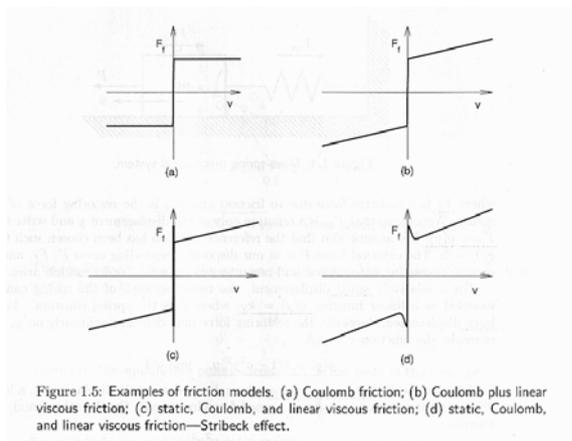


Figure 1.5: Examples of friction models. (a) Coulomb friction; (b) Coulomb plus linear viscous friction; (c) static, Coulomb, and linear viscous friction; (d) static, Coulomb, and linear viscous friction—Stribeck effect.

$F_{sp}$  due to the spring ( $= g(y)$ ):

$$\begin{aligned} g(y) &= ky, \text{ (small } y) \\ &= k(1 - a^2 y^2)y, |ay| < 1, \\ &\quad \text{(softening spring)} \\ &= k(1 + a^2 y^2)y, \\ &\quad \text{(hardening spring)} \end{aligned}$$

## Mass-Spring System - 3

Linear spring, static friction, Coulumb friction,  
linear viscous friction,  $F = 0$ :

$$m\ddot{y} + ky + c\dot{y} + \eta(y, \dot{y}) = 0$$

where:

$$\eta(y, \dot{y}) = \begin{cases} \mu_k mg \operatorname{sign}(\dot{y}), & \text{for } |\dot{y}| > 0 \\ -ky, & \text{for } \dot{y} = 0 \text{ and } |y| \leq \mu_s mg/k \\ -\mu_s mg \operatorname{sign}(y), & \text{for } \dot{y} = 0 \text{ and } |y| > \mu_s mg/k \end{cases}$$

State model (let  $x_1 = y, x_2 = \dot{y}$ ):

$$\begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= -\frac{k}{m}x_1 - \frac{c}{m}x_2 - \frac{1}{m}\eta(x_1, x_2) \end{aligned}$$

Remark:

1. An equilibrium set, rather than points;
2. A discontinuous function of the state.

## Negative-Resistance Oscillator - 1

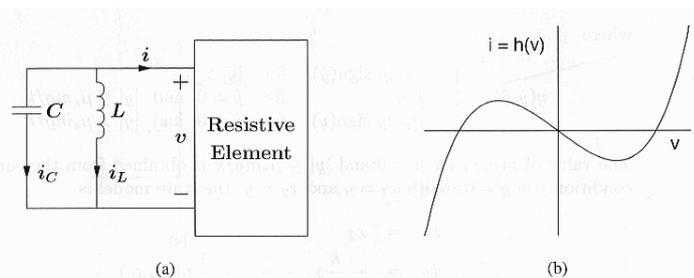


Figure 1.6: (a) Basic oscillator circuit; (b) Typical driving-point characteristic.

Kirchhoff's current law:

$$i_C + i_L + i = 0 \text{ or } C \frac{dv}{dt} + \frac{1}{L} \int_{-\infty}^t v(s) ds + h(v) = 0$$

( $dt$ ) and ( $\times L$ ):

$$CL \frac{d^2v}{dt^2} + v + Lh'(v) \frac{dv}{dt} = 0$$

$$\begin{aligned} h(0) &= 0 \\ h'(0) &< 0 \\ h(v) &\rightarrow \infty \text{ as } v \rightarrow \infty \\ h(v) &\rightarrow -\infty \text{ as } v \rightarrow -\infty \end{aligned}$$

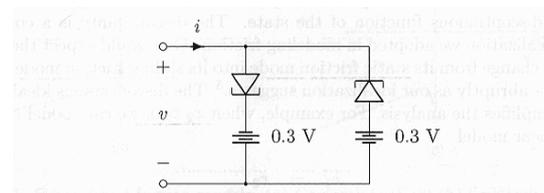


Figure 1.7: A negative-resistance twin-tunnel-diode circuit.

## Negative-Resistance Oscillator - 2

Let  $\tau = t/\sqrt{CL}, \epsilon = \sqrt{L/C}$ :

$$\ddot{v} + \epsilon h'(v)\dot{v} + v = 0$$

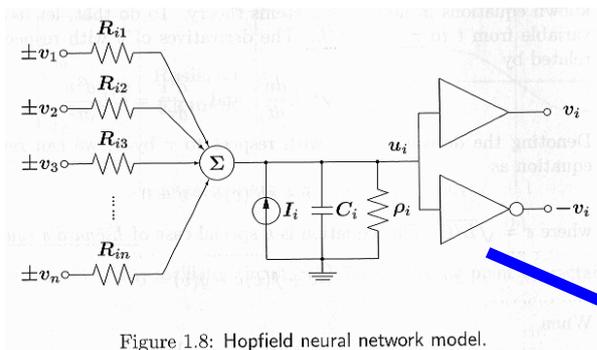
The **Lienard's equation**:

$$\ddot{v} + f(v)\dot{v} + g(v) = 0$$

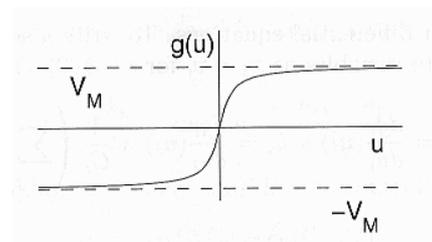
The **Van der Pol equation**:

$$\ddot{v} - \epsilon(1 - v^2)\dot{v} + v = 0$$

## Artificial Neural Network (ANN) - 1



$$v_i = g_i(u_i)$$



$g_i(\cdot)$ : a sigmoid function:

$$g_i(u_i) = \frac{2V_M}{\pi} \tan^{-1}\left(\frac{\lambda \pi u_i}{2V_M}\right), \lambda > 0$$

$$g_i(u_i) = V_M \frac{e^{\lambda u_i} - e^{-\lambda u_i}}{e^{\lambda u_i} + e^{-\lambda u_i}} = V_M \tanh(\lambda u_i), \lambda > 0$$

## Artificial Neural Network (ANN) - 2

Kirchhoff's current law:

$$\begin{aligned} C_i \frac{du_i}{dt} &= \sum_j \frac{1}{R_{ij}} (\pm v_j - u_i) - \frac{1}{\rho} u_i + I_i \\ &= \sum_j T_{ij} v_j - \frac{1}{R_i} u_i + I_i \end{aligned}$$

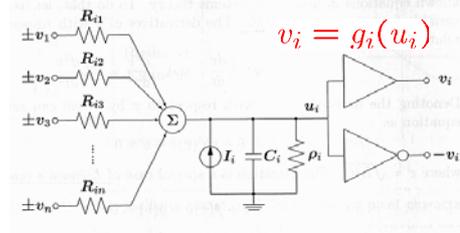


Figure 1.8: Hopfield neural network model.

State model (let  $x_i = v_i, i = 1, 2, \dots, n$ ):

$$\begin{aligned} \dot{x}_i &= \frac{dg_i}{du_i}(u_i) \times \dot{u}_i \\ &= \frac{dg_i}{du_i}(u_i) \times \frac{1}{C_i} \left[ \sum_j T_{ij} x_j - \frac{1}{R_i} u_i + I_i \right] \end{aligned}$$

Let  $h_i(x_i) = \frac{dg_i}{du_i}(u_i)|_{u_i=g_i^{-1}(x_i)}$ ,  
and  $h_i(x_i) > 0, \forall x_i \in (-V_M, V_M)$

Equilibrium points

$$0 = \sum_j T_{ij} x_j - \frac{1}{R_i} g^{-1}(x_i) + I_i$$

## Direct Model Reference Adaptive Control (MRAC) - 1

Plant model:  $\dot{y}_p = a_p y_p + k_p u$

Reference model:  $\dot{y}_m = a_m y_m + k_m r$

Input:  $u(t) = \theta_1^* r(t) + \theta_2^* y_p(t)$

$$\theta_1^* = \frac{k_m}{k_p} \text{ and } \theta_2^* = \frac{a_m - a_p}{k_p}$$

Gradient algorithm:

$$\begin{aligned} \dot{\theta}_1 &= -\gamma (y_p - y_m) r \\ \dot{\theta}_2 &= -\gamma (y_p - y_m) y_p \end{aligned}$$

Error state:

$$\begin{aligned} e_0 &= y_p - y_m, \\ \phi_1 &= \theta_1 - \theta_1^*, \\ \phi_2 &= \theta_2 - \theta_2^* \end{aligned}$$

## Error dynamics:

$$\begin{aligned} \dot{e}_0 &= a_m e_0 + k_p \phi_1 r(t) + k_p \phi_2 [e_0 + y_m(t)] \\ \dot{\phi}_1 &= -\gamma e_0 r(t) \\ \dot{\phi}_2 &= -\gamma e_0 [e_0 + y_m(t)] \end{aligned}$$

## Equilibrium points:

$$\begin{aligned} 0 &= a_m e_0 + k_p \phi_1 r(t) + k_p \phi_2 [e_0 + y_m(t)] \\ 0 &= -\gamma e_0 r(t) \\ 0 &= -\gamma e_0 [e_0 + y_m(t)] \end{aligned}$$

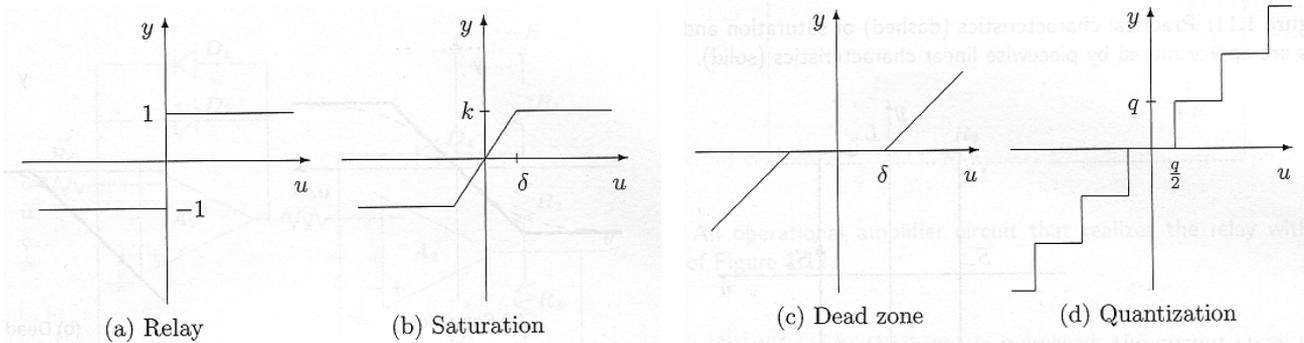
If  $r(t) = 0$ , i.e.,  $y_m(t) = 0$

$$\begin{aligned} 0 &= [a_m + k_p \phi_2] e_0 \\ 0 &= -\gamma e_0^2 \end{aligned}$$

At steady state:

$$\begin{aligned} e_0 &\rightarrow 0 \\ \phi_i &\rightarrow 0 \quad (?) \end{aligned}$$

## Common Nonlinearities



## Homework Assignment 1

- Due 9/30 in class:
  - 1) 1.11 (page 26)
  - 2) 1.15 (page 28)
  - 3) 1.21 (page 33)
- Outline of a technical paper/report:
  - Introduction
  - Background/Literature Survey
  - Problem Formulation
  - Analysis/Design/Proposed Solutions
  - Validation & Verification by Simulation/Experiment
  - Conclusions & Future Work
  - References
- 每一個作業：
  - 請註明：作業次別，姓名，學號，系級，日期
  - 請說明清楚該次作業之撰寫為：
    - 1) 自行獨立完成作業，且未參考過去之作業或解答或與其他同學討論。
    - 2) 自行獨立完成作業，曾與同學或學長討論之後，但未參考過去之作業或解答。
    - 3) 曾參考其他同學或過去之作業或解答之後，完成作業。
    - 4) 完全抄襲其他相關資料。

1.11 A phase-locked loop [64] can be represented by the block diagram of Figure 1.20. Let  $\{A, B, C\}$  be a minimal realization of the scalar, strictly proper transfer function  $G(s)$ . Assume that all eigenvalues of  $A$  have negative real parts,  $G(0) \neq 0$ , and  $\theta_i = \text{constant}$ . Let  $z$  be the state of the realization  $\{A, B, C\}$ .

(a) Show that the closed-loop system can be represented by the state equations

$$\dot{z} = Az + B \sin e, \quad \dot{e} = -Cz$$

(b) Find all equilibrium points of the system.

(c) Show that when  $G(s) = 1/(\tau s + 1)$ , the closed-loop model coincides with the model of a pendulum equation.

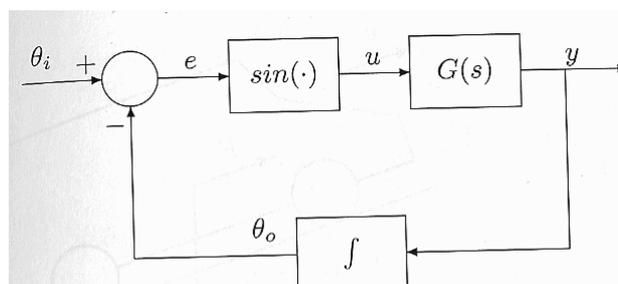


Figure 1.20: Exercise 1.11.

**1.15** Consider the inverted pendulum of Figure 1.24 [110]. The pivot of the pendulum is mounted on a cart that can move in a horizontal direction. The cart is driven by a motor that exerts a horizontal force  $F$  on the cart. The figure shows also the forces acting on the pendulum, which are the force  $mg$  at the center of gravity, a horizontal reaction force  $H$ , and a vertical reaction force  $V$  at the pivot. Writing horizontal and vertical Newton's laws at the center of gravity of the pendulum yields

$$m \frac{d^2}{dt^2}(y + L \sin \theta) = H \quad \text{and} \quad m \frac{d^2}{dt^2}(L \cos \theta) = V - mg$$

Taking moments about the center of gravity yields the torque equation

$$I\ddot{\theta} = VL \sin \theta - HL \cos \theta$$

while a horizontal Newton's law for the cart yields

$$M\ddot{y} = F - H - k\dot{y}$$

Here  $m$  is the mass of the pendulum,  $M$  is the mass of the cart,  $L$  is the distance from the center of gravity to the pivot,  $I$  is the moment of inertia of the pendulum with respect to the center of gravity,  $k$  is a friction coefficient,  $y$  is the displacement of the pivot,  $\theta$  is the angular rotation of the pendulum (measured clockwise), and  $g$  is the acceleration due to gravity.

- (a) Carrying out the indicated differentiation and eliminating  $V$  and  $H$ , show that the equations of motion reduce to

$$I\ddot{\theta} = mgL \sin \theta - mL^2\ddot{\theta} - mL\ddot{y} \cos \theta$$

$$M\ddot{y} = F - m(\ddot{y} + L\ddot{\theta} \cos \theta - L\dot{\theta}^2 \sin \theta) - k\dot{y}$$

- (b) Solving the foregoing equations for  $\ddot{\theta}$  and  $\ddot{y}$ , show that

$$\begin{bmatrix} \ddot{\theta} \\ \ddot{y} \end{bmatrix} = \frac{1}{\Delta(\theta)} \begin{bmatrix} m + M & -mL \cos \theta \\ -mL \cos \theta & I + mL^2 \end{bmatrix} \begin{bmatrix} mgL \sin \theta \\ F + mL\dot{\theta}^2 \sin \theta - k\dot{y} \end{bmatrix}$$

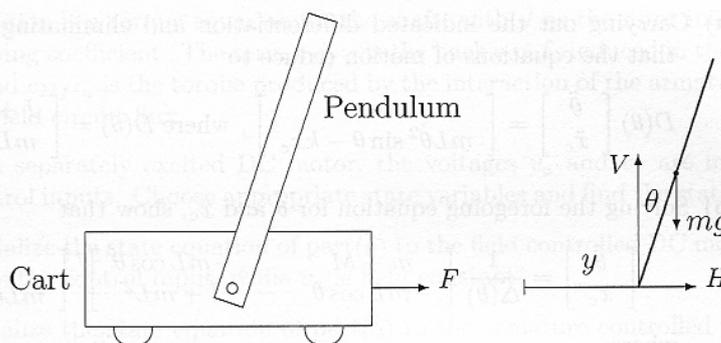


Figure 1.24: Inverted pendulum of Exercise 1.15.

where

$$\Delta(\theta) = (I + mL^2)(m + M) - m^2L^2 \cos^2 \theta \geq (I + mL^2)M + mI > 0$$

- (c) Using  $x_1 = \theta$ ,  $x_2 = \dot{\theta}$ ,  $x_3 = y$ , and  $x_4 = \dot{y}$  as the state variables and  $u = F$  as the control input, write down the state equation.

**1.21** The valves in the hydraulic system of Figure 1.30 obey the flow relationships  $w_1 = k_1\sqrt{p_1 - p_2}$  and  $w_2 = k_2\sqrt{p_2 - p_a}$ . The pump has the characteristic shown in Figure 1.29 for  $(p_1 - p_a)$  versus  $w_p$ . The various components and variables are defined in the previous two exercises.

- Using  $(p_1 - p_a)$  and  $(p_2 - p_a)$  as the state variables, find the state equation.
- Find all equilibrium points of the system.

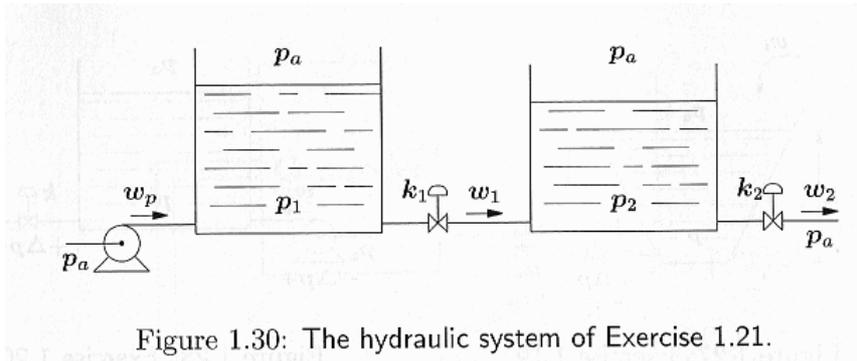


Figure 1.30: The hydraulic system of Exercise 1.21.