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### Problem 5-1:

[Ref: HW5\_劉仲恩 Chung-En Liu\_Digital Control\_HW5\_Design in State-Space Model\_20190510]

In this problem, we have to determine a linear state-feedback controller such that the

closed-loop poles match those given in the problem.

The system given in the problem is indicated in Equation (1-1) and Equation (1-2).

$x(k+1) = \begin{bmatrix} 1.0 & 0.1 \\ 0.5 & 0.1 \end{bmatrix} x(k) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(k)$	(1-1)
$y(k) = \begin{bmatrix} 1 & 1 \end{bmatrix} x(k)$	(1-2)

Since the closed-loop poles are situated at 0.1 and 0.25, the desired characteristic

equation can be written out as in Equation (1-3). We then get Equation (1-4) by expanding

Equation (1-3).

(z-0.1)(z-0.25)	(1-3)
$z^2 - 0.35z + 0.025$	(1-4)

We then set the state-feedback controller K as in Equation (1-5).

$K = \begin{bmatrix} K_1 \\ K_2 \end{bmatrix}^T$	(1-5)
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For state-feedback control system, the closed-loop characteristic equation can be expressed as of line two on page 14 in [1: Lian 2019], which is also shown here in Equation (1-6). By modifying Equation (1-6) and adopting the Eigenvalue Assignment

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as mentioned on page 14 and 24 in [1: Lian 2019], we obtain the closed-loop characteristic equation of the system as in Equation (1-7).

$\det(\lambda I - (F - HK)) = \lambda^n + (a_1 + k_1)\lambda^{n-1} + \dots + (a_n + k_n)$	(1-6)
$\det(zI - (F - HK))$	(1-7)

According to page 11 in [1: Lian 2019], we can see that the discrete time model with

given h can be expressed as in Equation (1-8). By comparing Equation (1-8) with the given system as indicated in Equation (1-1) and Equation (1-2), we can set matrices F and

H as in Equation (1-9) and Equation (1-10) respectively.

x(k+1) = Fx(k) + Hu(k)	(1-8)
$F = \begin{bmatrix} 1.0 & 0.1 \\ 0.5 & 0.1 \end{bmatrix}$	(1-9)
$H = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$	(1-10)

By substituting the values of matrices F and H from Equation (1-9) and Equation (1-

10) into Equation (1-7), Equation (1-7) can then be rewritten as Equation (1-11).

$$\det(zI - (\begin{bmatrix} 1.0 & 0.1 \\ 0.5 & 0.1 \end{bmatrix} - \begin{bmatrix} 1 \\ 0 \end{bmatrix} \begin{bmatrix} K_1 & K_2 \end{bmatrix})$$
(1-11)

By expanding Equation (1-11), we get Equation (1-12).

$$z^{2} + z(K_{1} - 1.1) + (0.5K_{2} - 0.1K_{1} + 0.05)$$
 (1-12)

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As Equation (1-12) should be equivalent to the desired characteristic equation as in

Equation (1-4), we get the following Equation (1-13) and Equation (1-14).

$K_1 - 1.1 = -0.35$	(1-13)
$0.5K_2 - 0.1K_1 + 0.05 = 0.025$	(1-14)

By solving Equation (1-13) and Equation (1-14), we obtain the values of  $K_1$  and

 $K_2$  as in Equation (1-15) and Equation (1-16).

$K_1 = 0.75$	(1-15)
$K_2 = 0.1$	(1-16)

The linear state-feedback controller can thus be expressed in the following.

$K = \begin{bmatrix} 0.75\\0.1\end{bmatrix}^T$	Ans. of Pro. 5-1

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# Problem 5-2:

Refs: [HW5\_王喻民 王喻民\_DCS\_HW-5\_Design in State-Space Model] → a [HW5\_林柏佑\_HW-5\_DesignInStateSpaceModel] → b

**(a)** 

The given system is denoted in Equation (2-1) and Equation (2-2)

$x(k+1) = \begin{bmatrix} 0.55 & 0.12\\ 0 & 0.67 \end{bmatrix} x(k) + \begin{bmatrix} 0.01\\ 0.16 \end{bmatrix} u(k)$	(2-1)
$y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} x(k)$	(2-2)

The discrete time state-space model can be written as Equation (2-3) and Equation (2-4)

x(k+1) = Fx(k) + Hu(k)	(2-3)
y(k) = Cx(k) + Du(k)	(2-4)

Hence, we can get  $F = \begin{bmatrix} 0.55 & 0.12 \\ 0 & 0.67 \end{bmatrix}$ ,  $H = \begin{bmatrix} 0.01 \\ 0.16 \end{bmatrix}$ , and  $C = \begin{bmatrix} 1 & 0 \end{bmatrix}$ .

A linear state-feedback controller is to be determined such that the desired closed-loop

characteristic equation in Equation (2-5):

$$z^2 - 0.63z + 0.21 = 0 \tag{2-5}$$

By the state-feedback law, u(k) = -Kx(k). Then, Equation (2-5) can be rewritten as

$$x(k+1) = \begin{bmatrix} 0.55 & 0.12 \\ 0 & 0.67 \end{bmatrix} x(k) + \begin{bmatrix} 0.01 \\ 0.16 \end{bmatrix} (-Kx(k))$$
$$= \left\{ \begin{bmatrix} 0.55 & 0.12 \\ 0 & 0.67 \end{bmatrix} - \begin{bmatrix} 0.01 \\ 0.16 \end{bmatrix} [k_1 \quad k_2] \right\} x(k)$$
(2-6)

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The characteristic equation is

$$det(zI - (F - HK))$$

$$= det(zI - \begin{bmatrix} 0.55 & 0.12 \\ 0 & 0.67 \end{bmatrix} + \begin{bmatrix} 0.01 \\ 0.16 \end{bmatrix} [k_1 & k_2])$$

$$= det(\begin{bmatrix} z - 0.55 + 0.01k_1 & -0.12 + 0.01k_2 \\ 0.16k_1 & z - 0.67 + 0.16k_2 \end{bmatrix})$$

$$= (z - 0.55 + 0.01k_1)(z - 0.67 + 0.16k_2)$$

$$- (0.16k_1)(-0.12 + 0.01k_2)$$

$$= z^2 + (0.01k_1 + 0.16k_2 - 1.22)z$$

$$+ (0.3685 + 0.0125k_1 - 0.088k_2)$$

$$= 0$$

$$(2-7)$$

By comparing the parameters of Equation (2-5) and Equation (2-7), we can get

$\int 0.01k_1 + 0.16k_2 - 1.22 = -0.63 \qquad \text{ (k_1 = 9.22)}$	Ans. of
$\left(0.3685 + 0.0125k_1 - 0.088k_2 = 0.21^{-1}\right)k_2 = 3.11$	Pro 5-2 a

Therefore, the linear state-feedback controller is  $K = \begin{bmatrix} 9.22 & 3.11 \end{bmatrix}$ 

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(b) The following figures are some code snippets and the simulation result.

The code in Figure 2-1 construct the original discrete-time system, and then use the state-feedback controller gain K, which is found in problem (a), to construct the closed-loop discrete-time system. Then, use Matlab function d2c() to construct the closed-loop continuous-time system.

### **Result 1: Simulation DT Controller only**

( <b>2</b> , <b>9</b> )
(2-8)

System in continuous time is derived:

sys = ss(A, B, C, []);	(2-9)

Transform the continuous system into discrete time in h = 0.2:

sys = c2d(sys,Ts);	(2-10)

Obtain the discrete state-space model:

[F,H,C,D] = ssdata(sys);	(2-11)
[F,H,C,D] = ssdata(sys);	(2-11)

Implement the controller:

$F2 = F-H^*K;$ (2-12)
-----------------------

Obtain the input, output and states with initial condition:

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sys2 = dss(F2, [], C, [], [], Ts);	
t = 0 : Ts : 4;	
[y, t, x] = initial(sys2, x0, t);	(2, 12)
u = -K * x';	(2-13)
x1 = x(:,1);	
x2 = x(:,2);	

Finally, plot the figures for states, input, and output:





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**Result 2: Simulation of DT Controller -> CT Controller** 

```
% Original discrete-time system
F = [0.55 0.12; 0 0.67];
H = [0.01; 0.16];
C = [1 0];
D = 0;
sys_ori = ss(F, H, C, D, 0.2);
% Linear state-feedback controller
K = [9.2222 3.1111];
% Closed-loop discrete-time system, h = 0.2
sys_cl = ss(F-H*K, H, C, D, 0.2)
% Closed-loop continuous-time system
sys_cl_ct = d2c(sys_cl)
Figure 2-1. System construction
```

The code in Figure 2-2 generates the input signal to do simulation. Because the

simulated system has already considered the state-feedback input, here the input signal is

zero.

```
% Generate input signal to simulate the system
t = 0:0.01:4;
u = zeros(size(t));
x0 = [1 0];
Figure 2-2. Input signal generation
```

The code in Figure 2-3 shows how to simulate the system and extract the input signals from states. Here use Matlab function *lsim()* to simulate the closed-loop continuous-time system. Store the output in variable y, time vector in t, and states in x.

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Then use x and the relation u = -Kx to extract input signal every 0.2 second. Finally,



5 1 0 0.5 Output Input -5 0 -10 -15 -0.5 5 10 1 2 3 15 20 0 4 Time (sec) Time (sec) 1 0.5 0 0.5 State 2 5.0-State 1 0 -1 -1.5 -0.5 0 2 3 2 3 4 0 1 4 1 Time (sec) Time (sec) Figure 2-4. Simulation Results [Ans. of Pro 5-2 b]

the simulation results are shown in Figure 2-4.

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### **<u>Result 3: Simulation of DT Controller + CT Plant</u>**



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## Problem 5-3

**Refs:** 

[HW5\_張峻豪 r07921012 張峻豪\_DCS\_HW5\_107 0510\_Design in State-Space Model] → a, c [HW5\_林柏宇 林柏宇\_DCS\_HW05\_20190510\_Design in state-space model] → b

**(a)** 

The system is described as:

x(k+1) = Fx(k) + Hu(k)	(3-1)
$\mathbf{y}(\mathbf{k}) = \mathbf{C}\mathbf{x}(\mathbf{k})$	(3-2)

Where

$\mathbf{F} = \begin{bmatrix} 0.25 & 0.5 \\ 1 & 2 \end{bmatrix}, \mathbf{H} = \begin{bmatrix} 1 \\ 4 \end{bmatrix}, \mathbf{C} = \begin{bmatrix} 1 & 0 \end{bmatrix}$	(3-3)
---	-------

(a) To bring the state to origin in two sampling intervals in closed-loop system, the

condition of deadbeat control should be satisfied.

The linear state-feedback controller is given:

$$u(k) = -Kx(k) + K_r r(k)$$
(3-4)

Where

$$\mathbf{K} = \begin{bmatrix} k_1 & k_2 \end{bmatrix} \tag{3-5}$$

The condition of deadbeat control for two time intervals:

$$det[\lambda I - F + HK] = \lambda^2$$
(3-6)

Substitute Equation (3-3) and Equation (3-5) into Equation (3-6):

$$\det \begin{bmatrix} \lambda - 0.25 + k_1 & -0.5 + k_2 \\ 4k_1 - 1 & \lambda - 2 + 4k_2 \end{bmatrix} = \lambda^2$$
(3-7)

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Summarize Equation (3-7):

$$\lambda^{2} + (k_{1} + 4k_{2} - 2.25)\lambda = \lambda^{2}$$
(3-8)

Therefore, the following condition should be met to satisfy Equation (3-8):

$$k_1 + 4k_2 - 2.25 = 0 \tag{3-9}$$

There are infinite solutions to Equation (3-9), one of solutions is:

v = [k		4]	Ans. of
$\kappa = [\kappa_1]$	$\kappa_2 = [-1.75]$	I	Pro 5-3 a

#### **(b)**

First, check out the controllability.

The controllability matrix is denoted in Equation (3-10):

$W_c = [H]$	$FH$ ] = $\begin{bmatrix} 1\\4 \end{bmatrix}$	$[ \begin{smallmatrix} 0.25 \\ 1 \end{smallmatrix}$		2.25 9	(3.10)
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As we can see,  $\det W_c = 0$ . Which means the system is not reachable and we cannot

#### reach arbitrary states from the origin.

However,  $\mathbf{x}(\mathbf{k}) = \begin{bmatrix} 2 \\ 8 \end{bmatrix}$  happen to be in the column space of  $W_c$ , so this point can be

reached.

Let's check out the following calculation process:

$$x(1) = Hx(0) + Fu(0)$$
  
=  $\begin{bmatrix} 0\\0 \end{bmatrix} + \begin{bmatrix} 1\\4 \end{bmatrix} u(0)$  (3.11)

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$$\begin{aligned} x(2) &= Hx(1) + Fu(1) \\ &= H[Hx(0) + Fu(0)] + Fu(1) \\ &= H^2x(0) + HFu(0) + Fu(1) \\ &= \begin{bmatrix} 0.5625 & 1.125 \\ 2.25 & 4.5 \end{bmatrix} x(0) + \begin{bmatrix} 2.25 \\ 9 \end{bmatrix} u(0) + \begin{bmatrix} 1 \\ 4 \end{bmatrix} u(1) \end{aligned}$$
(3.12)

Where  $x(2) = \begin{bmatrix} 2 \\ 8 \end{bmatrix}$  and  $x(0) = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$ , we can obtain:

$$\begin{cases} 2 = 2.25u(0) + u(1) \\ 8 = 9u(0) + 4u(1) \end{cases}$$
(3.13)

Therefore, we can find out the answer as:

(u(0) = 0)	Ans. of
u(1) = 2	Pro. 5-3 b

#### (c)

To estimate the states of system, the system should be observable. In advance, the

observer can be design to let the system meet the desired characteristic polynomial.

First of all, the observability matrix is:

$$W_o = \begin{bmatrix} C \\ CF \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0.25 & 0.5 \end{bmatrix}$$
(3-14)

From Equation (3-14), the observability matrix is full rank such that the system is

observable.

The observer is designed to estimate the state:

$$\hat{x}(k+1) = F\hat{x}(k) + Hu(k) + L[y(k) - C\hat{x}(k)]$$
 (3-15)

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Where

$\mathbf{L} = \begin{bmatrix} l_1 \\ l_2 \end{bmatrix}$	(3-16)
---	--------

The characteristic polynomial can be derived from:

$\det[\lambda I - F + LC] = 0$	(3-17)

Substitute Equation (3-3) and Equation (3-16) into Equation (3-17), and the

characteristic polynomial is:

$$\lambda^{2} + (-2.25 + l_{1})\lambda + (-2l_{1} + 0.5l_{2}) = 0$$
(3-18)

On the other hands, the desired characteristic polynomial:

$$(\lambda - 0.2)^2 = \lambda^2 - 0.4\lambda + 0.04 = 0$$
 (3-19)

Compare Equation (3-19) to Equation (3-20), the observer is derived:

$[l_1]$ [1.85]	Ans. of
$\mathbf{L} = \begin{bmatrix} l_2 \end{bmatrix} = \begin{bmatrix} 7.48 \end{bmatrix}$	Pro. 5-3 c

In conclusion, the states can be estimated by observer design. In advance, the

original characteristic polynomial can be brought to desired one.

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# References

### [1: Lian 2019]

Feng-Li Lian. (2019, May). 107-2\_dcs31\_StateSpaceDesign. [Online]. Available: http://cc.ee.ntu.edu.tw/~fengli/Teaching/DigitalControl/107-2\_dcs31\_StateSpace Design.pdf