Spring 2021

數位控制系統 Digital Control Systems

DCS-23 Controllability and Observability



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Introduction: Model and Analysis



- Plant (CT):
 - Input-Output Model:

$$\frac{Y_c(s)}{U_c(s)} = G_c(s) = \frac{B_c(s)}{A_c(s)}$$

• State-Space Model:

 $\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}u(t)$ $y(t) = \mathbf{C}\mathbf{x}(t) + \mathbf{D}u(t)$

- System Properties:
 - Stability
 - Controllability
 - Observability

Plant (DT):

• Input-Output Model:

$$\frac{Y_d(z)}{U_d(z)} = G_d(z) = \frac{B_d(z)}{A_d(z)}$$

• State-Space Model:

$$\mathbf{x}[k+1] = \mathbf{F}\mathbf{x}[k] + \mathbf{H}u[k]$$
$$y[k] = \mathbf{C}\mathbf{x}[k] + \mathbf{D}u[k]$$

Outline

- Controllability
- Controllable Canonical Form
- Observability
- Observable Canonical Form
- Kalman's Decomposition
- Loss of Controllability & Observability through Sampling

Introduction

Controllability

• Whether it is possible to steer a system

from a given initial state to another state?

 $x(k_0)$

Observability

 How to determine the state of a dynamic system from the observations of inputs and outputs



x(k)

Some Examples

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = \begin{bmatrix} 2 & 0 \\ 0 & 3 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u(k)$$
$$\begin{cases} x_1(k+1) = 2x_1(k) + u(k) \\ x_2(k+1) = 3x_2(k) + u(k) \end{cases}$$

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = \begin{bmatrix} 2 & 0 \\ 0 & 3 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(k)$$
$$\begin{cases} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = 2x_1(k) + u(k)$$
$$3x_2(k)$$
$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = \begin{bmatrix} 2 & 0 \\ 1 & 3 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(k)$$

 $\begin{cases} x_1(k+1) = 2x_1(k) + u(k) \\ x_2(k+1) = 3x_2(k) + x_1(k) \end{cases}$

- Definition: Controllability
 - The system is controllable
 - if it is possible to find a control sequence
 - such that an arbitrary state can be reached
 - From any initial state in finite time.

Consider the system: $\begin{cases} \mathbf{x}(k+1) = \mathbf{F}\mathbf{x}(k) + \mathbf{H}u(k) \\ y(k) = \mathbf{C}\mathbf{x}(k) \end{cases}$ $\mathbf{x} \in \mathbb{R}^n$ Initial state: $x_0 = x(0)$ The state at time n $\mathbf{x}(n) = \mathbf{F}^{n} \mathbf{x}(0) + \mathbf{F}^{n-1} \mathbf{H} u(0) + \dots + \mathbf{H} u(n-1)$ $= \mathbf{F}^{n} \mathbf{x}(0) + \begin{bmatrix} \mathbf{H} & \mathbf{F}\mathbf{H} & \cdots & \mathbf{F}^{n-1}\mathbf{H} \end{bmatrix} \begin{bmatrix} u(n-1) \\ u(n-2) \\ \vdots \\ u(0) \end{bmatrix}$ $= x_0^n + W_c U$ $\Rightarrow \mathbf{x}(n) - \mathbf{x_0^n} = \mathbf{W_c} \mathbf{U}$ \Rightarrow U = W_c⁻¹ [x(n) - x₀ⁿ] IF rank(W_c) = n That is, exit some control signals, such that: Initial state: $\mathbf{x}_0 = \mathbf{x}(0) \rightarrow \mathbf{x}(n)$

- Theorem: Controllability
 - The system is controllable

if and only the matrix Wc has rank n.

Controllability Matrix

$$\mathbf{W_c} = \left[\mathbf{H} \ \mathbf{F} \mathbf{H} \ \cdots \ \mathbf{F}^{n-1} \mathbf{H}
ight]$$

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Example: Controllable and Un-controllable systems

$$\mathbf{x}(k+1) = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(k)$$
$$\mathbf{W}_{\mathbf{c}} = \begin{bmatrix} \mathbf{H} & \mathbf{F}\mathbf{H} & \cdots & \mathbf{F}^{n-1}\mathbf{H} \end{bmatrix}$$
$$\mathbf{W}_{\mathbf{c}} = \begin{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$
$$\mathbf{W}_{\mathbf{c}} = \begin{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \begin{bmatrix} \mathbf{0} & \mathbf{0} \\ 1 \end{bmatrix} \begin{bmatrix} \mathbf{0} & \mathbf{0} \\ \mathbf{0} \end{bmatrix} = \mathbf{0}$$

 $\mathbf{x}(k+1) = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(k)$ $\mathbf{W}_{\mathbf{c}} = \begin{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} = \begin{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \end{bmatrix} \end{bmatrix}$ $= \begin{bmatrix} \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \end{bmatrix} \end{bmatrix}$ $= \mathbf{It is not controllable}$

Assume that F has the characteristic polynomial:

det
$$(\lambda \mathbf{I} - \mathbf{F}) = \lambda^n + a_1 \lambda^{n-1} + \dots + a_n$$

- Assume that **Wc** is nonsingular.
- Then, the system can be described by the following Controllable Canonical From:

$$\mathbf{z}(k+1) = \begin{bmatrix} -a_1 & -a_2 & \cdots & -a_{n-1} & -a_n \\ 1 & 0 & \cdots & 0 & 0 \\ 0 & 1 & \cdots & 0 & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \cdots & 1 & 0 \end{bmatrix} \mathbf{z}(k) + \begin{bmatrix} 1 \\ 0 \\ 0 \\ \vdots \\ 0 \end{bmatrix} u(k)$$
$$\mathbf{y}(k) = \begin{bmatrix} b_1 & \cdots & b_n \end{bmatrix} \mathbf{z}(k)$$

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 z_n]

- The advantage of using the Controllable Canonical From:
 IF the input is:

$$u(k) = -\begin{bmatrix} k_1 & k_2 & \cdots & k_n \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \\ \vdots \\ z_n \end{bmatrix}^{\left\lfloor 0 & 0 & \cdots & 1 & 0 \\ \end{matrix}}$$

$$\mathbf{z}(k+1) = \begin{bmatrix} -a_1 & -a_2 & \cdots & -a_{n-1} & -a_n \\ 1 & 0 & \cdots & 0 & 0 \\ 0 & 1 & \cdots & 0 & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & \cdots & 1 & 0 \end{bmatrix} \mathbf{z}(k)$$
$$\begin{bmatrix} -k_1 & -k_2 & \cdots & -k_{n-1} & -k_n \\ 0 & 0 & \cdots & 0 & 0 \\ 0 & 0 & \cdots & 0 & 0 \\ z(k) & - \begin{bmatrix} 1 \\ 0 \\ 0 \\ \vdots \end{bmatrix} \begin{bmatrix} k_1 & k_2 & \cdots & k_n \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \\ \vdots \end{bmatrix}$$

z(k+1) = $\begin{vmatrix} -(a_1 + k_1) & -(a_2 + k_2) & \cdots & -(a_{n-1} + k_{n-1}) & -(a_n + k_n) \\ 1 & 0 & \cdots & 0 & 0 \\ 0 & 1 & \cdots & 0 & 0 \end{vmatrix}$ $|\mathbf{z}(k)|$ ÷ · . \mathbf{O} The characteristic polynomial of the controlled system is: det ($\lambda \mathbf{I} - (\mathbf{F} - \mathbf{HK})$) $= \lambda^{n} + (a_{1} + k_{1})\lambda^{n-1} + \dots + (a_{n} + k_{n})$ $\mathbf{z}(k+1) = \mathbf{F}\mathbf{z}(k) + \mathbf{H}u(k) = \mathbf{F}\mathbf{z}(k) - \mathbf{H}\mathbf{K}\mathbf{z}(k)$ $u(k) = -\mathbf{K}\mathbf{z}(k)$ $= (\mathbf{F} - \mathbf{H}\mathbf{K}) \mathbf{z}(k)$

• Example:

$$\mathbf{x}(k+1) = \begin{bmatrix} -a_1 & -a_2 \\ 1 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(k)$$
$$y(k) = \begin{bmatrix} b_1 & b_2 \end{bmatrix} \mathbf{x}(k)$$

The input-output transfer function is:

$$G(z) = \mathbf{C}(z\mathbf{I} - \mathbf{F})^{-1}\mathbf{H} + \mathbf{D} = \frac{B(z)}{A(z)}$$

$$= \begin{bmatrix} b_1 & b_2 \end{bmatrix} \begin{pmatrix} z\mathbf{I} - \begin{bmatrix} -a_1 & -a_2 \\ 1 & 0 \end{bmatrix} \end{pmatrix}^{-1} \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$
$$= \begin{bmatrix} b_1 & b_2 \end{bmatrix} \begin{bmatrix} z + a_1 & a_2 \\ -1 & z \end{bmatrix}^{-1} \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$
$$= \frac{1}{z^2 + a_1 z + a_2} \begin{bmatrix} b_1 & b_2 \end{bmatrix} \begin{bmatrix} z & -a_2 \\ 1 & z + a_1 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \frac{b_1 z + b_2}{z^2 + a_1 z + a_2}$$

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Introduction

Controllability

• Whether it is possible to steer a system

from a given initial state to another state?

 $x(k_0)$

Observability

 How to determine the state of a dynamic system from the observations of inputs and outputs



x(k)

 $\mathbf{x}(k+1) = \mathbf{F}\mathbf{x}(k) + \mathbf{H}u(k)$ $y(k) = \mathbf{C}\mathbf{x}(k)$

 $\mathbf{x} \in \mathbb{R}^n$

Definition: Un-observable States

 $x_0 \neq 0$ is un-observable if \exists a finite $k_1 \geq n-1$ Initial state: $x_0 = x(0)$ such that when $x(0) = x_0 \& u(k) = 0$, for $0 \le k \le k_1$ then y(k) = 0, for $0 < k < k_1$

Definition: Observable

A system is observable if \exists a finite k such that the knowledge of $\begin{cases} u(0), ..., u(k-1) \\ y(0), ..., y(k-1) \end{cases}$

is sufficient to determine the initial state of the system



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Let
$$u(k) \equiv 0$$

and $y(0), y(1), ..., y(k-1)$ are given:
 $y(0) = \mathbf{C}\mathbf{x}(0)$
 $y(1) = \mathbf{C}\mathbf{x}(1) = \mathbf{C}(\mathbf{F}\mathbf{x}(0))$
 $y(n-1) = \mathbf{C}\mathbf{x}(n-1) = \cdots = \mathbf{C}\mathbf{F}^{n-1}\mathbf{x}(0)$
 $\begin{bmatrix} y(0)\\ y(1)\\ \vdots\\ y(n-1) \end{bmatrix} = \begin{bmatrix} \mathbf{C}\mathbf{x}(0)\\ \mathbf{C}\mathbf{F}\mathbf{x}(0)\\ \vdots\\ \mathbf{C}\mathbf{F}^{n-1}\mathbf{x}(0) \end{bmatrix} = \begin{bmatrix} \mathbf{C}\\ \mathbf{C}\\ \mathbf{C}\\ \mathbf{F}\\ \vdots\\ \mathbf{C}\mathbf{F}^{n-1}\mathbf{x}(0) \end{bmatrix}$
IF rank($\mathbf{W}_{\mathbf{0}}$) = $n \Rightarrow \mathbf{x}(0) = \mathbf{W}_{\mathbf{0}}^{-1}\mathbf{Y}$

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Theorem: Observability

The system is observable $\iff \operatorname{rank}(\mathbf{W}_0) = n$

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Example: <u>A system with unobservable states</u>

$$\mathbf{x}(k+1) = \begin{bmatrix} 1.1 & -0.3 \\ 1 & 0 \end{bmatrix} \mathbf{x}(k)$$
$$y(k) = \begin{bmatrix} 1 & -0.5 \end{bmatrix} \mathbf{x}(k)$$

The observability matrix is:

$$\mathbf{W}_{\mathbf{0}} = \begin{bmatrix} \mathbf{C} \\ \mathbf{C} \mathbf{F} \end{bmatrix} = \begin{bmatrix} 1 & -0.5 \end{bmatrix} \begin{bmatrix} 1.1 & -0.3 \\ 1 & 0 \end{bmatrix} \end{bmatrix}$$
$$= \begin{bmatrix} 1 & -0.5 \\ 0.6 & -0.3 \end{bmatrix} \Rightarrow \operatorname{rank}(\mathbf{W}_{\mathbf{0}}) = 1$$

The unobservable states belong to the null space of Wo:

that is, $\begin{vmatrix} 0.5 \\ 1 \end{vmatrix}$

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Example 3.10: <u>A system with unobservable states</u>



Observable Canonical Form

Observable Canonical Form

F:
$$\det(\lambda I - \mathbf{F}) = \lambda^n + a_1 \lambda^{n-1} + \dots + a_n = 0$$

W₀: nonsingular

T:
$$x \rightarrow z$$
, i.e., $z = Tx$

The transformed system is:

$$\mathbf{z}(k+1) = \begin{bmatrix} -a_1 & 1 & 0 & \cdots & 0 \\ -a_2 & 0 & 1 & & 0 \\ \vdots & \vdots & & \ddots & 0 \\ -a_{n-1} & \vdots & & & 1 \\ -a_n & 0 & & \cdots & 0 \end{bmatrix} \mathbf{z}(k) + \begin{bmatrix} b_1 \\ b_2 \\ \vdots \\ b_{n-1} \\ b_n \end{bmatrix} u(k)$$

 $y(k) = \begin{bmatrix} 1 & 0 & \cdots & 0 \end{bmatrix} \mathbf{z}(k)$

Observable Canonical Form

Observable Canonical Form Easy to find the observer gain: $\mathbf{x}(k+1) = \mathbf{F}\mathbf{x}(k) + \mathbf{H}u(k)$ $y(k) = \mathbf{C}\mathbf{x}(k)$ $\mathbf{x}_{o}(k+1) = \mathbf{F}\mathbf{x}_{o}(k) + \mathbf{H}u(k) + u_{o}(k)$ $u_{o}(k) = + L[y(k) - C x_{o}(k)]$ $x_o(k+1) = Fx_o(k) + Hu(k) + L[y(k) - Cx_o(k)]$ $x(k+1) - x_o(k+1) = F[x(k) - x_o(k)]$ $-\mathbf{L}[\mathbf{C}\mathbf{x}(k) - \mathbf{C}\mathbf{x}_{o}(k)]$ $\mathbf{x}_{e}(k) = \mathbf{x}(k) - \mathbf{x}_{o}(k)$ $-\mathbf{L} \mathbf{C} [\mathbf{x}(k) - \mathbf{x}_o(k)]$ $\mathbf{x}_e(k+1) = \mathbf{F} \mathbf{x}_e(k) - \mathbf{L} \mathbf{C} \mathbf{x}_e(k)$ $= [\mathbf{F} - \mathbf{L}\mathbf{C}]\mathbf{x}_{e}(k)$

- Observable Canonical Form
- Easy to find the observer gain:

 $\mathbf{x}_e(k+1) = [\mathbf{F} - \mathbf{L}\mathbf{C}] \mathbf{x}_e(k)$

$$\mathbf{LC} = \begin{bmatrix} l_1 \\ l_2 \\ \vdots \\ l_n \end{bmatrix} \begin{bmatrix} 1 \ 0 \ \cdots \ 0 \end{bmatrix} = \begin{bmatrix} l_1 \ 0 \ \cdots \ 0 \\ l_2 \ 0 \ \cdots \ 0 \\ \vdots \ \ddots \\ l_n \ 0 \ \cdots \ 0 \end{bmatrix}$$
$$\mathbf{F} = \begin{bmatrix} -a_1 \ 1 \ 0 \ \cdots \ 0 \\ -a_2 \ 0 \ 1 \ 0 \\ \vdots \ \vdots \ \cdots \ 0 \\ -a_n \ 0 \ \cdots \ 0 \end{bmatrix}$$

Observable Canonical Form

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 $\mathbf{x}_e(k+1) = [\mathbf{F} - \mathbf{L}\mathbf{C}]\mathbf{x}_e(k)$ $\mathbf{LC} = \begin{bmatrix} l_1 \\ l_2 \\ \vdots \\ l_n \end{bmatrix} \begin{bmatrix} 1 \ 0 \ \cdots \ 0 \end{bmatrix} = \begin{bmatrix} l_1 \ 0 \ \cdots \ 0 \\ l_2 \ 0 \ \cdots \ 0 \\ \vdots \ \ddots \\ l_n \ 0 \ \cdots \ 0 \end{bmatrix}$ $\mathbf{F} = \begin{bmatrix} -a_1 & 1 & 0 & \cdots & 0 \\ -a_2 & 0 & 1 & & 0 \\ \vdots & \vdots & \ddots & 0 \\ -a_{n-1} & \vdots & & & 1 \\ -a_n & 0 & \cdots & 0 \end{bmatrix}$ $\mathbf{F} - \mathbf{L} \mathbf{C} = \begin{bmatrix} -(a_1 + l_1) & 1 & 0 & \cdots & 0 \\ -(a_2 + l_2) & 0 & 1 & & 0 \\ \vdots & \vdots & \ddots & 0 \\ -(a_{n-1} + l_{n-1}) & \vdots & & & 1 \\ -(a_n + l_n) & 0 & \cdots & 0 \end{bmatrix}$

 $det(\lambda I - (\mathbf{F} - \mathbf{L} \mathbf{C})) = \lambda^n + (a_1 + l_1)\lambda^{n-1} + \dots + (a_n + l_n) = 0$

Example: (in observable canonical form)

$$\mathbf{x}(k+1) = \begin{bmatrix} -a_1 & 1\\ -a_2 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} b_1\\ b_2 \end{bmatrix} u(k)$$
$$y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}(k)$$

The input-output transfer function is:

$$G(z) = C(zI - F)^{-1}H + D = \frac{B(z)}{A(z)}$$

= $\begin{bmatrix} 1 & 0 \end{bmatrix} \left(zI - \begin{bmatrix} -a_1 & 1 \\ -a_2 & 0 \end{bmatrix} \right)^{-1} \begin{bmatrix} b_1 \\ b_2 \end{bmatrix}$
= $\begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} z + a_1 & -1 \\ a_2 & z \end{bmatrix}^{-1} \begin{bmatrix} b_1 \\ b_2 \end{bmatrix} = \frac{b_1 z + b_2}{z^2 + a_1 z + a_2}$
= $\frac{1}{z^2 + a_1 z + a_2} \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} z & 1 \\ -a_2 & z + a_1 \end{bmatrix} \begin{bmatrix} b_1 \\ b_2 \end{bmatrix}$

 $G(z) = \frac{b_1 z + b_2}{z^2 + a_1 z + a_2}$

 $G(z) = \frac{b_1 z + b_2}{z^2 + a_1 z + a_2}$

Observable canonical form:

$$\mathbf{x}(k+1) = \begin{bmatrix} -a_1 & 1\\ -a_2 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} b_1\\ b_2 \end{bmatrix} u(k)$$
$$y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}(k)$$

The input-output transfer function is:

Controllable canonical form:

$$\mathbf{x}(k+1) = \begin{bmatrix} -a_1 & -a_2 \\ 1 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(k)$$
$$y(k) = \begin{bmatrix} b_1 & b_2 \end{bmatrix} \mathbf{x}(k)$$

The input-output transfer function is:

Kalman showed that:

$$\mathbf{x}(k+1) = \begin{bmatrix} F_{11} & F_{12} & 0 & 0 \\ 0 & F_{22} & 0 & 0 \\ F_{31} & F_{32} & F_{33} & F_{34} \\ 0 & F_{42} & 0 & F_{44} \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} H_1 \\ 0 \\ H_3 \\ 0 \end{bmatrix} u(k)$$
$$y(k) = \begin{bmatrix} C_1 & C_2 & 0 & 0 \end{bmatrix} \mathbf{x}(k)$$
where $\mathbf{x}(k) = \begin{bmatrix} OC \\ O\bar{C} \\ \bar{O}\bar{C} \\ \bar{O}\bar{C} \end{bmatrix}$ and, $\mathbf{G}(z) = \mathbf{C} (zI - \mathbf{F})^{-1} \mathbf{H}$
$$= C_1 (zI - F_{11})^{-1} H_1$$

 $\mathbf{x}_{1}[k+1] = F_{11}\mathbf{x}_{1}[k] + H_{1}u[k]$

 $y[k] = C_1 \mathbf{x}_1[k]$

Kalman's Decomposition

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C.T. system D.T. system

If D.T. system is controllable, then C.T. system is controllable
 But, if C.T. system is controllable, D.T. system may not!

Loss of Observability:

Un-observability in C.T. system:

✓ zero over a time interval

>Un-observability in D.T. system:

✓ zero only at sampling instants

 May oscillate between sampling instants (hidden oscillation)

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- Example 3.12: The harmonic oscillator
- The CT model is:

$$\frac{dx}{dt} = \begin{bmatrix} 0 & w \\ -w & 0 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ w \end{bmatrix} u$$
$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}$$

The DT model (using zero-order hold) is:

$$\mathbf{x}(k+1) = \begin{bmatrix} \cos(wh) & \sin(wh) \\ -\sin(wh) & \cos(wh) \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 1 - \cos(wh) \\ \sin(wh) \end{bmatrix} u(k)$$
$$y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}(k)$$

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- Example 3.12: The harmonic oscillator
- The CT model is:

$$\frac{dx}{dt} = \begin{bmatrix} 0 & w \\ -w & 0 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ w \end{bmatrix} u$$
$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}$$

The controllability matrix is:

$$\mathbf{W}_{\mathbf{C}}^{\mathbf{C}} = \begin{bmatrix} \mathbf{B} & \mathbf{A}\mathbf{B} \end{bmatrix} = \begin{bmatrix} \begin{bmatrix} 0 \\ w \end{bmatrix} \begin{bmatrix} 0 & w \\ -w & 0 \end{bmatrix} \begin{bmatrix} 0 \\ w \end{bmatrix} = \begin{bmatrix} 0 & w \\ w^2 & 0 \end{bmatrix}$$

$$\bullet \text{ The observability matrix is:} \qquad \Rightarrow \det(\mathbf{W}_{\mathbf{C}}^{\mathbf{C}}) = -w^3$$

$$\mathbf{W}_{\mathbf{0}}^{\mathbf{C}} = \begin{bmatrix} \mathbf{C} \\ \mathbf{C}\mathbf{A} \end{bmatrix} = \begin{bmatrix} \begin{bmatrix} 1 & 0 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} 0 & w \\ -w & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & w \end{bmatrix}$$

$$\Rightarrow \det(\mathbf{W}_{\mathbf{0}}^{\mathbf{C}}) = w$$

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- Example 3.12: The harmonic oscillator
- The DT model (using zero-order hold) is:

$$\mathbf{x}(k+1) = \begin{bmatrix} \cos(wh) & \sin(wh) \\ -\sin(wh) & \cos(wh) \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 1 - \cos(wh) \\ \sin(wh) \end{bmatrix} u(k)$$

- $y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}(k)$
- The controllability matrix is:

det(

 $W_c^d = \begin{bmatrix} H & FH \end{bmatrix}$

$$= \begin{bmatrix} 1 - cwh & cwh(1 - cwh) + (swh)^2 \\ swh & -swh(1 - cwh) + (cwh)(swh) \end{bmatrix}$$
$$= \begin{bmatrix} 1 - cwh & cwh - (cwh)^2 + (swh)^2 \\ swh & -swh + 2(cwh)(swh) \end{bmatrix}$$
$$\mathbf{W_c^d} \quad = \cdots = -2(sinwh)(1 - (coswh))$$

Loss of Controllability & Observability through Sampling Feng-Li Lian © 2021 DCS23-CtrlObsv-32

- Example 3.12: The harmonic oscillator
- The DT model (using zero-order hold) is:

$$\mathbf{x}(k+1) = \begin{bmatrix} \cos(wh) & \sin(wh) \\ -\sin(wh) & \cos(wh) \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 1 - \cos(wh) \\ \sin(wh) \end{bmatrix} u(k)$$

- $y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}(k)$
- The observability matrix is:

$$\mathbf{W_o^d} = \begin{bmatrix} \mathbf{C} \\ \mathbf{CF} \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} \cos(wh) & \sin(wh) \\ -\sin(wh) & \cos(wh) \end{bmatrix}$$
$$= \begin{bmatrix} 1 & 0 \\ \cos(wh) & \sin(wh) \end{bmatrix}$$

 $det(\mathbf{W_o^d}) = \sin wh$

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Example 3.12: D.T. model of the harmonic oscillator $det(\mathbf{W_c^d}) = -2(\sin wh)(1 - \cos wh)$ $\Rightarrow \sin wh = 0 \Rightarrow wh = 0, n\pi$ $\Rightarrow 1 - \cos wh = 0 \Rightarrow \cos wh = 1 \Rightarrow wh = 2n\pi$ det($\mathbf{W}_{\mathbf{0}}^{\mathbf{d}}$) = sin wh $\Rightarrow \sin wh = 0 \Rightarrow wh = 0, n\pi \Rightarrow w = \frac{\pi}{h} \Rightarrow \frac{2\pi}{h}$ $\Rightarrow w_N = \frac{\pi}{h}$ SUMMARY

Models	Controllability	Observability
СТ	ОК	ОК
DT	Lost when $wh = n\pi$	Lost when $wh = n\pi$