

Spring 2019

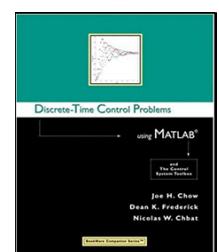
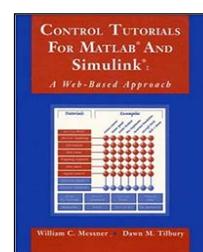
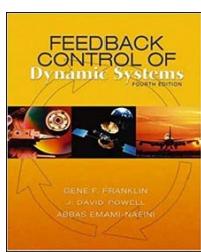
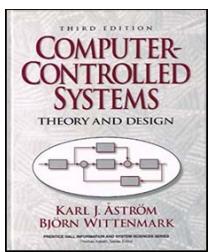
數位控制系統
Digital Control Systems

DCS-11
Discrete-Time Systems –
State Space Model

Feng-Li Lian

NTU-EE

Feb19 – Jun19



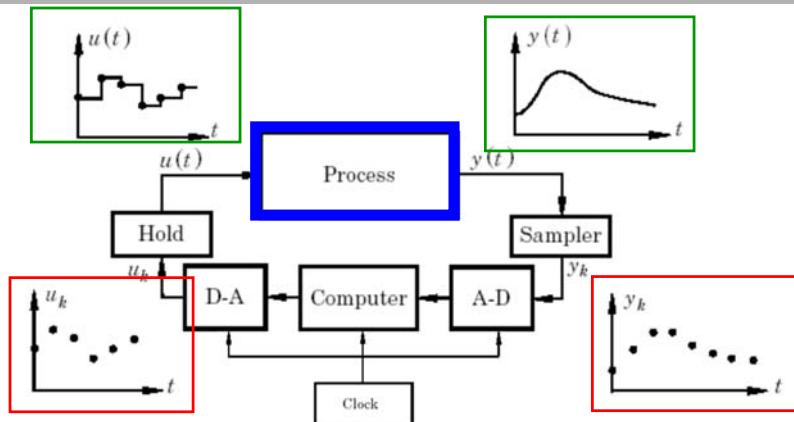
Review:

Feng-Li Lian © 2019
DCS11-SSModel-2

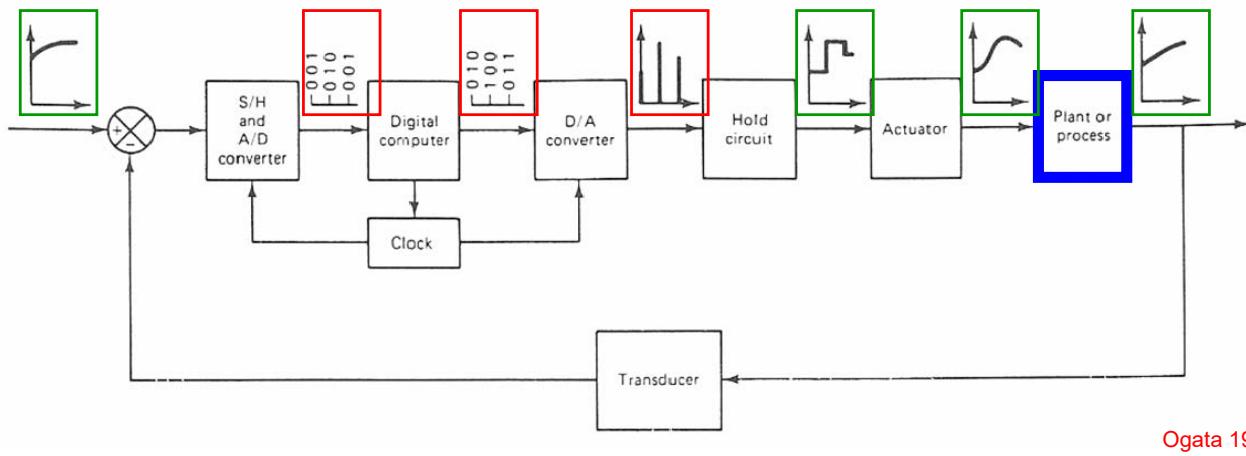
- **System Modeling at Control Tutorials for Matlab & Simulink:**
 - <http://ctms.engin.umich.edu/CTMS/index.php?example=Introduction§ion=SystemModeling>
- **System Models in LS:**
 - <http://cc.ee.ntu.edu.tw/~fengli/Teaching/LinearSystems/971lsNotePhoto080924.pdf>
- **Solving 1st-order Differential Equations:**
 - https://case.ntu.edu.tw/CASTUDIO/Files/speech/Ref/CS0101S1B02_03.pdf
- **Systems of Linear 1st-order Differential Equations:**
 - http://case.ntu.edu.tw/CASTUDIO/Files/speech/Ref/CS0101S1B02_16.pdf
- **Solving State-Space Equations:**
 - <http://cc.ee.ntu.edu.tw/~fengli/Teaching/LinearSystems/971lsNotePhoto081001.pdf>

Introduction: Digital Signals and Systems

Feng-Li Lian © 2019
DCS11-SSModel-3



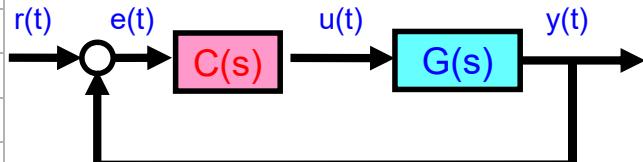
Astrom & Wittenmark 1997



Ogata 1995

Introduction: CT and DT Plant-Controller

Feng-Li Lian © 2019
DCS11-SSModel-4



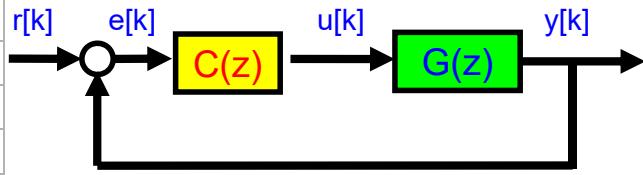
$G(s) \rightarrow C(s)$

- Transform CT plant into DT plant
- By DT plant, design DT controller

$G(s) \rightarrow C(s)$

- By CT plant, design CT controller
- Transform CT controller into DT controller

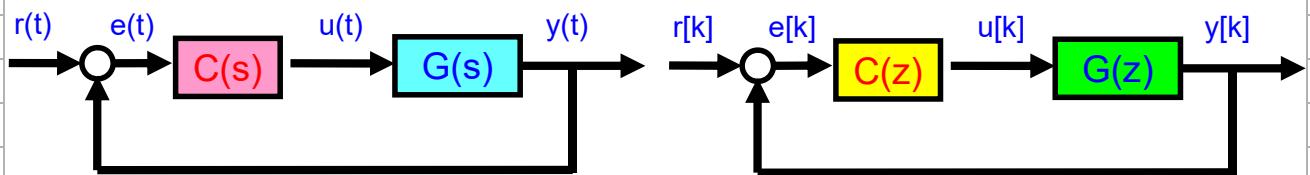
$G(z) \rightarrow C(z)$



$G(z) \rightarrow C(z)$

Introduction: From CT Plant to DT Plant

Feng-Li Lian © 2019
DCS11-SSModel-5

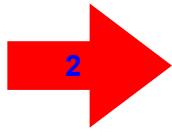


▪ Plant (CT):

- Input-Output Model:

$$u(t)$$

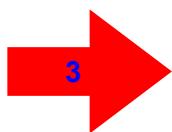
$$y(t)$$



$$u[k]$$

$$y[k]$$

$$G(s) = \frac{Y(s)}{U(s)}$$

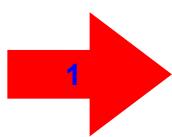


$$G(z) = \frac{Y(z)}{U(z)}$$

- State-Space Model:

$$\dot{x}(t) = Ax(t) + Bu(t)$$

$$y(t) = Cx(t) + Du(t)$$



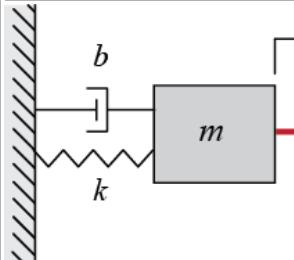
- State-Space Model:

$$x[k+1] = Fx[k] + Hu[k]$$

$$y[k] = Cx[k] + Du[k]$$

Introduction: Example - Mass-Spring-Damper System

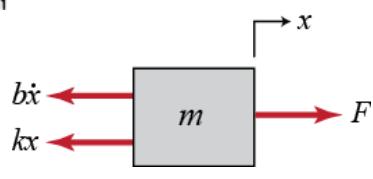
Feng-Li Lian © 2019
DCS11-SSModel-6



$$F(t) - b\dot{x}(t) - kx(t) = m\ddot{x}(t)$$

$$\Rightarrow F(s) - bsX(s) - kX(s) = ms^2X(s)$$

$$\Rightarrow \frac{X(s)}{F(s)} = \frac{1}{ms^2 + bs + k} = \frac{Y(s)}{U(s)} = G(s)$$



$$y = x = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \end{bmatrix} u$$

$$x = \begin{bmatrix} x \\ \dot{x} \end{bmatrix}$$

$$\dot{x} = \begin{bmatrix} \dot{x} \\ \ddot{x} \end{bmatrix} = \begin{bmatrix} -\frac{k}{m}x - \frac{b}{m}\dot{x} + \frac{1}{m}F(t) \end{bmatrix}$$

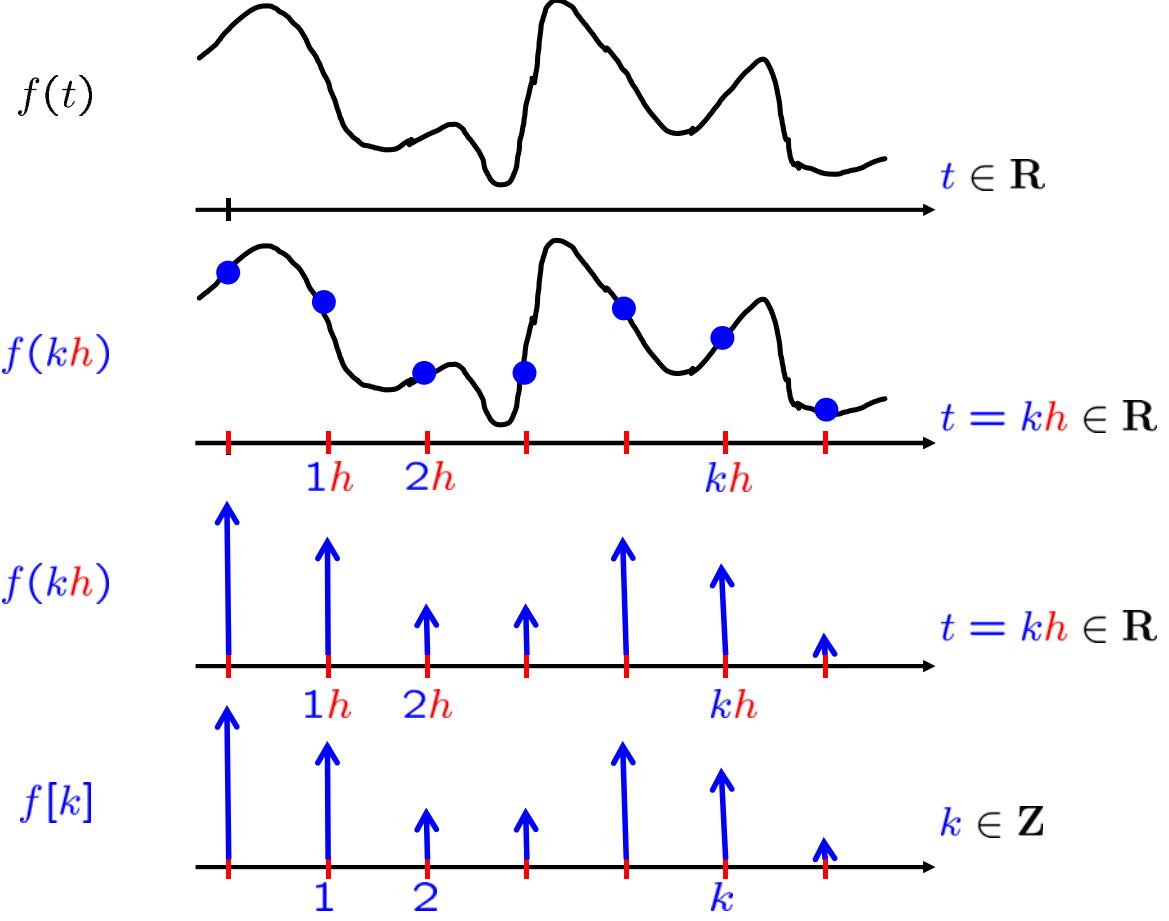
$$= \begin{bmatrix} 0 & 1 \\ -\frac{k}{m} & -\frac{b}{m} \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix} F(t)$$

$$= \begin{bmatrix} 0 & 1 \\ -\frac{k}{m} & -\frac{b}{m} \end{bmatrix} \begin{bmatrix} x \\ \dot{x} \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix} u$$

CTMS

Sampling a CT Signal

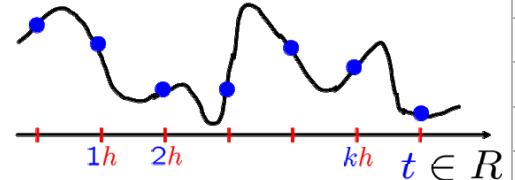
Feng-Li Lian © 2019
DCS11-SSModel-7



Sampling a CT Signal

Feng-Li Lian © 2019
DCS11-SSModel-8

- $\mathbf{Z} = \{\dots, -1, 0, 1, 2, \dots\}$
- Sampling Period: h (sec)
- Sampling Rate: $f_s = \frac{1}{h}$ (Hz)
- Sampling Frequency: $w_s = \frac{2\pi}{h}$ (rad/s)
- Sampling Instants: t_k



$$\mathbf{T} = \{t_k = kh, k \in \mathbf{Z}\}$$

- CT Signals: $f(t) : f(t) \in \mathbf{R}, t \in \mathbf{R}$
- DT Signals: $f(t_k) : f(t_k) \in \mathbf{R}, t_k \in \mathbf{T}$

$$f(t_k) \in \mathbf{D}$$
- DT Signals: $f[k] : f[k] \in \mathbf{R}, k \in \mathbf{Z}$

$$f[k] \in \mathbf{D}$$

- Consider the following LTI system:

$$\begin{aligned}\dot{\mathbf{x}}(t) &= \mathbf{A}\mathbf{x}(t) + \mathbf{B}u(t) \\ \mathbf{y}(t) &= \mathbf{C}\mathbf{x}(t) + \mathbf{D}u(t)\end{aligned}$$

$$\dot{\mathbf{x}}(t) = \frac{d}{dt}\mathbf{x}(t)$$

$$\begin{array}{ll}\mathbf{x}(t) \in \mathbb{R}^n & \mathbf{A} \in \mathbb{R}^{n \times n} \\ \mathbf{u}(t) \in \mathbb{R}^r & \mathbf{B} \in \mathbb{R}^{n \times r} \\ \mathbf{y}(t) \in \mathbb{R}^p & \mathbf{C} \in \mathbb{R}^{p \times n} \\ & \mathbf{D} \in \mathbb{R}^{p \times r}\end{array}$$

For SISO system,
 $r = 1, p = 1$

$$\frac{d}{dt}\mathbf{x}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}u(t)$$

See: Solution of 1st-order DE

$$\Rightarrow \mathbf{x}(t) = e^{\mathbf{A}(t-t_0)}\mathbf{x}(t_0) + \int_{t_0}^t e^{\mathbf{A}(t-\tau)}\mathbf{B}u(\tau)d\tau$$

$$\text{Let } t = t_{k+1} = kh + h \quad \& \quad t_0 = t_k = kh$$

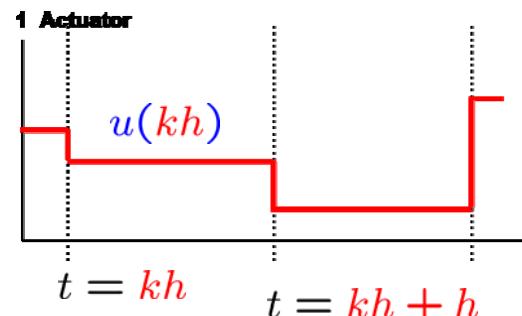
$$\Rightarrow \mathbf{x}(kh+h) = e^{\mathbf{A}(h)}\mathbf{x}(kh) + \int_{kh}^{kh+h} e^{\mathbf{A}(kh+h-\tau)}\mathbf{B}u(\tau)d\tau$$

Sampling a CT State-Space System: Piecewise Constant Input

Let $u(\tau)$ be piecewise constant through h

$$u(\tau) = u(kh), \quad kh \leq \tau < kh + h$$

$$\text{Let } \eta = kh + h - \tau$$



$$\Rightarrow \mathbf{x}(kh+h) = e^{\mathbf{A}h}\mathbf{x}(kh) + \left(\int_0^h e^{\mathbf{A}\eta} d\eta \right) \mathbf{B}u(kh)$$

$$\text{Let } \mathbf{F} = e^{\mathbf{A}h} \quad \& \quad \mathbf{H} = \left(\int_0^h e^{\mathbf{A}\eta} d\eta \right) \mathbf{B}$$

$$\Rightarrow \mathbf{x}((k+1)h) = \mathbf{F}\mathbf{x}(kh) + \mathbf{H}u(kh)$$

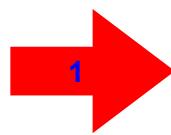
Then,

$$\begin{aligned}\mathbf{x}[k+1] &= \mathbf{F}\mathbf{x}[k] + \mathbf{H}u[k] \\ y[k] &= \mathbf{C}\mathbf{x}[k] + \mathbf{D}u[k]\end{aligned}$$

Sampling a CT State-Space System: CT (A, B) to DT (F, H)



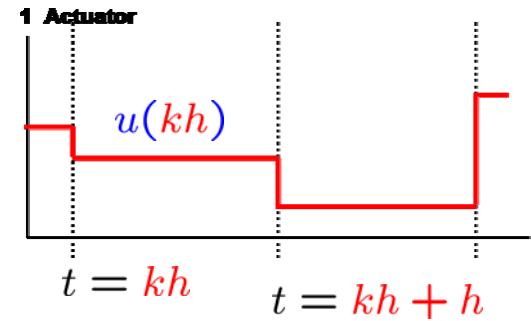
$$\begin{aligned}\dot{x}(t) &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t)\end{aligned}$$



$$\begin{aligned}x[k+1] &= Fx[k] + Hu[k] \\ y[k] &= Cx[k] + Du[k]\end{aligned}$$

- No approximation
- Control signal = Piecewise constant

$$F = e^{Ah} \quad \& \quad H = \left(\int_0^h e^{A\eta} d\eta \right) B$$



$$\Rightarrow F(h) = e^{Ah} \quad \& \quad H(h) = \left(\int_0^h e^{A\eta} d\eta \right) B$$

Sampling a CT State-Space System: CT (A, B) to DT (F, H)

$$\Rightarrow F(h) = e^{Ah} \quad \& \quad H(h) = \left(\int_0^h e^{A\eta} d\eta \right) B$$

$$\begin{aligned}\Rightarrow \frac{d}{dh} F(h) &= \frac{d}{dh} (e^{Ah}) = A(e^{Ah}) = A F(h) \\ &= (e^{Ah}) A = F(h) A\end{aligned}$$

$$\Rightarrow \frac{d}{dh} H(h) = \frac{d}{dh} \left(\int_0^h e^{A\eta} d\eta \right) B \quad MN \neq NM$$

$$= (e^{Ah}) B = F(h) B \neq B F(h)$$

$$\Rightarrow \frac{d}{dh} \begin{bmatrix} F(h) & H(h) \\ 0 & I \end{bmatrix} = \begin{bmatrix} F(h) & H(h) \\ 0 & I \end{bmatrix} \begin{bmatrix} A & B \\ 0 & 0 \end{bmatrix}$$

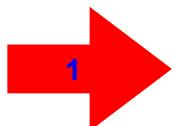
$$\Rightarrow \begin{bmatrix} F(h) & H(h) \\ 0 & I \end{bmatrix} = \exp \left(\begin{bmatrix} A & B \\ 0 & 0 \end{bmatrix} h \right)$$

$$\Rightarrow \begin{bmatrix} A & B \\ 0 & 0 \end{bmatrix} = \frac{1}{h} \ln \begin{bmatrix} F(h) & H(h) \\ 0 & I \end{bmatrix}$$

Sampling a CT State-Space System: CT (A, B) to DT (F, H)

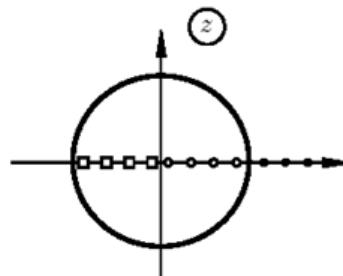
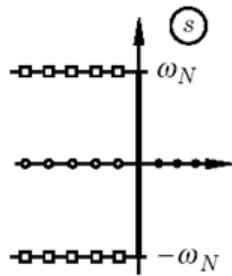
Feng-Li Lian © 2019
DCS11-SSModel-13

$$\begin{aligned}\dot{\mathbf{x}}(t) &= \mathbf{A}\mathbf{x}(t) + \mathbf{B}u(t) \\ y(t) &= \mathbf{C}\mathbf{x}(t) + \mathbf{D}u(t)\end{aligned}$$



$$\begin{aligned}\mathbf{x}[k+1] &= \mathbf{F}\mathbf{x}[k] + \mathbf{H}u[k] \\ y[k] &= \mathbf{C}\mathbf{x}[k] + \mathbf{D}u[k]\end{aligned}$$

$$\Rightarrow \mathbf{F}(h) = e^{\mathbf{A}h} \quad \& \quad \mathbf{H}(h) = \left(\int_0^h e^{\mathbf{A}\eta} d\eta \right) \mathbf{B}$$



Sampling a CT State-Space System: Example

Feng-Li Lian © 2019
DCS11-SSModel-14

- Consider the following SS Model (Double Integrator):

$$\begin{aligned}\dot{\mathbf{x}} &= \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u & = \mathbf{Ax} + \mathbf{Bu} \\ y &= [1 \ 0] \mathbf{x} & = \mathbf{Cx}\end{aligned}$$

$$\mathbf{F} = e^{\mathbf{A}h} = \mathbf{I} + \mathbf{Ah} + \frac{\mathbf{A}^2 h^2}{2!} + \dots$$

$$= \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} + \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} h + \frac{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}^2 h^2}{2!} + \dots$$

$$= \begin{bmatrix} 1 & h \\ 0 & 1 \end{bmatrix}$$

- Consider the following SS Model (Double Integrator):

$$\begin{aligned}\dot{\mathbf{x}} &= \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u & = \mathbf{Ax} + \mathbf{Bu} \\ y &= [1 \ 0] \mathbf{x} & = \mathbf{Cx}\end{aligned}$$

$$\mathbf{H} = \left(\int_0^h e^{\mathbf{A}\eta} d\eta \right) \mathbf{B} \quad e^{\mathbf{A}\tau} = \mathbf{I} + \mathbf{A}\tau + \frac{\mathbf{A}^2\tau^2}{2!} + \dots$$

$$\sum_{k=0}^{\infty} \frac{\mathbf{A}^k h^{k+1}}{(k+1)!}$$

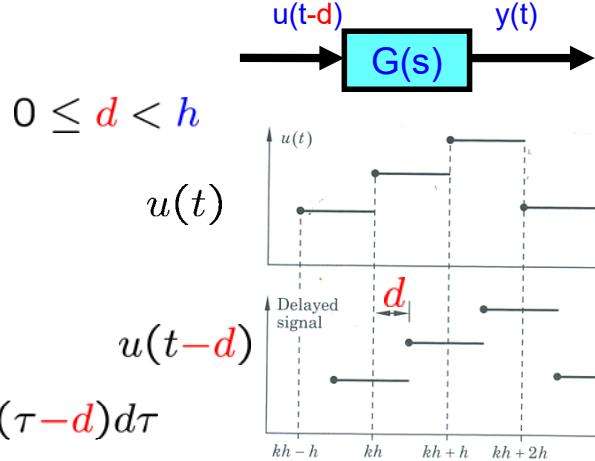
$$\begin{aligned}\mathbf{H} &= \sum_{k=0}^{\infty} \frac{\mathbf{A}^k h^{k+1}}{(k+1)!} \mathbf{B} = \mathbf{I}h + \mathbf{A}\frac{h^2}{2!}\mathbf{B} + \mathbf{A}^2\frac{h^3}{3!}\mathbf{B} + \dots \\ &= \left(\begin{bmatrix} h & 0 \\ 0 & h \end{bmatrix} + \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \frac{h^2}{2} \right) \begin{bmatrix} 0 \\ 1 \end{bmatrix} = \begin{bmatrix} h^2/2 \\ h \end{bmatrix}\end{aligned}$$

▪ Homework 2-2

Sampling a CT State-Space System with Time Delay

- Consider the LTI system with delayed input:

$$\begin{aligned}\dot{\mathbf{x}}(t) &= \mathbf{Ax}(t) + \mathbf{Bu}(t-d) \\ y(t) &= \mathbf{Cx}(t) + \mathbf{Du}(t-d) \\ \Rightarrow \mathbf{x}(t) &= e^{\mathbf{A}(t-t_0)} \mathbf{x}(t_0) + \int_{t_0}^t e^{\mathbf{A}(t-\tau)} \mathbf{B} u(\tau-d) d\tau\end{aligned}$$



$$\text{Let } t = t_{k+1} = kh + h \quad \& \quad t_0 = t_k = kh$$

$$\begin{aligned}\Rightarrow \mathbf{x}(kh+h) &= e^{\mathbf{A}(h)} \mathbf{x}(kh) + \int_{kh}^{kh+h} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} u(\tau-d) d\tau \\ &\quad \int_{kh}^{kh+h} \Rightarrow \int_{kh}^{kh+d} + \int_{kh+d}^{kh+h} \\ &+ \int_{kh}^{kh+d} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} u(\tau-d) d\tau + \int_{kh+d}^{kh+h} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} u(\tau-d) d\tau\end{aligned}$$

Sampling a CT State-Space System with Time Delay

Feng-Li Lian © 2019
DCS11-SSModel-17

$$\Rightarrow \mathbf{x}(kh + h)$$

$$= e^{\mathbf{A}(h)} \mathbf{x}(kh)$$

$$+ \int_{kh}^{kh+d} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} u(\tau-d) d\tau$$

$$u((k-1)h) = u[k-1]$$

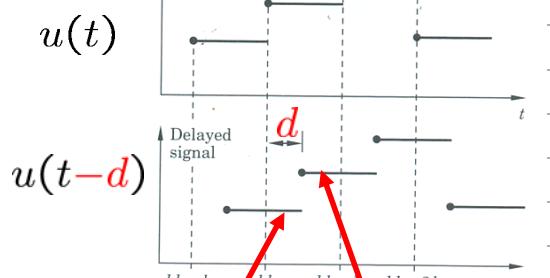
$$+ \int_{kh+d}^{kh+h} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} u(\tau-d) d\tau$$

$$u(kh) = u[k]$$

$$= \mathbf{F} \mathbf{x}(kh) + \mathbf{H}_1 u((k-1)h) + \mathbf{H}_0 u(kh)$$

$$\mathbf{H}_1 = \int_{kh}^{kh+d} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} d\tau = e^{\mathbf{A}(h-d)} \int_0^d e^{(\mathbf{A}\eta)} \mathbf{B} d\eta$$

$$\mathbf{H}_0 = \int_{kh+d}^{kh+h} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} d\tau = \int_0^{h-d} e^{(\mathbf{A}\eta)} \mathbf{B} d\eta$$



$u[k-1] \quad u[k]$

Sampling a CT State-Space System with Time Delay

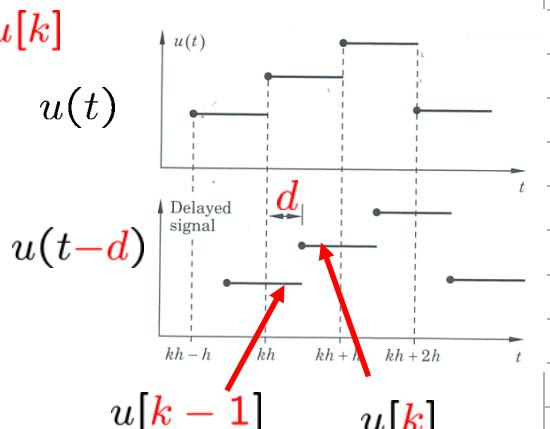
Feng-Li Lian © 2019
DCS11-SSModel-18

$$\Rightarrow \mathbf{x}[k+1]$$

$$= \mathbf{F} \mathbf{x}[k] + \mathbf{H}_1 u[k-1] + \mathbf{H}_0 u[k]$$

$$\mathbf{H}_1 = e^{\mathbf{A}(h-d)} \int_0^d e^{(\mathbf{A}\eta)} \mathbf{B} d\eta$$

$$\mathbf{H}_0 = \int_0^{h-d} e^{(\mathbf{A}\eta)} \mathbf{B} d\eta$$



$u[k-1] \quad u[k]$

$$\begin{bmatrix} \mathbf{x}[k+1] \\ \mathbf{u}[k] \end{bmatrix} = \begin{bmatrix} \mathbf{F} & \mathbf{H}_1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \mathbf{x}[k] \\ \mathbf{u}[k-1] \end{bmatrix} + \begin{bmatrix} \mathbf{H}_0 \\ \mathbf{I} \end{bmatrix} \mathbf{u}[k]$$

▪ Homework 2-4

- Consider the LTI system with longer delayed input:



If $d \geq h$

$$\text{Then, } d = (m-1)h + \eta$$

$$0 \leq \eta < h$$

$$m \in N$$

$$\Rightarrow \mathbf{x}(kh+h) = \mathbf{F}\mathbf{x}(kh)$$

$$+ \mathbf{H}_1 u(kh - (m-1)h) + \mathbf{H}_0 u(kh - mh)$$

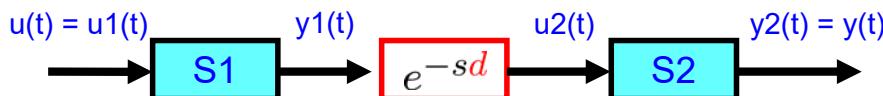
$$\mathbf{H}_1 = \int_{kh+(m-1)h}^{kh+(m-1)h+d} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} d\tau$$

$$\mathbf{H}_0 = \int_{kh+(m-1)h+d}^{kh+(m-1)h+h} e^{\mathbf{A}(kh+h-\tau)} \mathbf{B} d\tau$$

$$\begin{bmatrix} \mathbf{x}[k+1] \\ \mathbf{u}[k-(m-1)] \\ \mathbf{u}[k-m] \\ \vdots \\ \mathbf{u}[k] \end{bmatrix} = \begin{bmatrix} \mathbf{F} & \mathbf{H}_1 & \mathbf{H}_0 & 0 & \cdots & 0 \\ 0 & 0 & \mathbf{I} & 0 & \cdots & 0 \\ 0 & 0 & 0 & \ddots & \ddots & \vdots \\ 0 & 0 & \cdots & 0 & \cdots & 0 \end{bmatrix} \begin{bmatrix} \mathbf{x}[k] \\ \mathbf{u}[k-m] \\ \mathbf{u}[k-m-1] \\ \vdots \\ \mathbf{u}[k-1] \end{bmatrix} + \begin{bmatrix} 0 \\ \vdots \\ 0 \\ \mathbf{I} \end{bmatrix} \mathbf{u}[k]$$

Sampling a CT State-Space System with Inner Time Delay

- Consider the LTI system with inner time delay:



$$\mathbf{S1}: \dot{\mathbf{x}}_1(t) = \mathbf{A}_1 \mathbf{x}_1(t) + \mathbf{B}_1 u_1(t)$$

$$y_1(t) = \mathbf{C}_1 \mathbf{x}_1(t) + \mathbf{D}_1 u_1(t)$$

$$\mathbf{S2}: \dot{\mathbf{x}}_2(t) = \mathbf{A}_2 \mathbf{x}_2(t) + \mathbf{B}_2 u_2(t)$$

$$y_2(t) = \mathbf{C}_2 \mathbf{x}_2(t) + \mathbf{D}_2 u_2(t) \quad u_2(t) = y_1(t - d)$$

- By using sampling interval, h , and $0 < d \leq h$

- Then, the sampled-data representation is as follows:

$$\mathbf{x}_1(kh+h) = \mathbf{F}_1(h) \mathbf{x}_1(kh) + \mathbf{H}_1(h) \mathbf{u}(kh)$$

$$\mathbf{x}_2(kh+h) = \mathbf{F}_{21} \mathbf{x}_1(kh-h) + \mathbf{F}_2(h) \mathbf{x}_2(kh)$$

$$+ \mathbf{H}_{21} \mathbf{u}(kh-h) + \mathbf{H}_2(h-d) \mathbf{u}(kh)$$

- Where

$$\mathbf{F}_i(\eta) = e^{\mathbf{A}_i \eta}, \quad i = 1, 2$$

$$\mathbf{F}_{21}^a(\eta) = \int_0^\eta e^{\mathbf{A}_2 s} \mathbf{B}_2 \mathbf{C}_1 e^{\mathbf{A}_1(\eta-s)} ds$$

$$\mathbf{H}_1(\eta) = \int_0^\eta e^{\mathbf{A}_1 s} \mathbf{B}_1 ds$$

$$\mathbf{H}_2(\eta) = \int_0^\eta e^{\mathbf{A}_2 s} \mathbf{B}_2 \mathbf{C}_1 \mathbf{H}_1(\tau - s) ds$$

$$\mathbf{F}_{21} = \mathbf{F}_{21}^a(h) \mathbf{F}_1(h - d)$$

$$\mathbf{H}_{21} = \mathbf{F}_{21}^a(h) \mathbf{H}_1(h - d) + \mathbf{F}_{21}^a(h - d) \mathbf{H}_1(d)$$

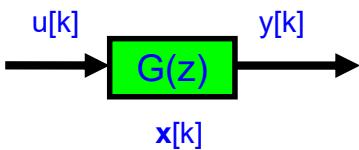
$$+ \mathbf{F}_2(h - d) \mathbf{H}_2(d)$$

- Reference:

- Bjorn Wittenmark, "Sampling of a system with a time delay,"
IEEE Transactions on Automatic Control, Vol. 30, No. 5, pp. 507-510, May 1985.
- <https://ieeexplore.ieee.org/document/1103985>

■ Homework 2-1

Solution of DT State-Space System



$$\mathbf{x}[k+1] = \mathbf{F}\mathbf{x}[k] + \mathbf{H}\mathbf{u}[k]$$

$$y[k] = \mathbf{C}\mathbf{x}[k] + \mathbf{D}\mathbf{u}[k]$$

$$\mathbf{x}[k_0 + 1] = \mathbf{F}\mathbf{x}[k_0] + \mathbf{H}\mathbf{u}[k_0]$$

$$\mathbf{x}[k_0 + 2] = \mathbf{F}\mathbf{x}[k_0 + 1] + \mathbf{H}\mathbf{u}[k_0 + 1]$$

$$= \mathbf{F}\{\mathbf{F}\mathbf{x}[k_0] + \mathbf{H}\mathbf{u}[k_0]\} + \mathbf{H}\mathbf{u}[k_0 + 1]$$

$$= \mathbf{F}^2\mathbf{x}[k_0] + \mathbf{F}\mathbf{H}\mathbf{u}[k_0] + \mathbf{H}\mathbf{u}[k_0 + 1]$$

$$\mathbf{x}[k_0 + (k - k_0)]$$

$$\mathbf{x}[k] = \mathbf{F}^{k-k_0}\mathbf{x}[k_0] + \mathbf{F}^{k-k_0-1}\mathbf{H}\mathbf{u}[k_0] + \cdots + \mathbf{H}\mathbf{u}[k-1]$$

$$= \mathbf{F}^{k-k_0}\mathbf{x}[k_0] + \sum_{j=k_0}^{k-1} \mathbf{F}^{k-j-1}\mathbf{H}\mathbf{u}[j]$$

$$y[k] = \mathbf{C}\mathbf{F}^{k-k_0}\mathbf{x}[k_0] + \sum_{j=k_0}^{k-1} \mathbf{C}\mathbf{F}^{k-j-1}\mathbf{H}\mathbf{u}[j] + \mathbf{D}\mathbf{u}[k]$$