

Solution of HW02 for Units 3A, 3B: Dynamic Response, Control Systems

Assigned: October 8, 2021

Due: October 21, 2021 (23:59)

1. (Laplace Transform)

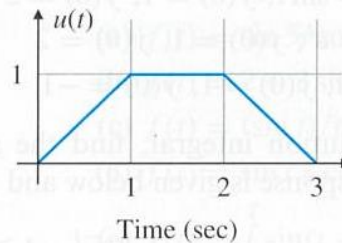
3.12 Consider the standard second-order system

$$G(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

- Write the Laplace transform of the signal in Fig. 3.47.
- What is the transform of the output if this signal is applied to $G(s)$?
- Find the output of the system for the input shown in Fig. 3.47.

Figure 3.47

Plot of input signal for Problem 3.12



Solution:

- (a) The input signal may be written as:

$$u(t) = t - (t-1) * 1(t-1) - (t-2) * 1(t-2) + (t-3) * 1(t-3),$$

where $1(t-\tau)$ denotes a delayed unit step. The Laplace transform of the input signal is:

$$U(s) = \frac{1}{s^2}(1 - e^{-s} - e^{-2s} + e^{-3s}).$$

We can verify this in MATLAB:

```
>> ilaplace(1/s^2*(1-exp(-s)-exp(-2*s)+exp(-3*s)))
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ans =

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t-heaviside(t-1)*(t-1)-heaviside(t-2)*(t-2)+heaviside(t-3)*(t-3)
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(b) The Laplace transform of the output if this input signal is applied is:

$$Y(s) = G(s)U(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \left(\left(\frac{1}{s^2} \right) \right) (1 - e^{-s} - e^{-2s} + e^{-3s}).$$

(c) However to make the mathematical manipulation easier, consider only the response of the system to a (unit) ramp input:

$$Y_1(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \left(\left(\frac{1}{s^2} \right) \right).$$

Partial fractions yields the following:

$$Y_1(s) = \frac{1}{s^2} - \frac{\frac{2\zeta}{\omega_n}}{s} + \frac{\frac{2\zeta}{\omega_n}(s + 2\zeta\omega_n - \frac{\omega_n}{2\zeta})}{(s + \omega_n\zeta)^2 + (\omega_n\sqrt{1-\zeta^2})^2}.$$

Use the following Laplace transform pairs for the case $0 \leq \zeta < 1$:

$$L^{-1}\left\{ \left\{ \frac{s + z_1}{(s + a)^2 + \omega^2} \right\} \right\} = \sqrt{\frac{(z_1 - a)^2 + \omega^2}{\omega^2}} e^{-at} \sin(\omega t + \phi),$$

where

$$\phi \equiv \tan^{-1}\left(\frac{\omega}{z_1 - a}\right).$$

$$L^{-1}\left\{ \left\{ \frac{1}{s^2} \right\} \right\} = t \quad \text{unit ramp}$$

$$L^{-1}\left\{ \left\{ \frac{1}{s} \right\} \right\} = 1(t) \quad \text{unit step}$$

and the following Laplace transform pairs for the case $\zeta = 1$:

$$L^{-1}\left\{ \left\{ \frac{1}{(s + a)^2} \right\} \right\} = te^{-at}.$$

$$L^{-1}\left\{ \left\{ \frac{s}{(s + a)^2} \right\} \right\} = (1 - at)e^{-at}.$$

$$L^{-1}\left\{ \left\{ \frac{1}{s^2} \right\} \right\} = t \quad \text{unit ramp,}$$

$$L^{-1}\left\{ \left\{ \frac{1}{s} \right\} \right\} = 1(t) \quad \text{unit step,}$$

the following is derived:

$$y_1(t) = \begin{cases} t - \frac{2\zeta}{\omega_n} + \frac{e^{-\zeta\omega_n t}}{\omega_n\sqrt{1-\zeta^2}} \sin(\omega_n\sqrt{1-\zeta^2}t + \tan^{-1} \frac{2\zeta\sqrt{1-\zeta^2}}{2\zeta^2-1}) & 0 \leq \zeta < 1 \\ & t \geq 0 \\ t - \frac{2}{\omega_n} + \frac{2}{\omega_n} e^{-\omega_n t} \left(\frac{\omega_n}{2} t + 1 \right) & \zeta = 1 \\ & t \geq 0 \end{cases}.$$

Since $u(t)$ consists of a ramp and three delayed ramp signals, using superposition (the system is linear), then:

$$y(t) = y_1(t) - y_1(t-1) - y_1(t-2) + y_1(t-3) \quad t \geq 0.$$

2. (Laplace Transform)

3.16 For a second-order system with transfer function

$$G(s) = \frac{5}{s^2 + s + 4},$$

Determine the following:

- (a) The DC gain and whether the system is stable.
- (b) The final value of the output if the input is applied with a step of 2 units or $R(s) = \frac{2}{s}$.

Solution:

- (a) The system is stable and therefore the Final Value Theorem is applicable, and the DC gain is

$$G(0) = \frac{5}{4} = 1.25$$

By finding the roots of $s^2 + s + 4 = 0$, the poles are located at $s = -0.5 \pm j1.94$. Since the poles are at the LHS of s -plane, the system is stable.

- (b) With a step input of 2 units or $R(s) = \frac{2}{s}$, the final value can be calculated as

$$\lim_{s \rightarrow 0} sG(s)R(s) = 2G(0) = 2.5 \text{ units.}$$

3. (Block diagram)

3.20 Find the transfer functions for the block diagrams in Fig. 3.50.

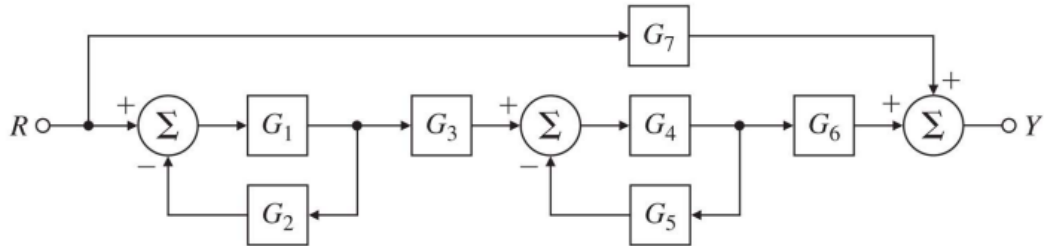
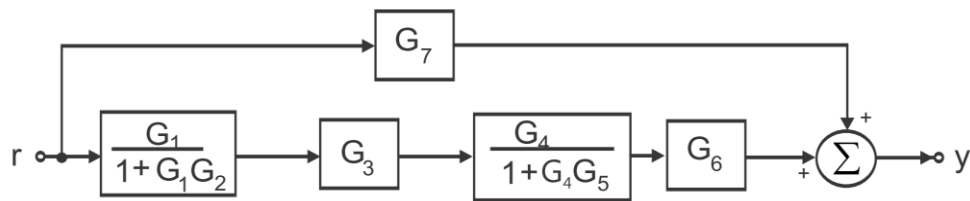
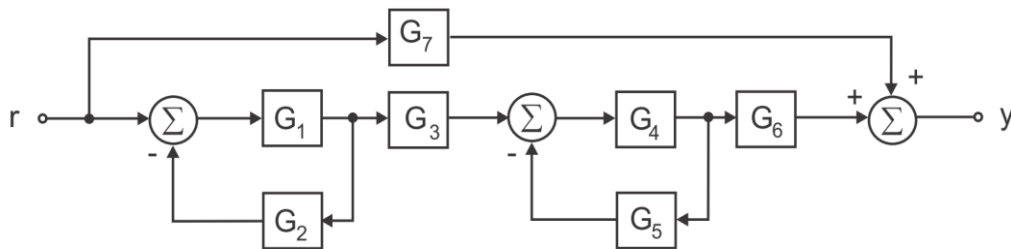


Figure 3.50 Block diagrams for Problem 3.20

Solution:

(b)



$$\frac{Y}{R} = G_7 + \frac{G_1 G_3 G_4 G_6}{(1 + G_1 G_2)(1 + G_4 G_5)}$$

4. (Time domain specification)

26. For the unity feedback system shown in Fig. 3.55, specify the gain and pole location of the compensator so that the overall closed-loop response to a unit-step input has an overshoot of no more than 25%, and a 1% settling time of no more than 0.1 sec. Verify your design using MATLAB.

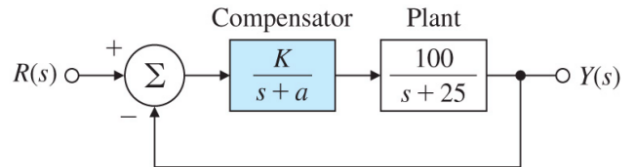


Figure 3.55: Unity feedback system for Problem 3.26

Solution:

$$\frac{Y(s)}{R(s)} = \frac{100K}{s^2 + (25 + a)s + 25a + 100K} = \frac{100K}{s^2 + 2\zeta\omega_n s + \omega_n^2}.$$

Using the given information:

$$\begin{aligned} R(s) &= \frac{1}{s} && \text{unit step,} \\ M_p &\leq 25\%, \\ t_s &\leq 0.1 \text{ sec.} \end{aligned}$$

Solve for ζ :

$$\begin{aligned} M_p &= e^{-\pi\zeta/\sqrt{1-\zeta^2}}, \\ \zeta &= \sqrt{\frac{(\ln M_p)^2}{\pi^2 + (\ln M_p)^2}} \geq 0.4037. \end{aligned}$$

Solve for ω_n :

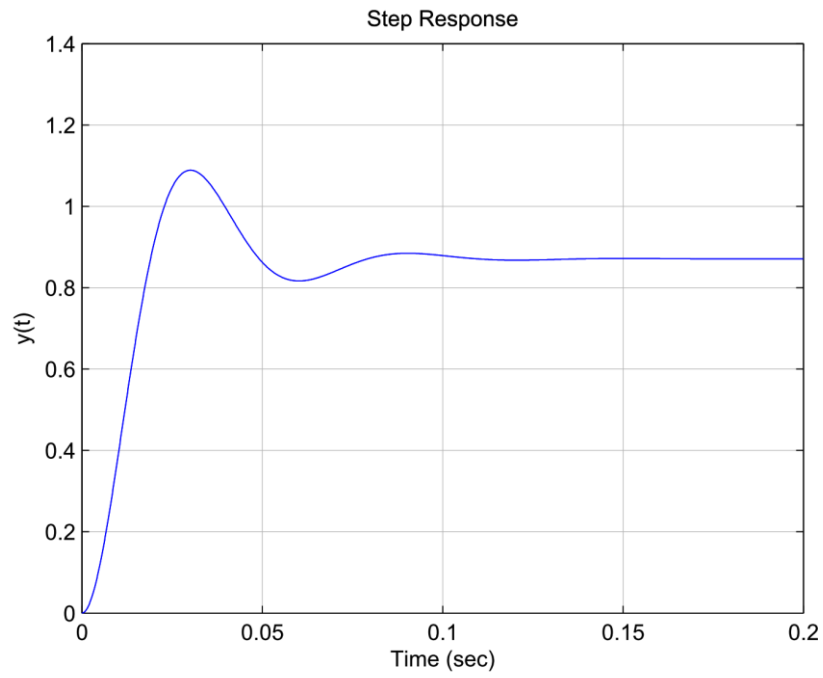
$$e^{-\zeta\omega_n t_s} = 0.01 \quad \text{For a 1\% settling time.}$$

$$\begin{aligned} t_s &\leq \frac{4.605}{\zeta\omega_n} = 0.1, \\ \implies \omega_n &\approx 114.07. \end{aligned}$$

Now find a and K :

$$\begin{aligned}2\zeta\omega_n &= (25 + a), \\a &= 2\zeta\omega_n - 25 = 92.10 - 25 = 67.10, \\ \omega_n^2 &= (25a + 100K), \\ K &= \frac{\omega_n^2 - 25a}{100} \approx 113.34.\end{aligned}$$

The step response of the system using MATLAB is shown next.



Step response for Problem 3.26.