

Fall 2021 (110-1)

控制系統
Control Systems

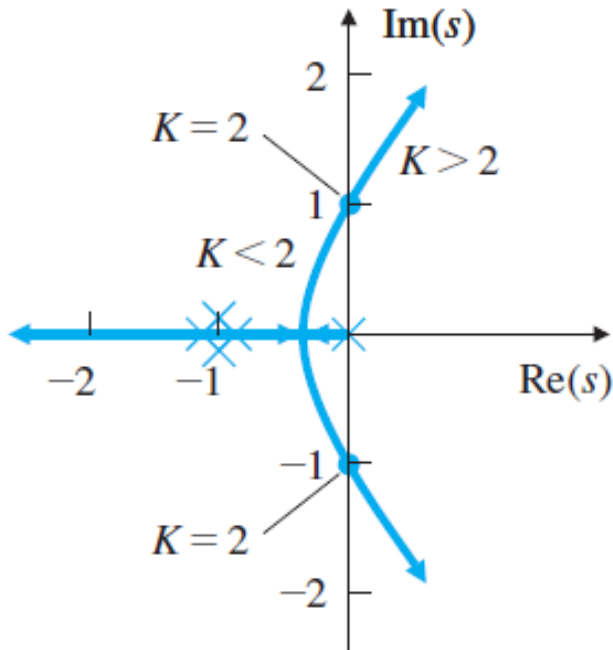
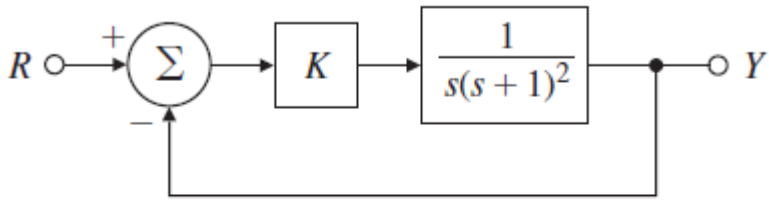
Unit 6F
Stability Margins

Feng-Li Lian

NTU-EE

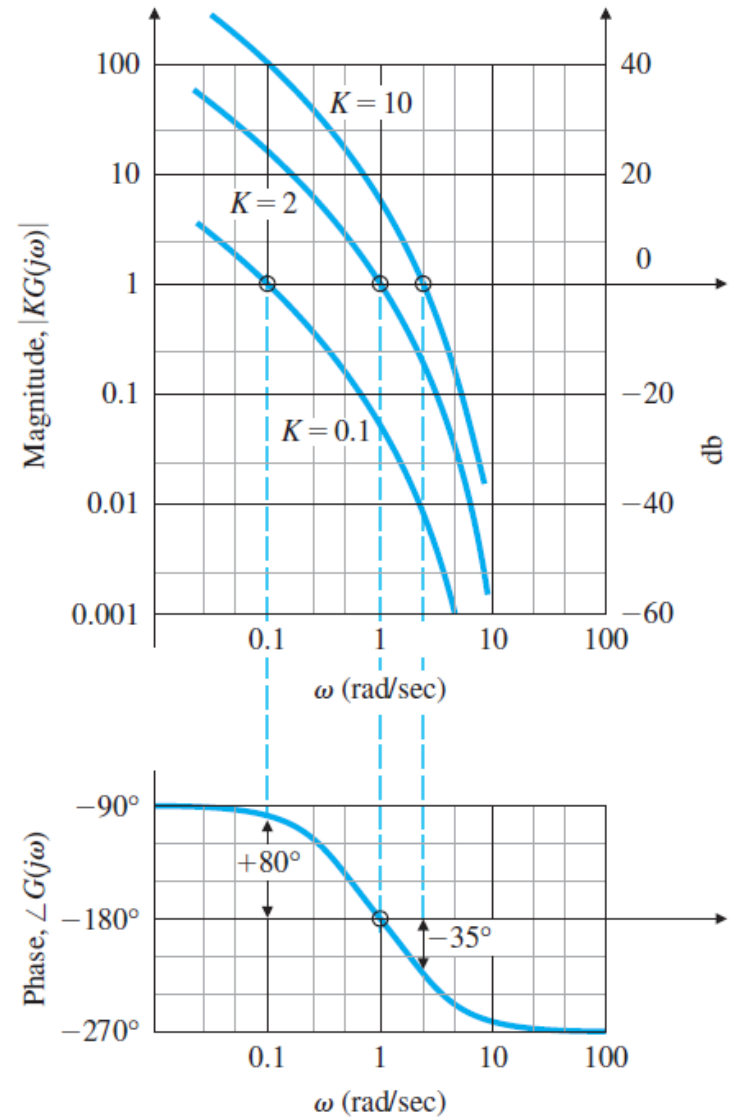
Sep 2021 – Jan 2022

In U6D



(b) $|K G(j\omega)| < 1$

$\angle G(j\omega) = -180^\circ$



- Gain Margin (GM) & Phase Margin (PM)
- Another measure of stability, originally defined by Smith (1958)
- Combine the two margins into Vector Margin / Complex Margin

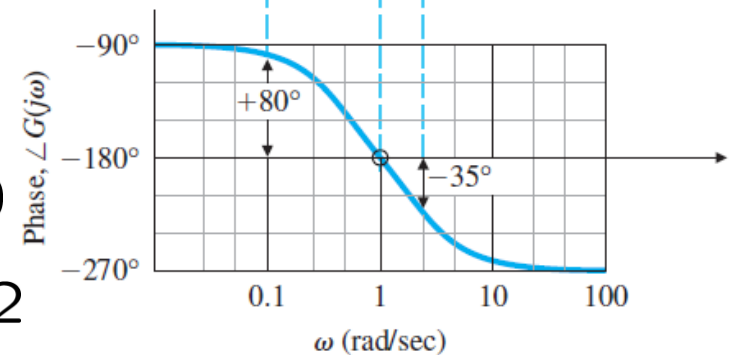
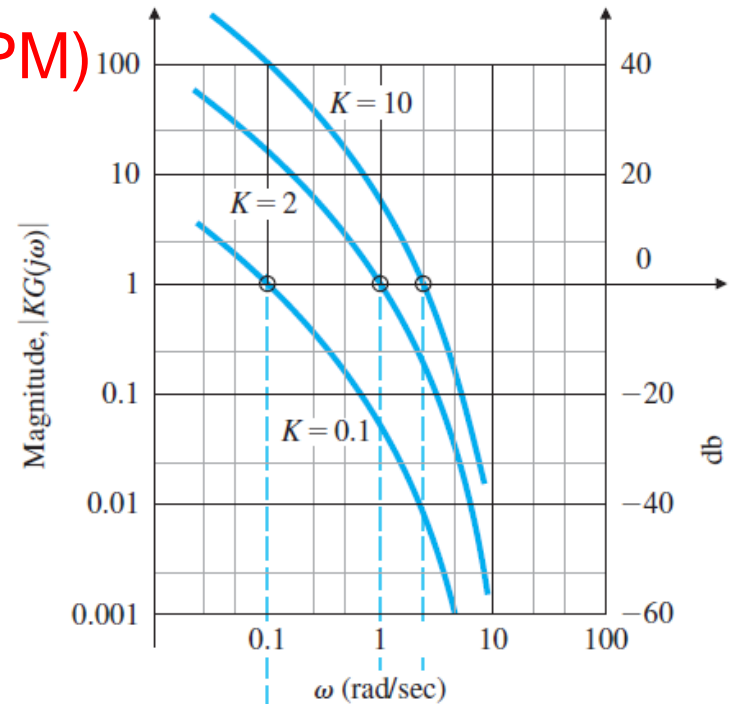
$$\angle G(j\omega) = -180^\circ$$

$$|K G(j\omega)| = 1$$

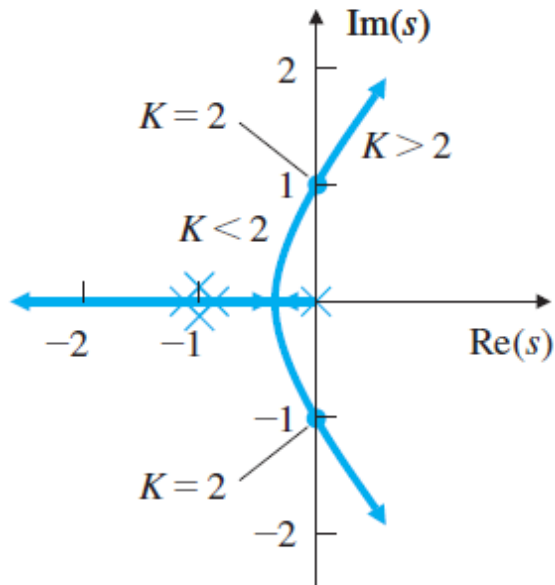
$$\Rightarrow K=2 \quad |K G(j\omega)| = 1 \quad \text{GM} = 1$$

$$\Rightarrow K=0.1 \quad |K G(j\omega)| = 0.05 \quad \text{GM} = 20$$

$$\Rightarrow K=10 \quad |K G(j\omega)| = 5 \quad \text{GM} = 0.2$$



Unstable



$$\angle G(j\omega) = -180^\circ$$

(b)

$$|K G(j\omega)| = 1$$

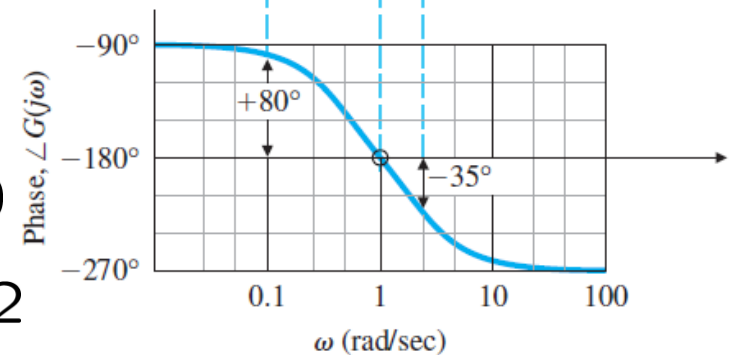
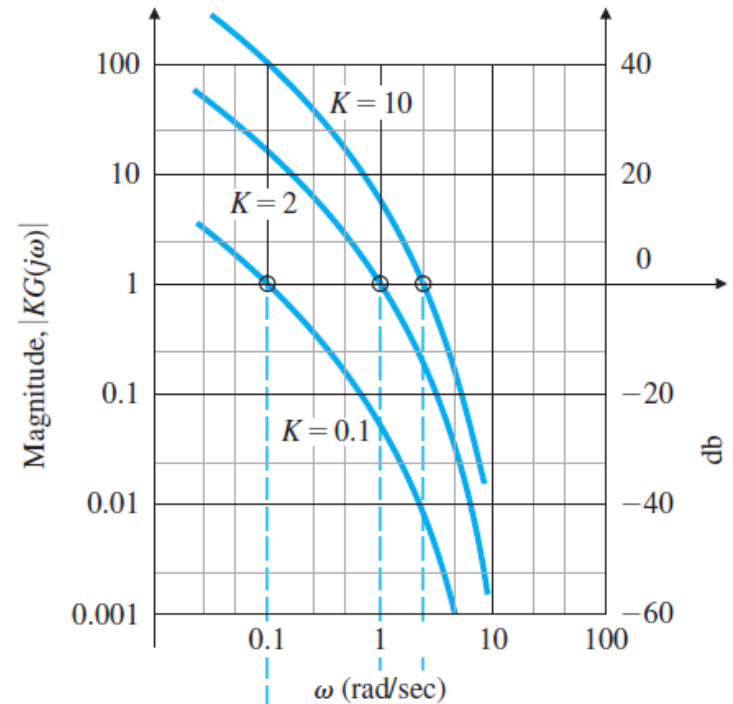
$$\Rightarrow K=2 \quad |K G(j\omega)| = 1 \quad \text{GM} = 1$$

$$\Rightarrow K=0.1 \quad |K G(j\omega)| = 0.05 \quad \text{GM} = 20$$

$$\Rightarrow K=10 \quad |K G(j\omega)| = 5 \quad \text{GM} = 0.2$$

$$\mathcal{T}(s) = \frac{Y(s)}{R(s)} = \frac{K G(s)}{1 + K G(s)}$$

■ Unstable



Gain Margin (GM) & Phase Margin (PM)

$$|KG(j\omega)| = 1$$

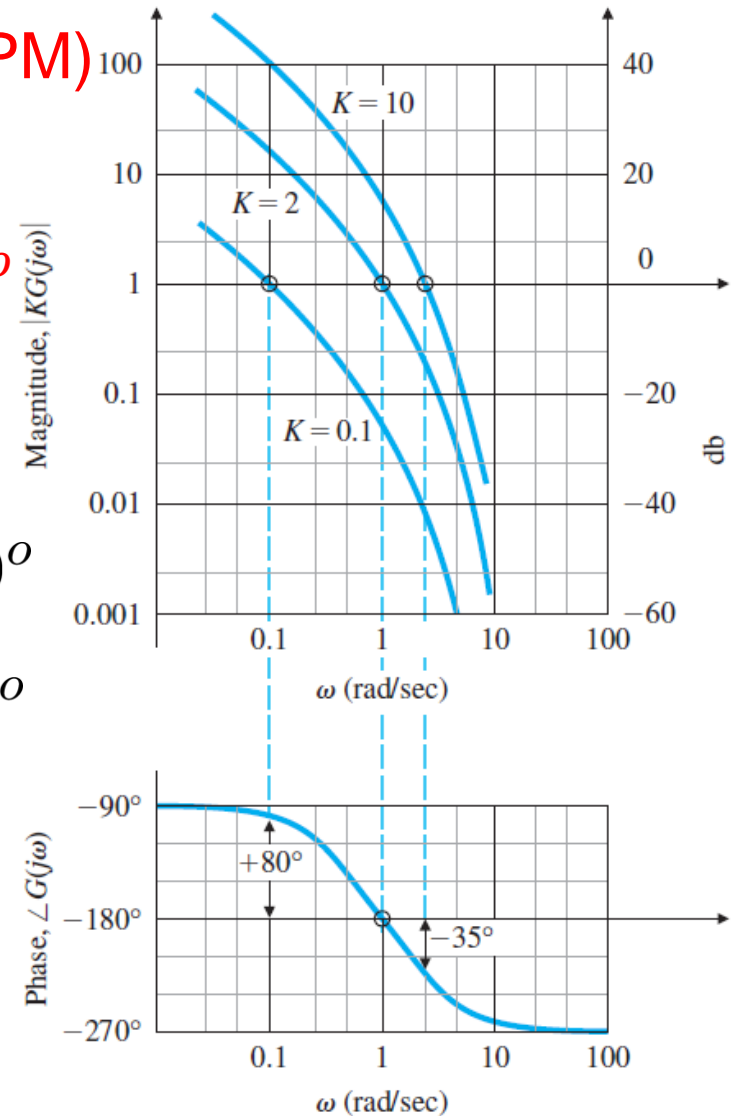
$$\Rightarrow \angle G(j\omega) = ? \quad \blacksquare \text{ Exceeds } -180^\circ$$

$$\Rightarrow K=2 \quad |KG(j\omega)| = 1 \quad \text{PM} = 0^\circ$$

$$\Rightarrow K=0.1 \quad |KG(j\omega)| = 1 \quad \text{PM} = +80^\circ$$

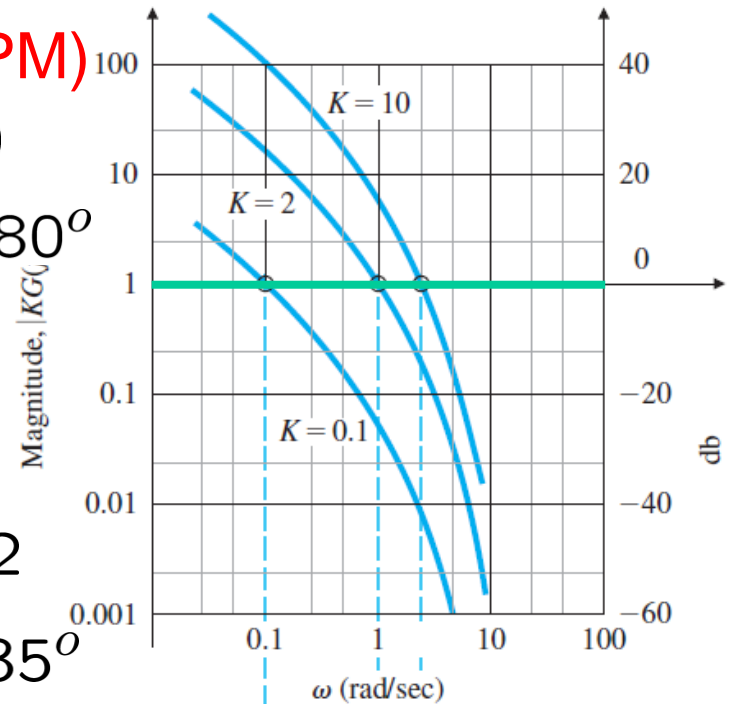
$$\Rightarrow K=10 \quad |KG(j\omega)| = 1 \quad \text{PM} = -35^\circ$$

Unstable

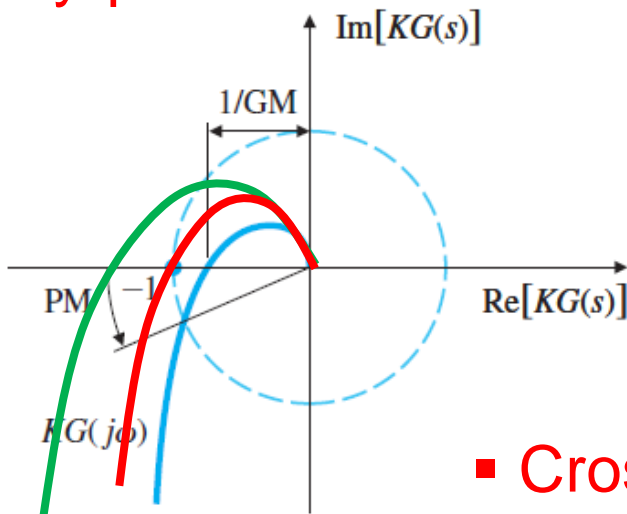


Gain Margin (GM) & Phase Margin (PM)

- ⇒ $K = 0.1$ $|KG(j\omega)| = 0.05$ GM = 20
 $|KG(j\omega)| = 1$ PM = +80°
- ⇒ $K = 2$ $|KG(j\omega)| = 1$ GM = 1
 $|KG(j\omega)| = 1$ PM = 0°
- ⇒ $K = 10$ $|KG(j\omega)| = 5$ GM = 0.2
 $|KG(j\omega)| = 1$ PM = -35°



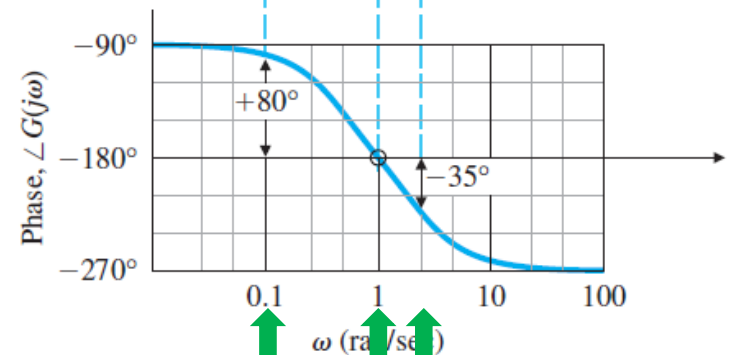
Nyquist Plot



$K = 0.1$

$K = 2$

$K = 10$

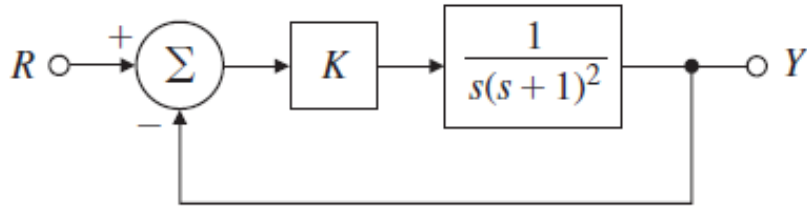


$|KG(j\omega)| = 1$

▪ Crossover Frequency, ω_c

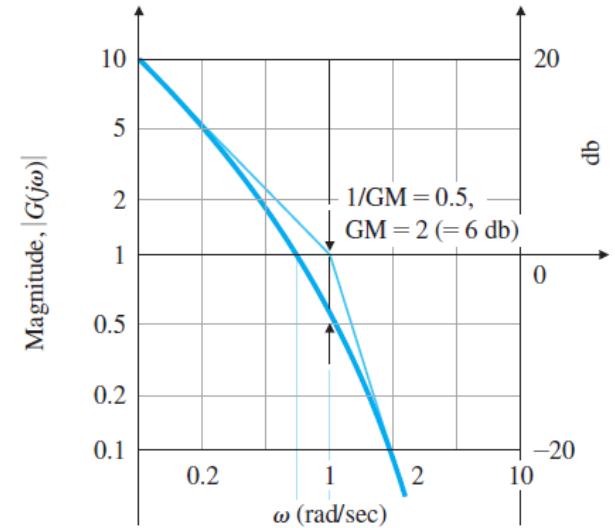
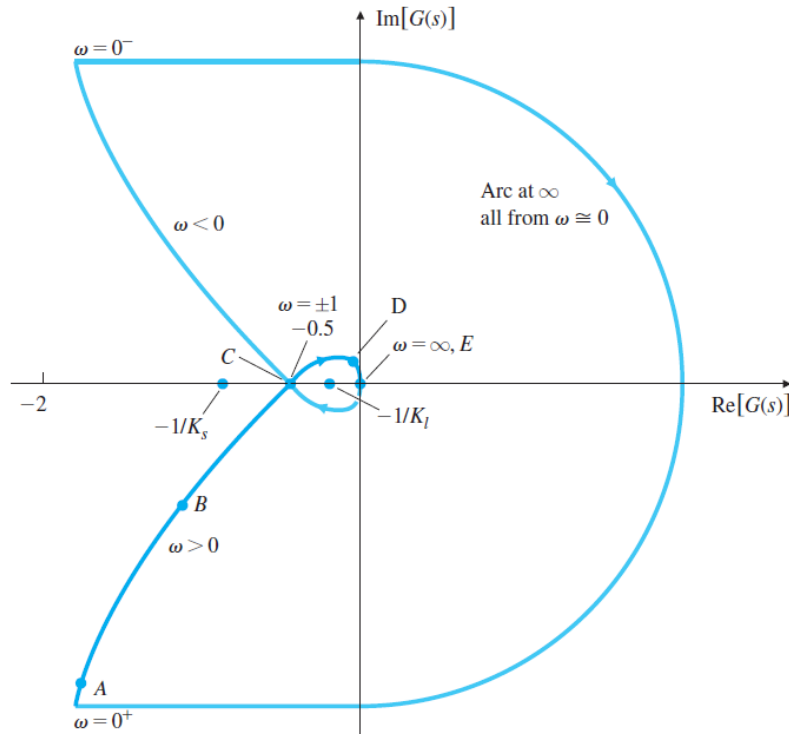
OR 0 db

- In Example 6.9:

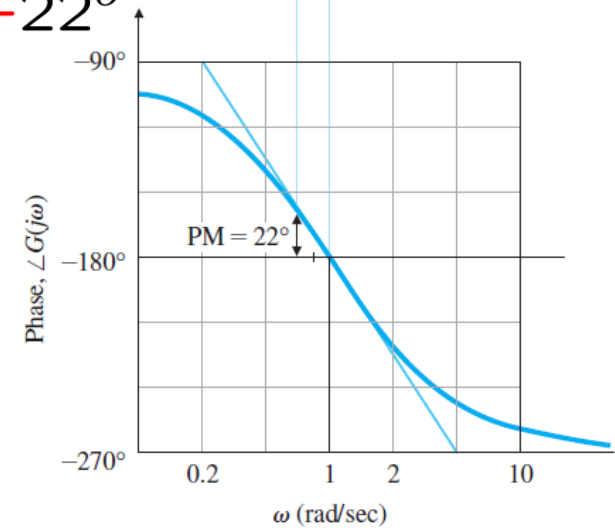


$GM = 2$

- Nyquist Plot

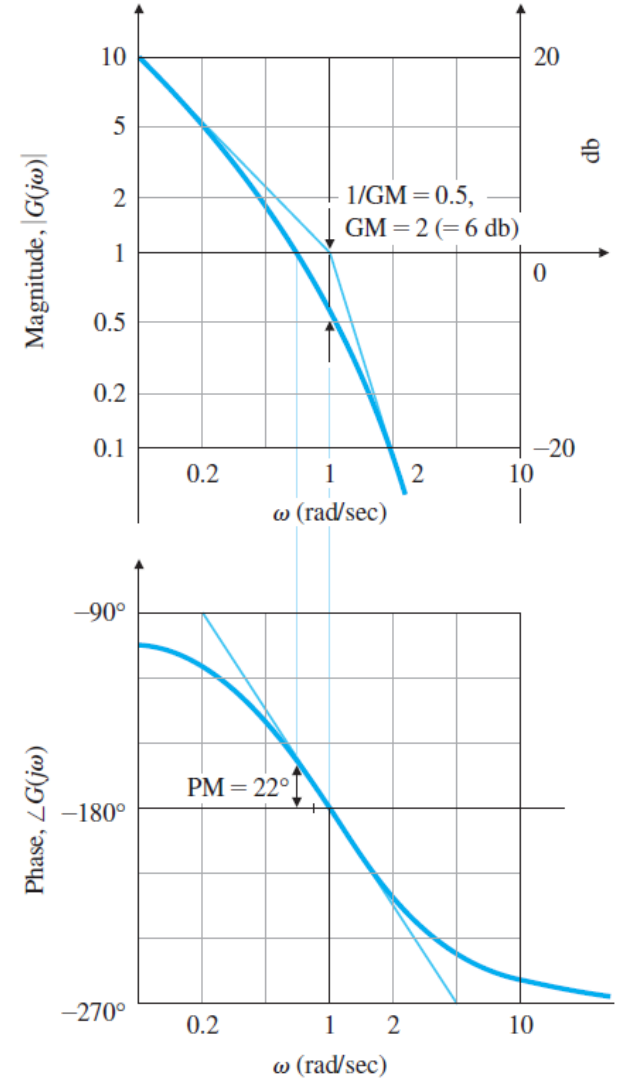
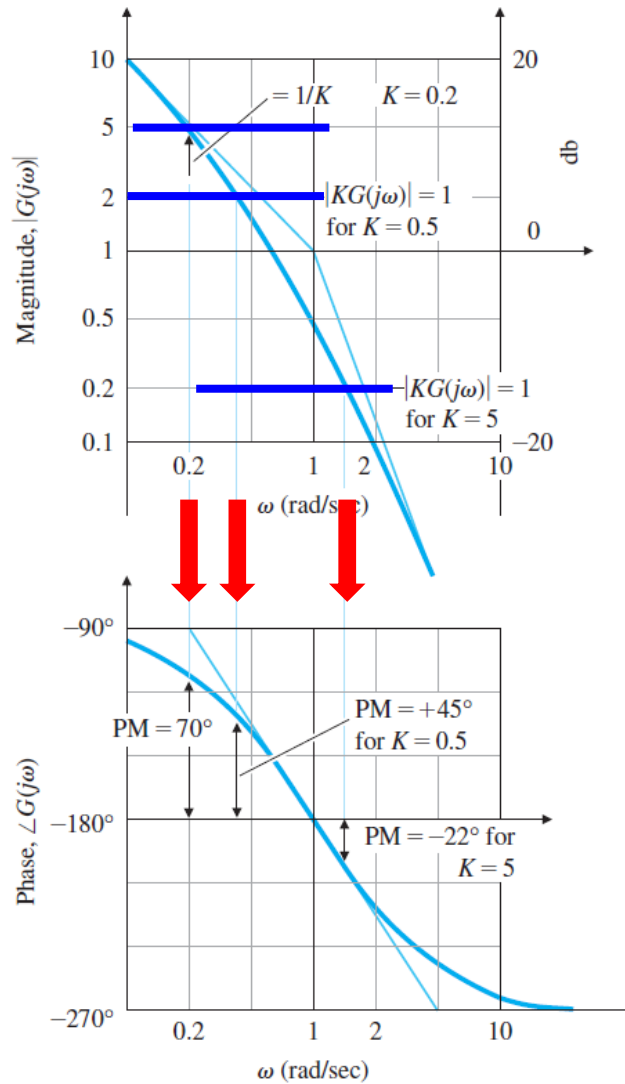


$PM = +22^\circ$



Stability Margins

- PM vs K
- $K = 5$
- $|KG(j\omega)| = 1$
- $PM = -22^\circ$
- $K = 0.5$
- $|KG(j\omega)| = 1$
- $PM = +45^\circ$
- $K = 0.2$
- $|KG(j\omega)| = 1$
- $PM = +70^\circ$



- The **PM** is more commonly used

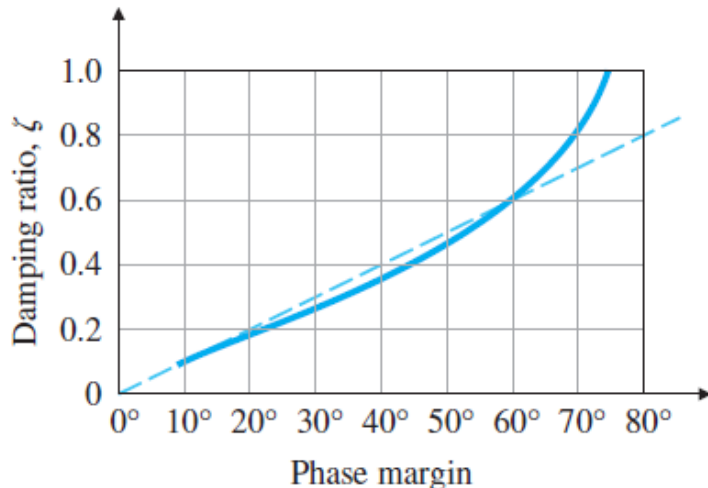
to specify control system performance

because it is more closely to the **damping ratio** of the system.

- For the **open-loop** 2nd-order system: $G(s) = \frac{\omega_n^2}{s(s + 2\zeta\omega_n)}$
- With **unity feedback**, produces the **closed-loop** system:

$$T(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

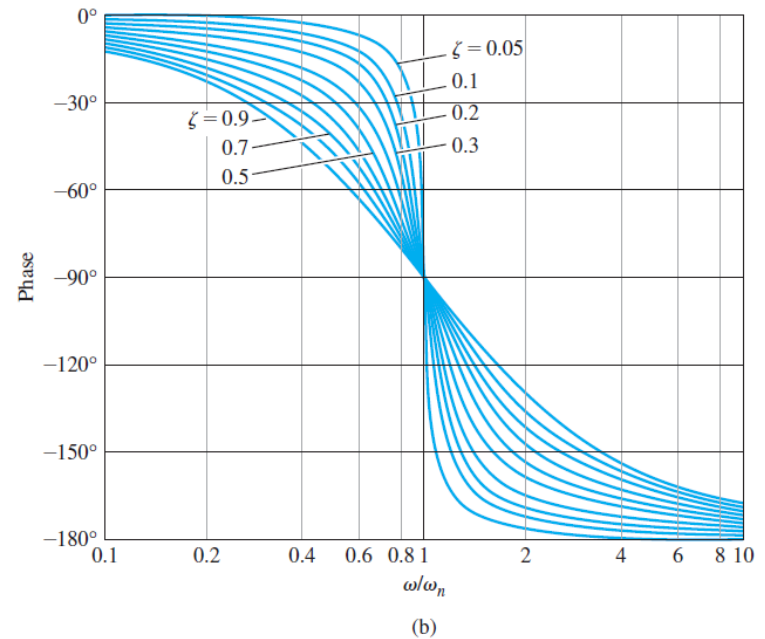
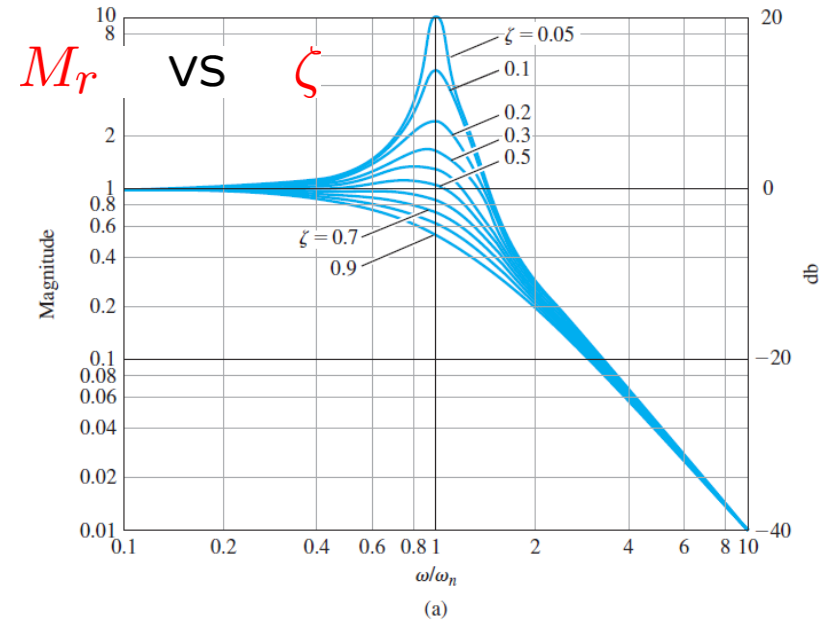
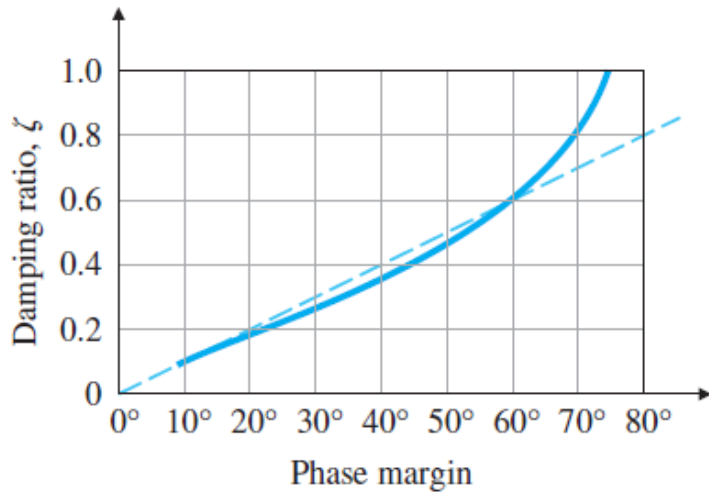
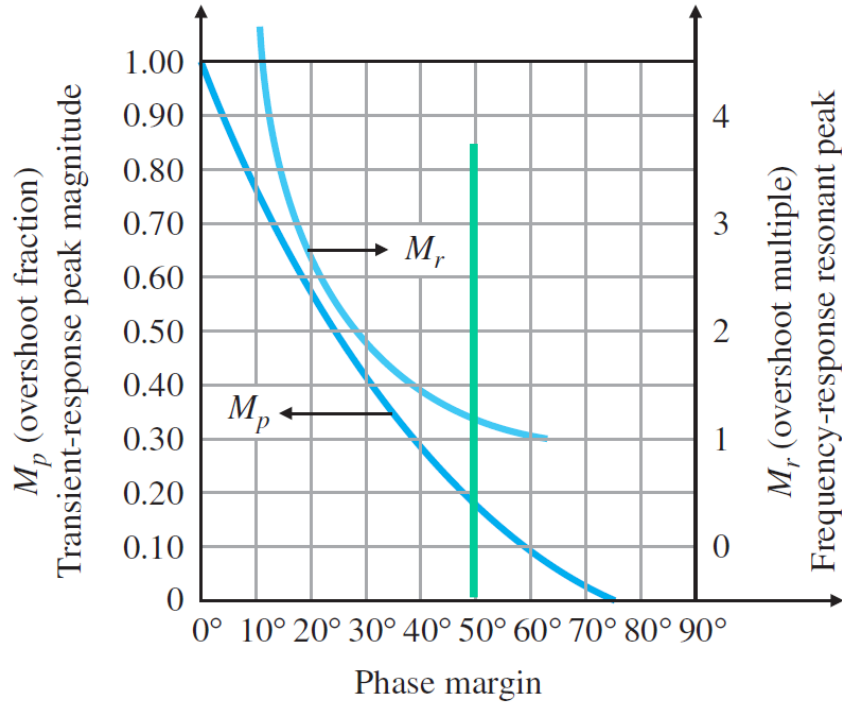
- The relationship between the **PM** and ζ is:



$$PM = \tan^{-1} \left[\frac{2\zeta}{\sqrt{\sqrt{1 + 4\zeta^4} - 2\zeta^2}} \right]$$

$$\zeta \approx \frac{PM}{100} \quad \text{for } PM < 60^\circ$$

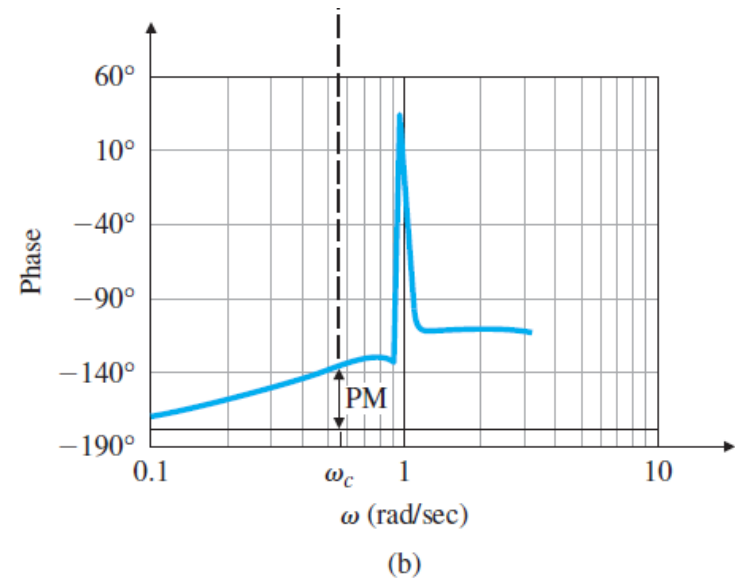
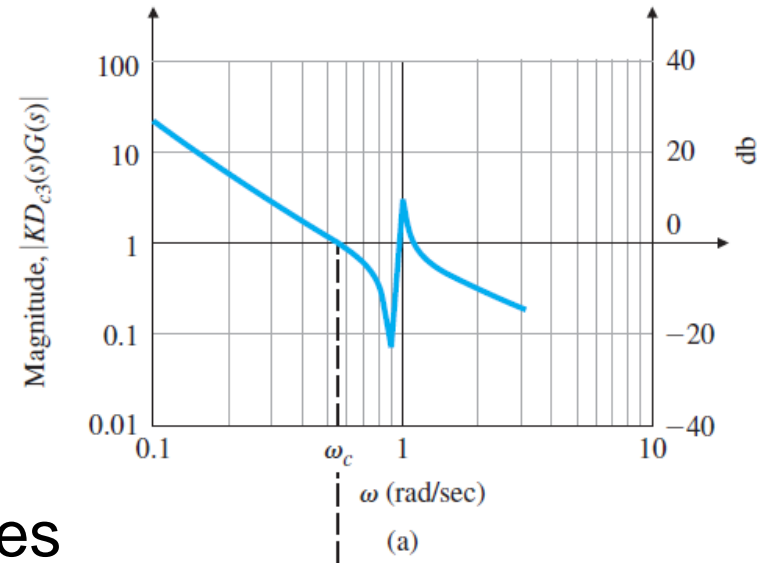
Stability Margins



- In some cases,
the **PM** and **GM** are **not useful indicators of stability**.
- For **1st- and 2nd-order** systems,
the **phase** never crosses the **180°** line;
- Hence, the **GM** is always ∞ and **not a useful** design parameter.
- For **higher-order** systems,
it is possible to have more than one frequency
where $|K G(j\omega)| = 1$ or where $\angle KG(j\omega) = 180^\circ$
- And the margins as previously defined need clarification.
- An example as follows:

- In Chapter 10
- The magnitude **crosses 1** three times
- Define **PM** by the **first crossing**
- Because the **PM at this crossing**
was the **smallest** of these 3 values

and thus the **most conservative** assessment of stability



- **Vector Margin** (or **Complex Margin**)

- The **distance** to the **-1 point**

from the **closest approach**

of the Nyquist Plot

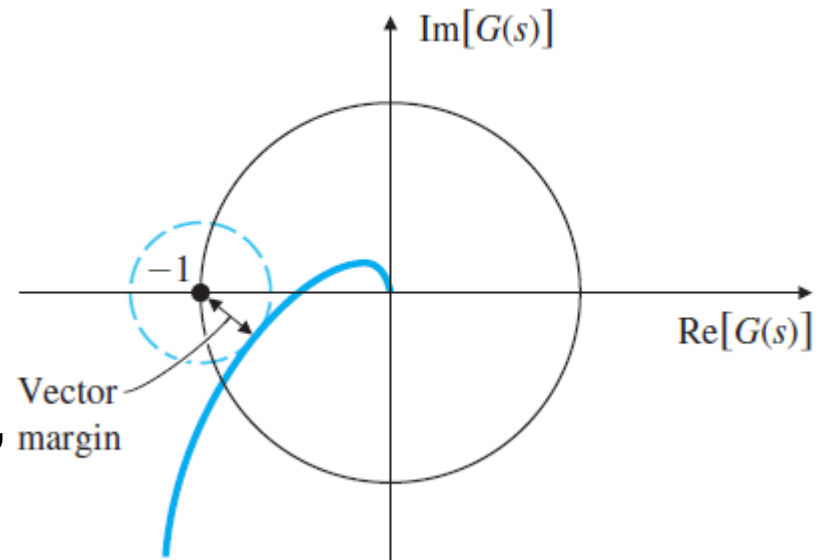
- **Vector margin** is

a **single** margin parameter,

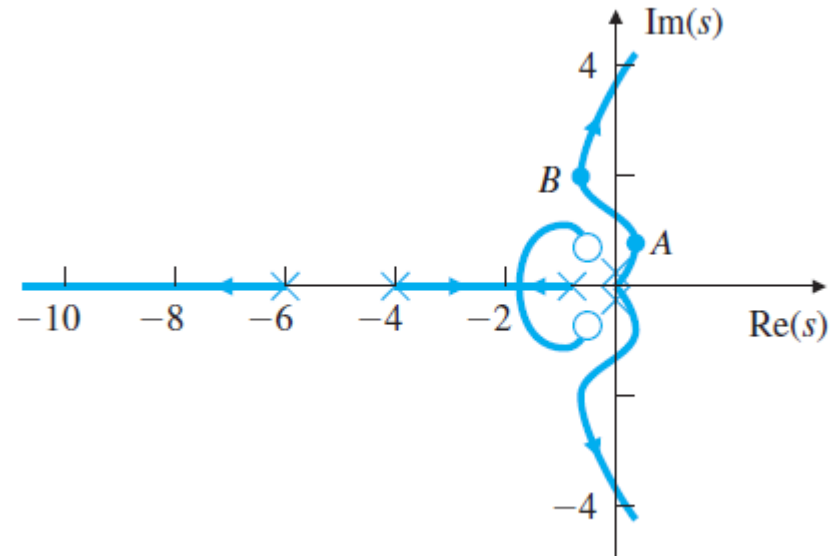
- It removes all the **ambiguities**

in assessing stability

that come with **using GM and PM** in combination.



- Conditionally Stable Systems
- Point A:
 - Increase gain
 - make **stable**
- Point B:
 - Increase/decrease gain
 - make **unstable**

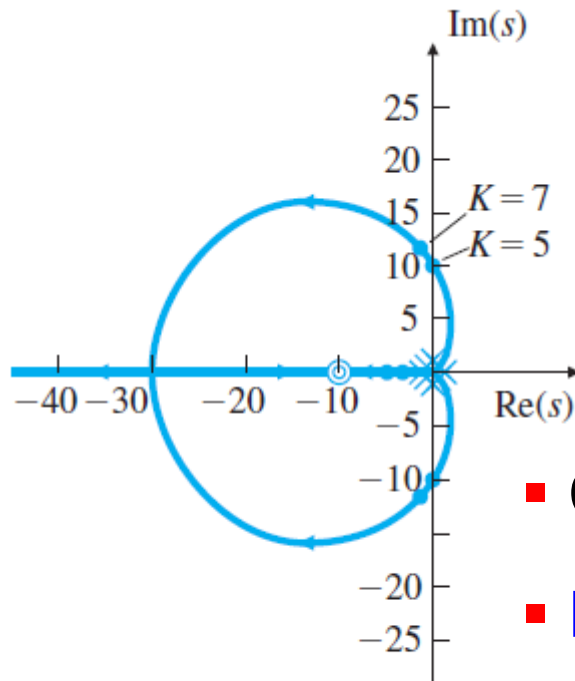


Examples

Example 6.12: Stability Properties for a Conditionally Stable System

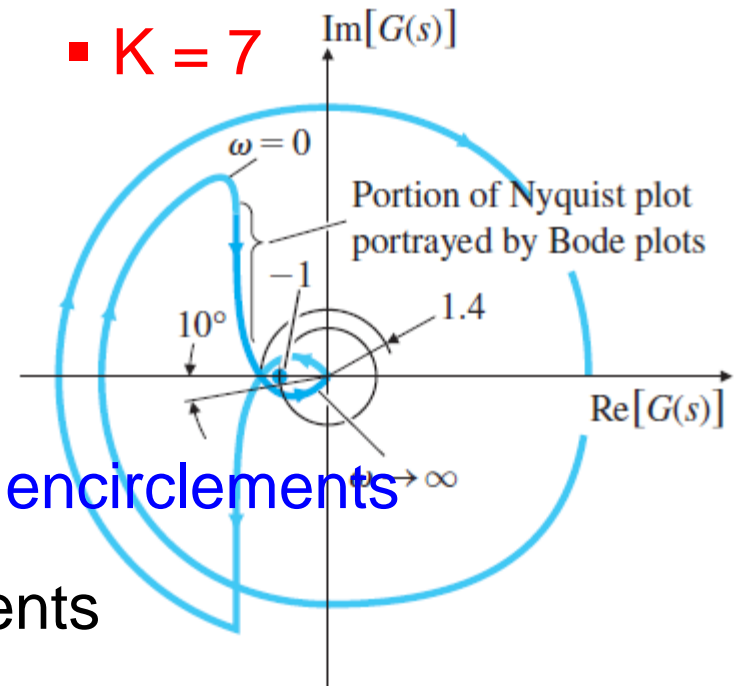
$$K G(s) = \frac{K (s + 10)^2}{s^3}$$

- Unstable: $K < 5$
- Stable: $K > 5$
- **Conflicting !!!**
- $PM = +10^\circ$ (Stable)
- $GM = 0.6$ (Unstable)



(a)

- Counting Nyquist encirclements
- No net encirclements
- \rightarrow Stable for $K = 7$



(b)

Examples

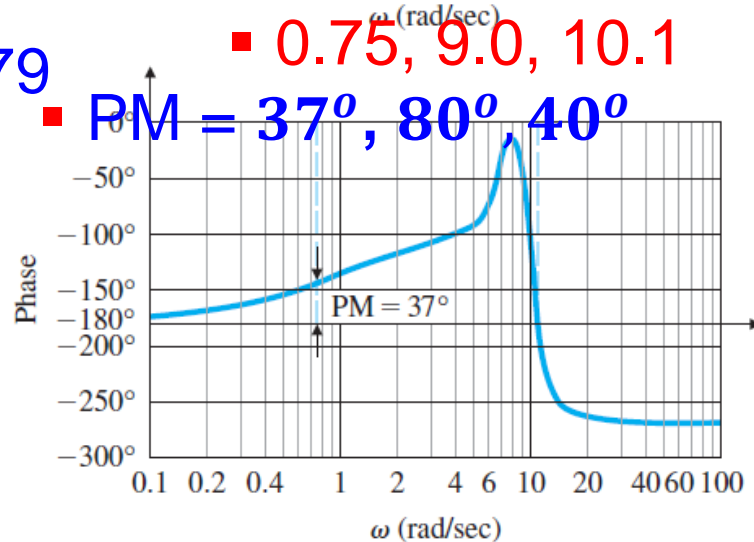
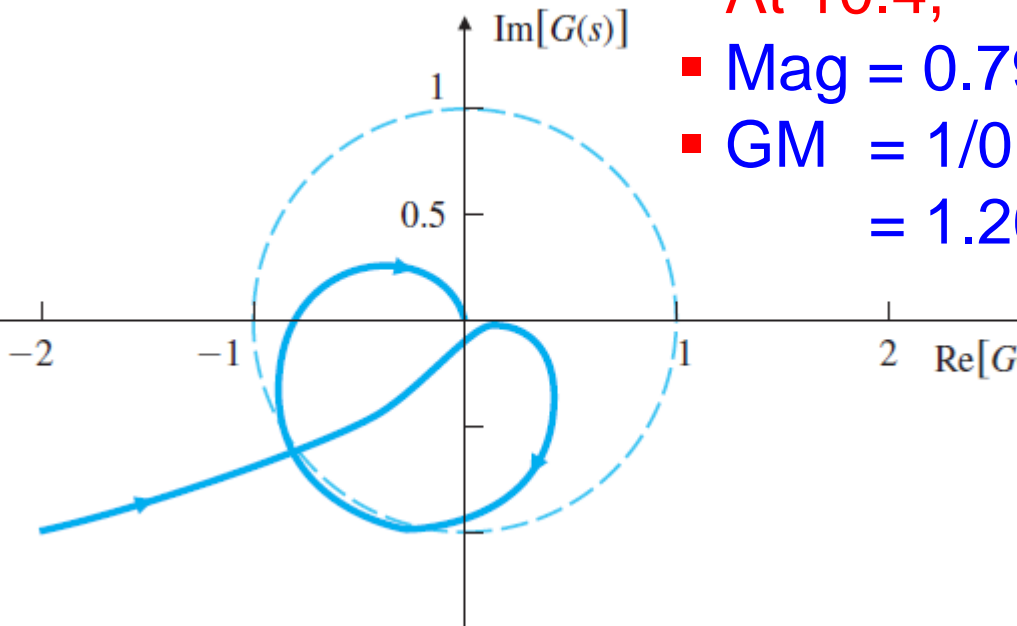
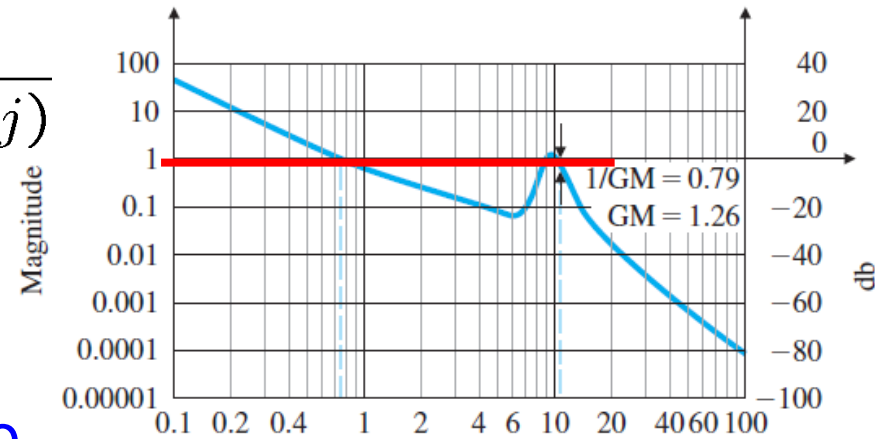
Example 6.13: Nyquist Plot for a System with Multiple Crossover Frequencies

$$G(s) = \frac{85 (s + 1) (s^2 + 2s + 43.25)}{s^2 (s^2 + 2s + 82) (s^2 + 2s + 101)}$$

$$= \frac{85 (s + 1) (s + 1 \pm 6.5j)}{s^2 (s + 1 \pm 9j) (s + 1 \pm 10j)}$$

3 crossover frequencies

- At 10.4,
- Mag = 0.79
- GM = 1/0.79 = 1.26



- 0.75, 9.0, 10.1
- PM = 37°, 80°, 40°