Fall 2021 (110-1)

# 控制系統 Control Systems

## Unit 5B Guidelines for Determining a Root Locus

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- Definition I:
- The root locus is the set of values of s
- for which 1 + KL(s) = 0 is satisfied

as the real parameter K varies from 0 to +  $\infty$ 

Typically, 1 + KL(s) = 0 is the characteristic equation of the system, and in this case the roots on the locus are the closed-loop poles of that system.

### Definition II:

- The root locus of L(s) is the set of points in the s-plane where the phase of L(s) is 180°.
- To test whether a point in the s-plane is on the locus, we define the angle to the test point from a zero as ψ and the angle to test point from a pole as φ as follows:

$$\sum \psi_i \ - \ \sum \phi_i \ = \ 180^o \ + \ 360^o$$
 (  $l \ - \ 1$  )

 $L(s) = \frac{b(s)}{a(s)}$ 

 $1 + \frac{k}{a(s)} = 0$ 

 $\Rightarrow L(s) = -\frac{1}{\kappa}$ 



#### Formal Definition of Root Locus

• Illustrative Example: 
$$L(s) = \frac{s+1}{s(s+5)[(s+2)^2+4)]}$$
  
 $s_0 = -1+2j$   
 $\angle L(s_0) = 180^o + 360^o (l-1)$   
 $= \sum \psi_i - \sum \phi_i$   
 $= \angle (s_0+1)$   
 $-\angle (s_0) - \angle (s_0+5)$   
 $-\angle [(s_0+2)^2+4]$   
 $= 90^o - 116.6^o - 0^o$   
 $-76^o - 26.6^o$   
 $= -129.2^o \neq 180^o$   
 $\Rightarrow s_0$  is not on the root locus

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- Rule 1:
- The n branches of the locus start at the poles of L(s) and
- m of these branches end on the zeros of L(s).
- a(s) + K b(s) = 0,
- If K = 0, then a(s) = 0, whose roots are the poles.
- When  $K \rightarrow \infty$ , then b(s) = 0 (m zeros) or  $s \rightarrow \infty$ . (the rest n-m)

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Real Axis (seconds<sup>-1</sup>)

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Rule 3:  
As 
$$K \rightarrow \infty$$
,  $L(s) = -\frac{1}{K}$   $\Rightarrow L(s) = 0$   
1) *m* roots will be found  
to approach the zeros of L(s)  
2)  $s \rightarrow \infty$  because  $n \ge m$   
that is,  
 $n - m$  roots approach  $s \rightarrow \infty$   
 $\Rightarrow 1 + K \frac{b(s)}{a(s)} = 0$   
 $\Rightarrow 1 + K \frac{s^m + b_1 s^{m-1} + b_2 s^{m-2} + \dots + b_{m-1} s + b_m}{s^n + a_1 s^{n-1} + a_2 s^{n-2} + \dots + a_{n-1} s + a_n} = 0$   
Can be approximated by  
 $\Rightarrow 1 + K \frac{1}{(s - \alpha)^{n-m}} = 0$ 

Rule 3:

$$\Rightarrow 1 + K \frac{1}{(s-\alpha)^{n-m}} = 0$$

• The search point:  $s_0 = R e^{j\phi}$   $l = 1, 2, \cdots, (n - m)$  $(n - m) \phi_l = 180^o + 360^o (l - 1)$  $\Rightarrow \phi_l = \frac{180^o + 360^o (l - 1)}{(n - m)}$ 

• For this example:  $L(s) = \frac{1}{s [(s + 4)^2 + 16)]}$  (n - m) = 3  $\phi_{1,2,3} = 60^{\circ}, 180^{\circ}, 300^{\circ},$ or  $\pm 60^{\circ}, 180^{\circ}$ 





CS5B-RLGuidelines - 11 Rules for Determining a Positive (180°) Root Locus Feng-Li Lian © 2021 s + 1 $L(s) = \frac{3}{s \left[ (s + 4)^2 + 16 \right]}$ • Rule 3: Determine asymptotic lines: s + 1 $=\frac{1}{s^3+8s^2+32s+0}$  $\Rightarrow a(s) + K b(s) = 0$  $(s^{3} + 8s^{2} + 32s) + K(s+1) = 0$  $\Rightarrow s^{n} + a_{1}s^{n-1} + a_{2}s^{n-2} + \dots + a_{n-1}s + a_{n}$  $+K(s^{m}+b_{1}s^{m-1}+b_{2}s^{m-2}+\cdots+b_{m-1}s+b_{m}) = 0$  $= (s - r_1) (s - r_2) \cdots (s - r_{n-1}) (s - r_n) = 0$ ● If m < n - 1:  $= -\sum r_i$  $\Rightarrow a_1 = -r_1 - r_2 \cdots - r_{n-1} - r_n$  And this term is independent of K The open-loop sum and closed-loop sum are the same

and are equal to  $-a_1$ 

 $\Rightarrow -\sum r_i = -\sum p_i$ 

CS5B-RLGuidelines - 12 Rules for Determining a Positive (180°) Root Locus Feng-Li Lian © 2021 Rule 3:  $L(s) = \frac{1}{s[(s + 4)^2 + 16)]}$ For large values of *K*: of the roots  $r_i$  approach the zeros  $z_i$ • m • n - m of the roots  $r_i$  approach the branches of the asymptotic system whose poles add up to  $(n - m) \alpha$  $\overline{(s-\alpha)^{n-m}}$  $\Rightarrow -\sum r_i = -(n-m)\alpha - \sum z_i = -\sum p_i$  $2.67 \times \sqrt{3} = 4.62$  $\Rightarrow \alpha = \frac{\sum p_i - \sum z_i}{(n-m)}$ 4 For this example: maginary axis 2  $\Rightarrow \alpha = \frac{-4-4+0}{3-0}$ 0 -2 $=-\frac{8}{3}=-2.67$ -4 $-\frac{6}{10}$  $= \pm 60^{\circ}, 180^{\circ}$ -5 0 5  $\phi_{1.2.3}$ Real axis

 $\Rightarrow \alpha = \frac{\sum p_i - \sum z_i}{(n-m)}$ 

- Rule 3:
- For large values of K:
  - *m* of the roots  $r_i$  approach the zeros  $z_i$
  - n m of the roots  $r_i$  approach the branches of the asymptotic system



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• Rule 4: 
$$\sum \psi_i - \sum \phi_i = 180^o + 360^o (l - 1)$$

The angle of departure of a branch of the locus from a single pole is given by

 $\sum \phi_i$ 

$$\phi_{dep} \,=\, \sum \psi_i \,-\, \sum_{i 
eq dep} \phi_i \,-\, 180^o \,\,-\, 360^o \,(\, l \,-\, 1\,)$$

the sum of the angles to the remaining poles

 $\sum \psi_i$  the sum of the angles to all the zeros

• The angle of departure of a branch of the locus from repeated poles with multiplicity *q* is given by

$$q \, \phi_{l,dep} = \sum \psi_i - \sum_{i \neq l,dep} \phi_i - 180^o - 360^o (l-1)$$
  
 $l = 1, 2, \cdots, q$ 



- Rule 5:
- The locus can have multiple roots at points on the locus and the branches will approach a point of q roots at angles separated by  $180^{\circ} - 360^{\circ}(l-1)$

 $\boldsymbol{q}$ 

• And will depart at angles with same separation.



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- Rule 5:
- Continuation Locus:
- $L(s) = \frac{b(s)}{a(s)} = \frac{1}{s(s+1)}$  $K_1 = \frac{1}{4} \implies K = K_1 + K_2 = \frac{1}{4} + K_2$  $\Rightarrow s^2 + s + \frac{1}{4} + K_2 = 0$  $\Rightarrow \left(s + \frac{1}{2}\right)^2 + K_2 = 0$  $K_2 = 0 \implies s_{1,2} = -\frac{1}{2}$  is 0.5 $2 \phi_{dep} = -180^o - 360^o (l-1)$  if -0.5 $\phi_{dep} = \pm 90^{\circ}$  $^{-1}$  $-1.5_{-2}^{-1}$  $\phi_{arr} = 0^{\circ}, 180^{\circ}$



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$$L(s) = \frac{1}{s(s^2 + 8s + 32)}$$

s = tf( 's' )

sysL = ( 1 )/(s\*( s^2+8\*s+32 )); sysL = 1/(s\*( (s+4)^2+16 ) );

rlocus( sysL );



Summary of the Rules for Determining a Root Locus

- Rule 1:
- The n branches of the locus start at the poles of L(s) and
- m of these branches end on the zeros of L(s).
- Rule 2:
- The loci are on the real axis to the left

of an odd number of poles and zeros.

- Rule 3:
- For large s and K,

n-m branches of the loci are asymptotic to lines at angles  $\phi$  radiating out from the point s =  $\alpha$  on the real axis, where  $\frac{180^{\circ} + 360^{\circ} (l - 1)}{n - m}$   $\alpha = \frac{\sum p_i - \sum z_i}{n - m}$ 

 $l = 1, 2, \cdots, n-m$ 

Summary of the Rules for Determining a Root Locus

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 $\cdot, q$ 

• Rule 4: 
$$\sum \psi_i - \sum \phi_i = 180^o + 360^o (l - 1)$$

The angle of departure of a branch of the locus from repeated poles with multiplicity q is given by

$$q \phi_{l,dep} = \sum \psi_i - \sum_{i \neq l,dep} \phi_i - 180^o - 360^o (l-1)$$
  
 $l = 1, 2, \cdots$ 

The angle of arrival of a branch at a zero with multiplicity q is given by

$$q \psi_{l,arr} = \sum \phi_i - \sum_{i \neq l,arr} \psi_i + 180^o + 360^o (l-1)$$

Rule 5:

• The locus can have multiple roots at points on the locus and the branches will approach a point of q roots at angles separated by  $180^{\circ} - 360^{\circ}(l-1)$ 

#### $\boldsymbol{q}$

• And will depart at angles with same separation.

The positive root locus

is a plot of all possible locations for roots to the equation 1 + K L(s) = 0 for some real positive value of K.

The purpose of design

 is to select a particular value of K
 that will meet the specifications
 for static and dynamic response.



- Compute the error constant of the control system
- For example, the steady-state error in tracking a ramp input is giving by the velocity constant:

$$K_v = \lim_{s \to 0} s K L(s) \qquad K_v = \lim_{s \to 0} s L(s)$$
$$= \lim_{s \to 0} s K \frac{1}{s [(s + 4)^2 + 16)]}$$
$$= \frac{K}{32}$$
$$= \frac{65}{32} \approx 2 \sec^{-1}$$