Spring 2020

控制系統 Control Systems

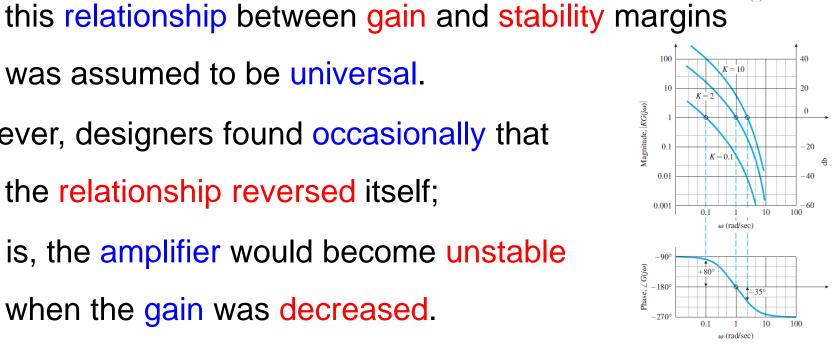
Unit 6E
The Nyquist Stability Criterion

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NTU-EE

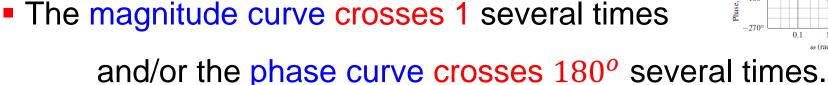
Mar 2020 – Jul 2020

- For most systems,
 - an increasing gain eventually causes instability
- In very early days of feedback control design,
- was assumed to be universal.
- However, designers found occasionally that the relationship reversed itself;
- That is, the amplifier would become unstable when the gain was decreased.
- The confusion motivated Harry Nyquist of Bell Tele Lab in 1932
- The Nyquist Stability Criterion



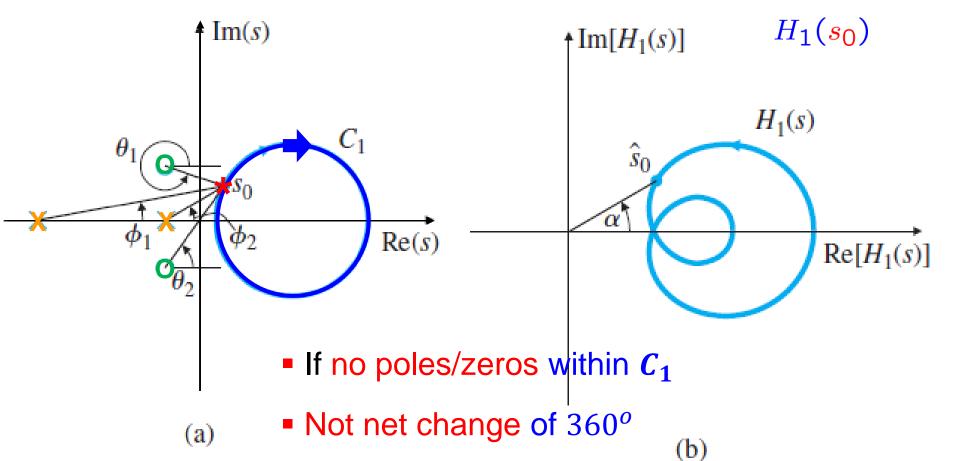
- Nyquist Stability Criterion:
- Based on the Argument Principle in complex variable theory.
- Relate OL frequency response
 to the number of CL poles in the RHP
- Determine stability

from frequency response of a complex system



- Deal with (a) OL unstable systems,
 - (b) non-minimum-phase systems,
 - (c) systems with pure delays

- -- ()
 - $H_1(s_0) = \overrightarrow{v} = |\overrightarrow{v}| e^{j \alpha}$
- argument: $\alpha = \theta_1 + \theta_2 (\phi_1 + \phi_2)$
- Contour Evaluation

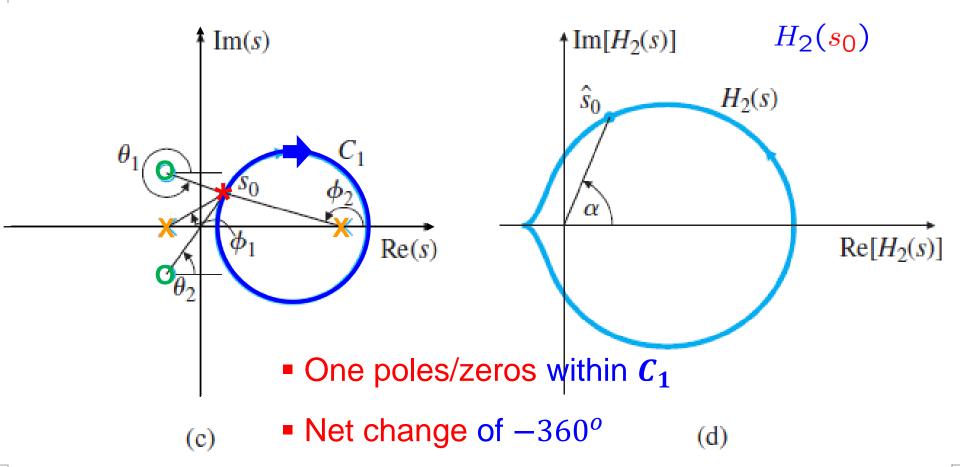


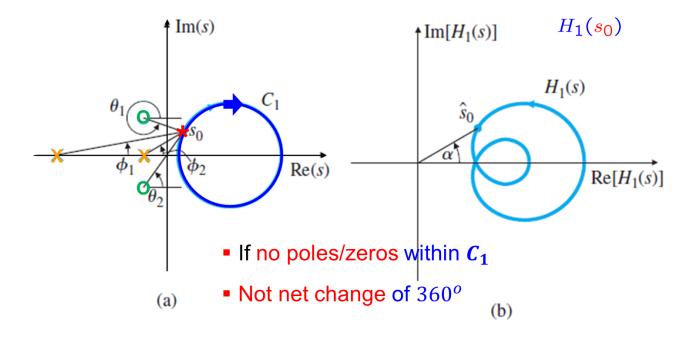
$$\mathbf{H}_{2}(s)$$

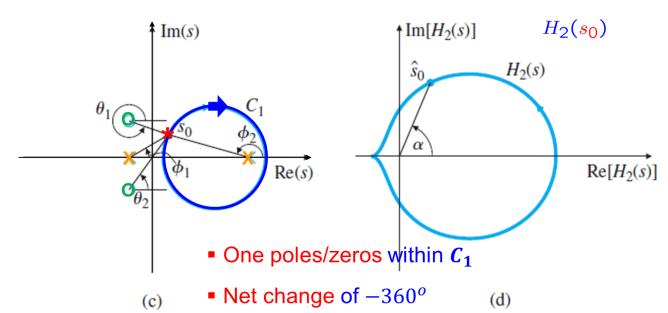
$$H_2(s_0) = \overrightarrow{v} = |\overrightarrow{v}| e^{j \alpha}$$

argument:
$$\alpha = \theta_1 + \theta_2 - (\phi_1 + \phi_2)$$

Contour Evaluation

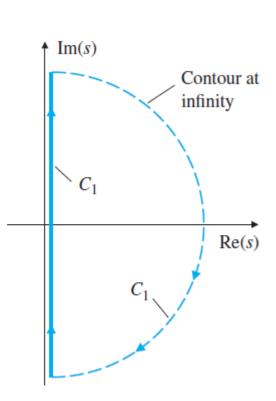


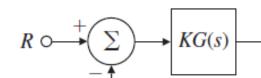




- The essence of the Argument Principle
- A contour map of a complex function
 will encircle the origin Z P times,
- where Z is the number of zeros
 and P is the number of poles
 of the function inside the contour.
- For controller design,

let the C_1 contour encircle entire RHP, where a pole would cause an unstable system.

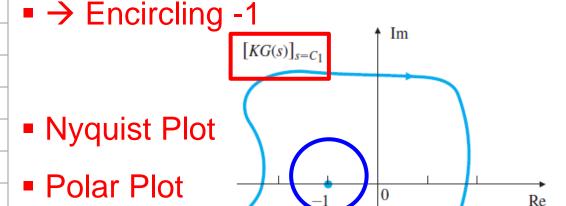


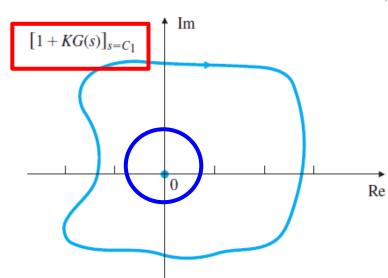


■ → Encircling the origin!

$$T(s) = \frac{Y(s)}{R(s)} = \frac{K G(s)}{1 + K G(s)}$$

- CL roots: 1 + KG(s) = 0
- The contour evaluation for 1 + KG(s) = 0
- Equivalently, the contour evaluation for KG(s) = 0
- Equivalently, the contour evaluation for $M \circ (s) = 0$





- $1 + KG(s) = 1 + K\frac{b(s)}{a(s)} = \frac{a(s) + Kb(s)}{a(s)}$
- poles of 1 + KG(s) = poles of G(s)
- A clockwise contour C₁ enclosing a zero of 1 + KG(s)
- result in KG(s) encircling the -1 point in a clockwise direction
- Likewise, C₁ enclose a pole of 1 + KG(s) (if there is an unstable OL pole)
- there will be a counterclockwise encirclement of the -1 point.
- Furthermore, two poles or zeros are in the RHP, KG(s) will encircle the -1 point twice, and so on.
- Net number of CW encirclements N = Z P
 - Z = zeros in RHP, P = poles in RHP

Procedure for Determining Nyquist Stability

- 1. Plot KG(s) for $-j\infty \le s \le +j\infty$. Do this by first evaluating $KG(j\omega)$ for $\omega = 0$ to ω_h , where ω_h is so large that the magnitude of $KG(j\omega)$ is negligibly small for $\omega > \omega_h$, then reflecting the image about the real axis and adding it to the preceding image. The magnitude of $KG(j\omega)$ will be small at high frequencies for any physical system. The Nyquist plot will always be symmetric with respect to the real axis. The plot is normally created by the NYQUIST Matlab function.
- 2. Evaluate the number of clockwise encirclements of -1, and call that number N. Do this by drawing a straight line in any direction from -1 to ∞ . Then count the net number of left-to-right crossings of the straight line by KG(s). If encirclements are in the counterclockwise direction, N is negative.
- 3. Determine the number of unstable (RHP) poles of G(s), and call that number P.
- 4. Calculate the number of unstable closed-loop roots Z:

$$Z = N + P. (6.28)$$

-90°

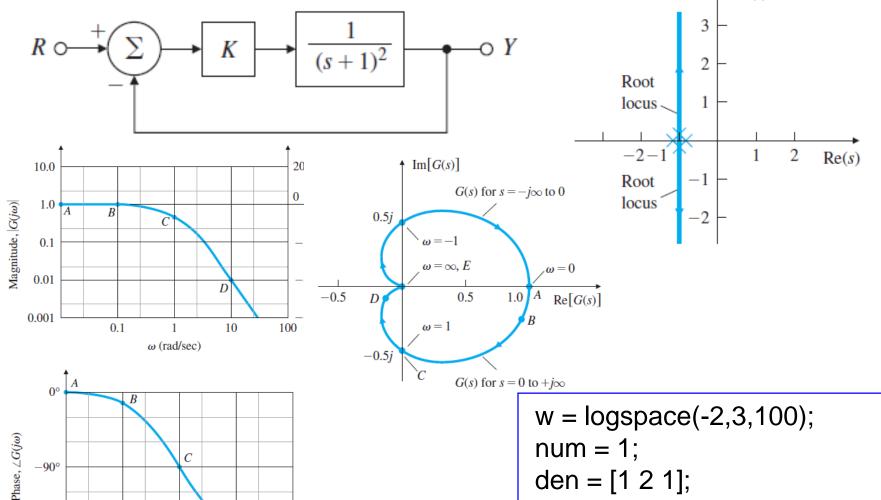
 -180°

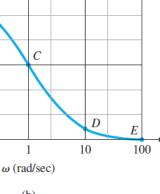
0.1

(b)

Im(s)

Example 6.8: Nyquist Plot for a Second-Order System





num = 1; $den = [1 \ 2 \ 1];$

[re,im] = nyquist(num, den, w); plot(re, im, re, -im);

 -180°

0.1

Example 6.8: Nyquist Plot for a Second-Order System

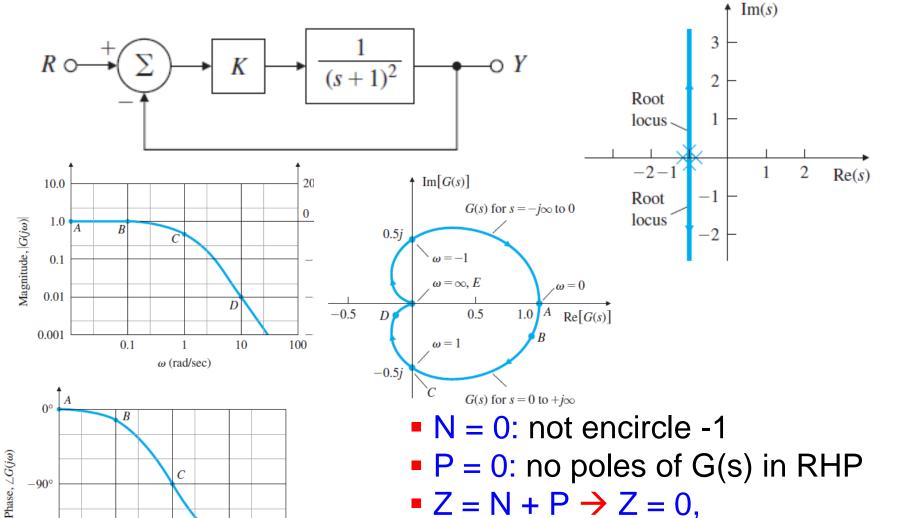
D

100

10

ω (rad/sec)

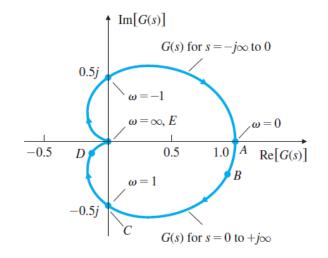
(b)



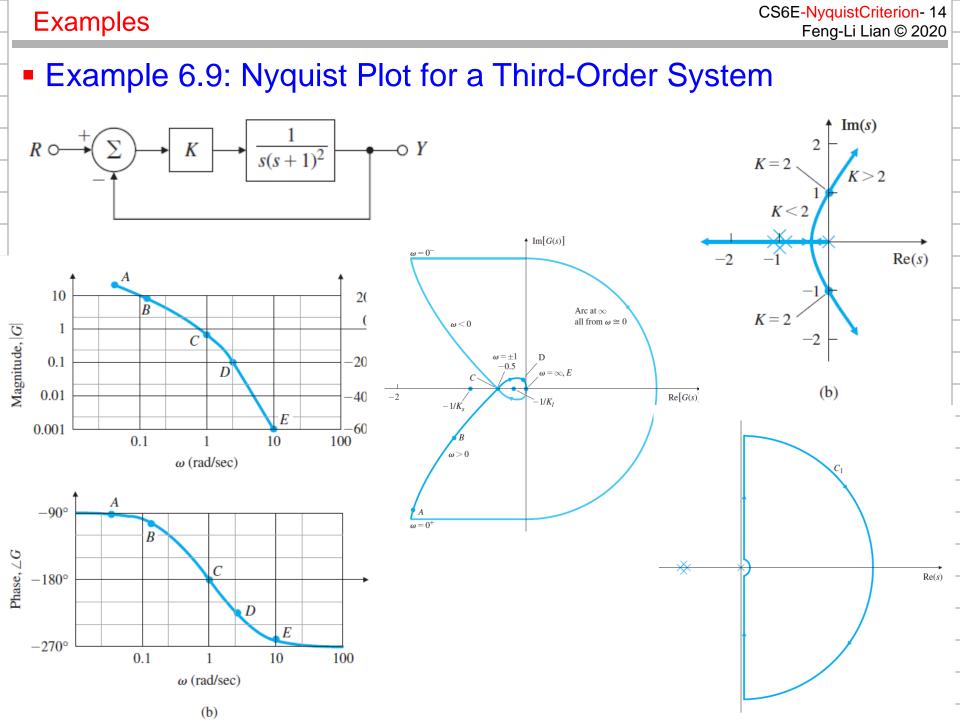
K > 0 also holds

no unstable roots for K = 1

- Example 6.8: Nyquist Plot for a Second-Order System
- Another viewpoint:
 - 1 + KG(s) for the origin point
 - KG(s) for the -1 point
 - G(s) for the -1/K point

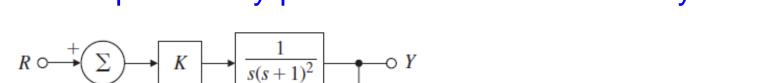


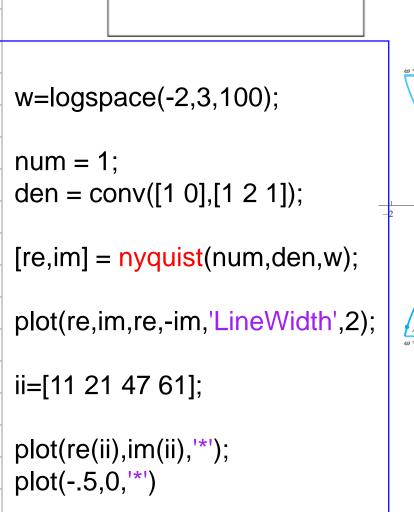
- No encirclement of G(s) on -1/Kfor any K > 0
- Hence, K > 0 is stable

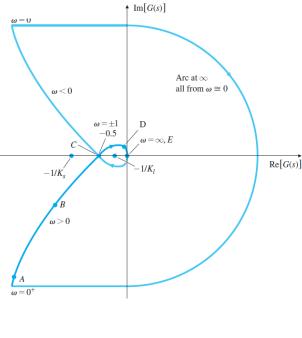


Examples

Example 6.9: Nyquist Plot for a Third-Order System



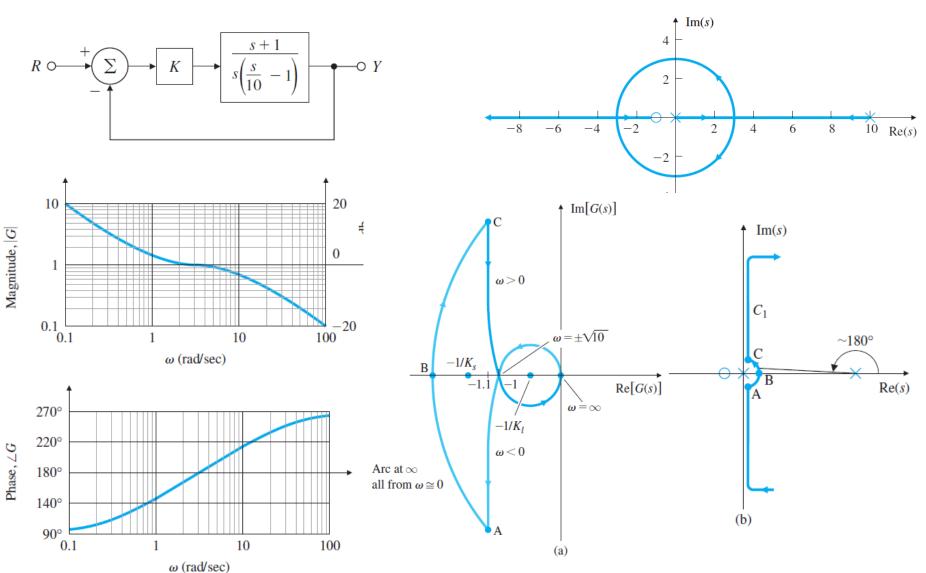




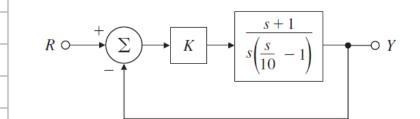
Examples

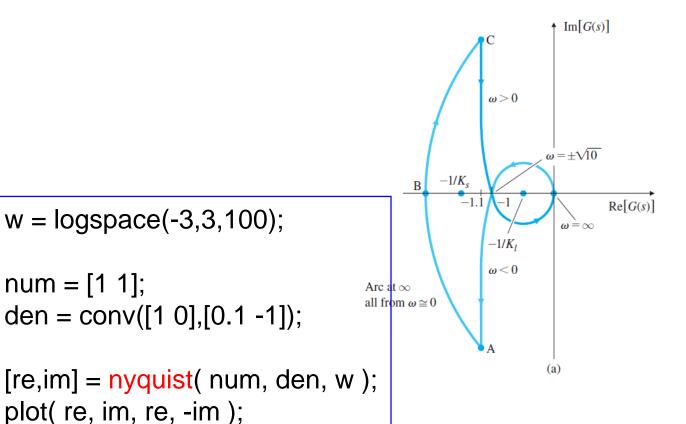
(b)

Example 6.10: Nyquist Plot for an Open-Loop Unstable System

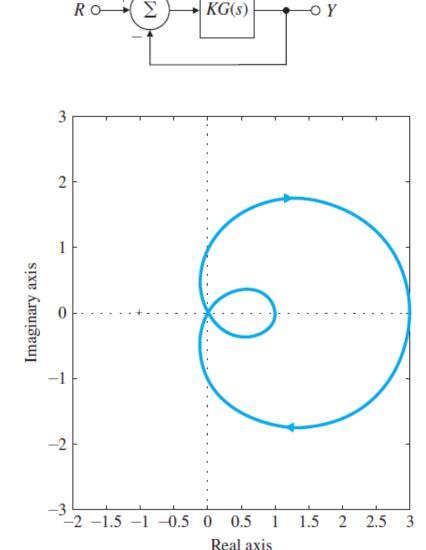


Example 6.10: Nyquist Plot for an Open-Loop Unstable System





Example 6.11: Nyquist Plot Characteristics



$$G(s) = \frac{s^2 + 3}{(s+1)^2}$$

- Never cross negative-real axis
- Stable for K > 0