Spring 2020

控制系統 Control Systems

Unit 52 Guidelines for Determining a Root Locus

Feng-Li Lian & Ming-Li Chiang

NTU-EE

Mar 2020 – Jul 2020

as follows:

Definition I:

for which 1 + KL(s) = 0 is satisfied

The root locus is the set of values of s

Definition II:

- as the real parameter K varies from 0 to + ∞ • Typically, 1 + KL(s) = 0 is the characteristic equation of the system, and
- in this case the roots on the locus are the closed-loop poles of that system. $\Rightarrow L(s) = -\frac{1}{K}$
 - The root locus of L(s) is the set of points in the s-plane where the phase of L(s) is 180° .
- To test whether a point in the s-plane is on the locus, the angle to the test point from a zero as ψ and we define the angle to test point from a pole as ϕ

$$\sum \psi_i - \sum \phi_i = 180^o + 360^o (l - 1)$$

K is real and positive, the phase of L(s) is 180° ,

the positive locus or 180° locus

 $\Rightarrow L(s) = -\frac{1}{V}$

K is real and negative,

the phase of L(s) is 0° ,

the negative locus or 0° locus

 $L(s_0) = 180^o + 360^o (l - 1)$

Illustrative Example:
$$L(s)$$

 $s_0 = -1 + 2j$

Illustrative Example:
$$L(s)$$

Illustrative Example:
$$L(s)$$

Illustrative Example:
$$L(s) = \frac{s+1}{s(s+5)[(s+2)^2+4)]}$$

Illustrative Example:
$$L(s) =$$

Illustrative Example:
$$L(s)$$

Illustrative Example:
$$L(s)$$

$$\Rightarrow$$

Formal Definition of Root Locus

CS52-RLGuidelines - 4 Feng-Li Lian © 2020 s + 1

Illustrative Example:

Istrative Example:
$$L(s) = \frac{1}{s(s+5)[(s+2)^2+4)]}$$

$$\angle L(s_0) = 180^o + 360^o (l - 1)$$

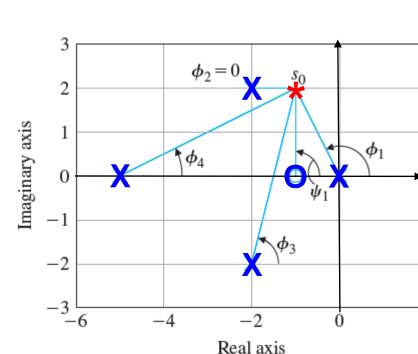
$$=\sum \psi_i - \sum \phi_i$$

$$= \angle(s_0 + 1) \\ - \angle(s_0) - \angle(s_0 + 5)$$

$$-\angle[(s_0+2)^2+4]$$

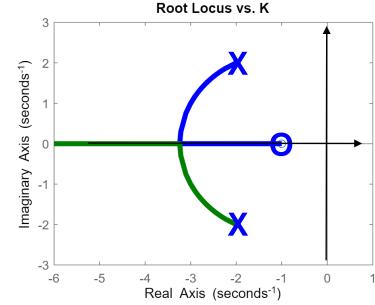
$$= 90^{o} - 116.6^{o} - 0^{o} - 76^{o} - 26.6^{o}$$

$$= -129.2^{o} \neq 180^{o}$$
 is not on the root locus



 \Rightarrow s₀ is not on the root locus

$$L(s) = \frac{s+1}{s^2 + 4s + 8}$$

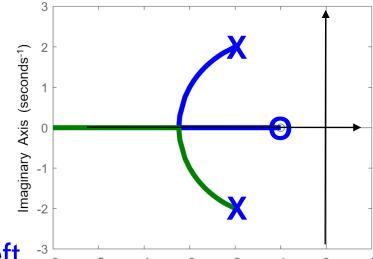


- Rule 1:
- The n branches of the locus start at the poles of L(s) and
- m of these branches end on the zeros of L(s).
- a(s) + K b(s) = 0,
- If K = 0, then a(s) = 0, whose roots are the poles.
- When $K \rightarrow \infty$, then b(s) = 0 (m zeros) or $s \rightarrow \infty$. (the rest n-m)

Rules for Determining a Positive (180°) Root Locus

CS52-RLGuidelines - 6 Feng-Li Lian © 2020 Root Locus vs. K

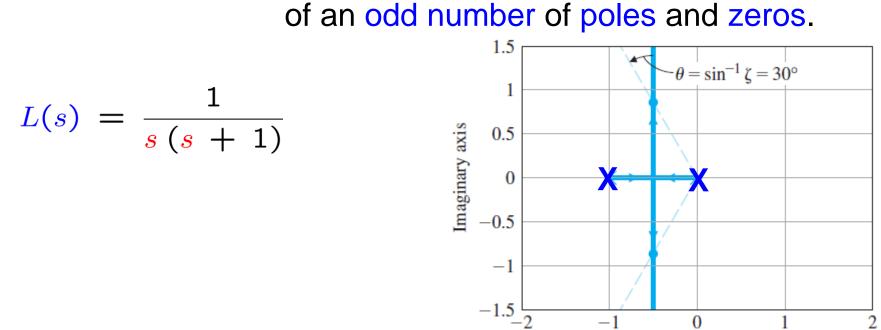
$$L(s) = \frac{s+1}{s^2 + 4s + 8}$$



Real axis

- Rule 2:
- The loci are on the real axis to the left -3-6

Real Axis (seconds⁻¹)



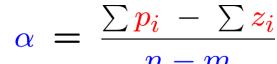
- Rule 3:
- For large s and K,

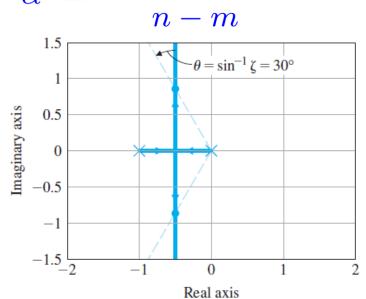
n-m branches of the loci are asymptotic

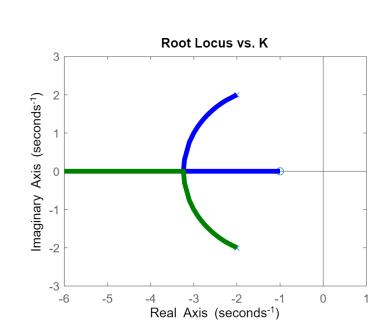
to lines at angles ϕ radiating out

from the point $s = \alpha$ on the real axis, where

$$\phi_l = \frac{180^o + 360^o (l-1)}{n-m}$$
 $l = 1, 2, \dots, n-m$







CS52-RLGuidelines - 8 Feng-Li Lian © 2020

Rule 3:

$$ullet$$
 As $K o \infty$, $L(s) = -rac{1}{K} \Rightarrow L(s) = 0$

1) *m* roots will be found to approach the zeros of L(s)

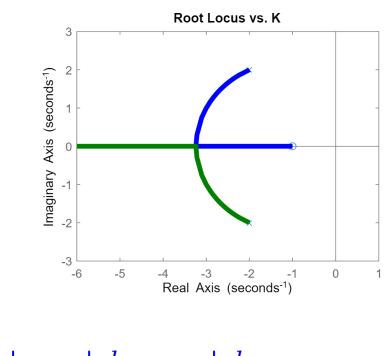
to approach the zeros of L(s)
2)
$$s \rightarrow \infty$$
 because $n >= m$

that is, n-m roots approach $s \rightarrow \infty$

$$\Rightarrow 1 + \frac{b(s)}{a(s)} = 0$$

$$\Rightarrow 1 + K \frac{s^n + a_1 s^n}{s^n + a_1 s^n}$$

 $\Rightarrow 1 + K \frac{s^m + b_1 s^{m-1} + b_2 s^{m-2} + \dots + b_{m-1} s + b_m}{s^n + a_1 s^{n-1} + a_2 s^{n-2} + \dots + a_{m-1} s + a_n} = 0$ Can be approximated by



$$\Rightarrow 1 + K \frac{1}{(s-\alpha)^{n-m}} = 0$$

Feng-Li Lian © 2020

 $l = 1, 2, \cdots, (n-m)$

CS52-RLGuidelines - 9

2

Rule 3:

$$\Rightarrow 1 + K \frac{1}{(s-\alpha)^{n-m}} = 0$$

$$R e^{j\phi}$$

ch point:
$$s_0 = R e^{j\phi}$$

h point:
$$s_0 = R e^{j\phi}$$

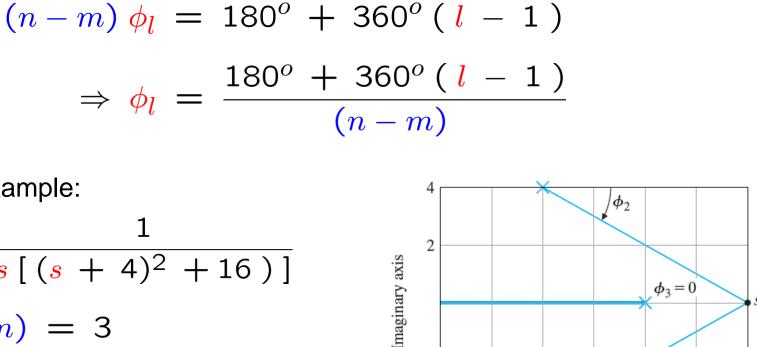
The search point:
$$s_0 = R e^{j\phi}$$

earch point:
$$s_0 = R e^{j\phi}$$

e search point:
$$s_0 = R e^{j\phi}$$

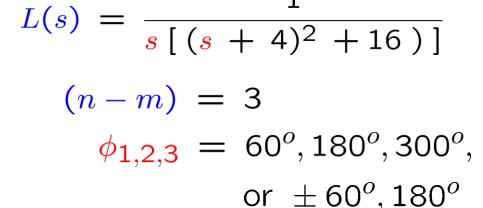
$$\frac{1}{(s-\alpha)^{n-m}} = 0$$

$$\frac{1}{m} = 0$$



Real axis

For this example: $L(s) = \frac{1}{s[(s + 4)^2 + 16)]}$



• Rule 3:

Determine asymptotic lines:

$$a(s) - s^n \perp a_1 s^{n-1} \perp a_2$$

$$a(s) = s^n + a_1 s^{n-1} + a_2 s^{n-2} + \dots + a_{n-1} s + a_n$$

$$s^n + a_1 s^{n-1} + a_2 s^n$$

 $\Rightarrow b_1 = -z_1 - z_2 \cdots - z_{n-1} - z_n$

$$s^n + a_1 s^{n-1} + a_2 s^n$$

$$-a_1s^{n-1}+a_2s^n$$

$$-a_1s^{n-1}+a_2s^n$$

$$^{1}+a_{2}s^{n-2}+$$

 $\Rightarrow a_1 = -p_1 - p_2 \cdots - p_{n-1} - p_n = -\sum_{i} p_i$

 $b(s) = s^m + b_1 s^{m-1} + b_2 s^{m-2} + \dots + b_{m-1} s + b_m$

 $= (s-z_1)(s-z_2)\cdots(s-z_{m-1})(s-z_m)$

$$a_1 + \cdots + a_n$$

$$= (s-p_1)(s-p_2)\cdots(s-p_{n-1})(s-p_n)$$

 $=-\sum z_i$

- Determine asymptotic lines:

$$\Rightarrow a(s) + K b(s) = 0$$

$$\Rightarrow s^n + a_1 s^{n-1} + a_2 s^n$$

$$\Rightarrow s^n + a_1 s^{n-1} + a_2 s^{n-2} + \dots + a_{n-1} s + a_n$$

$$a^{n-1} + a_2 s^{n-1}$$

$$a^{-1} + a_2 s^{n-1}$$

$$+a_2s^{n-2}$$

$$+ u_2s$$
 $+$

$$+K(s^m+b_1s^{m-1}+b_2s^{m-2}+\cdots+b_{m-1}s+b_m) = 0$$

$$= (s-r_1)(s-r_2)\cdots(s-r_{n-1})(s-r_n) = 0$$

$$\Rightarrow a_1 = -r_1 - r_2 \cdots - r_{n-1} - r_n$$

- And this term is independent of K
- The open-loop sum and closed-loop sum are the same and are equal to $-a_1$ $\Rightarrow -\sum r_i = -\sum p_i$

 $= -\sum r_i$

Rule 3:

For large values of *K*:

 $\overline{(s-\alpha)^{n-m}}$

 $\Rightarrow \alpha = \frac{\sum p_i - \sum z_i}{(n-m)}$

 $\Rightarrow \alpha = \frac{-4 - 4 + 0}{3 - 0}$

 $=-\frac{8}{3}=-2.67$

 $= \pm 60^{\circ}, 180^{\circ}$

For this example:

of the roots r_i approach the zeros z_i

 $\Rightarrow -\sum r_i = -(n-m)\alpha - \sum z_i = -\sum p_i$

• n - m of the roots r_i approach the branches of the asymptotic system

whose poles add up to $(n-m)\alpha$

maginary axis

CS52-RLGuidelines - 12

 $2.67 \times \sqrt{3} = 4.62$

Real axis

Feng-Li Lian © 2020

- $L(s) = \frac{1}{s \left[(s + 4)^2 + 16 \right]}$

- Rule 4:
- The angle of departure of a branch of the locus from a single pole is given by

$$\phi_{dep} = \sum \psi_i - \sum_{i \neq dep} \phi_i - 180^o$$

$$\sum \psi_i$$
 the sum of the angles to all the zeros

 $\sum \phi_i$ the sum of the angles to the remaining poles

The angle of departure of a branch of the locus from repeated poles with multiplicity q is given by

$$rac{q \; \phi_{l,dep}}{\phi_{l,dep}} \; = \; \sum \psi_i - \sum_{i
eq l,dep} \phi_i - 180^o - 360^o (l-1)$$

CS52-RLGuidelines - 13

Feng-Li Lian © 2020

- Rule 4:
- The angle of arrival of a branch at a zero with multiplicity q is given by

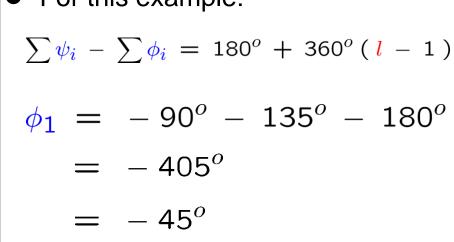
is given by
$$\frac{q\ \psi_{l,arr}}{\psi_{l,arr}} = \sum_{i\neq l,arr} \frac{\psi_i}{\psi_i} + 180^o + 360^o (l-1)$$

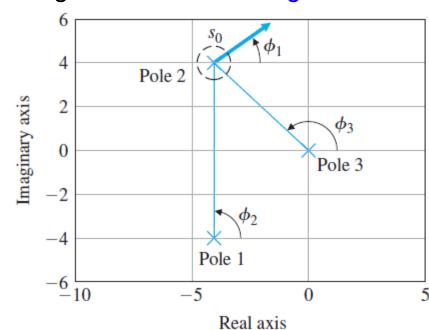
$$\frac{l}{l} = 1,2,\cdots,\frac{q}{l}$$

$$\sum \psi_i$$
 the sum of the angles to the remaining zeros

 $\sum \phi_i$ the sum of the angles to all the poles

For this example:

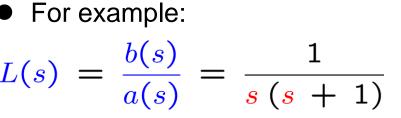


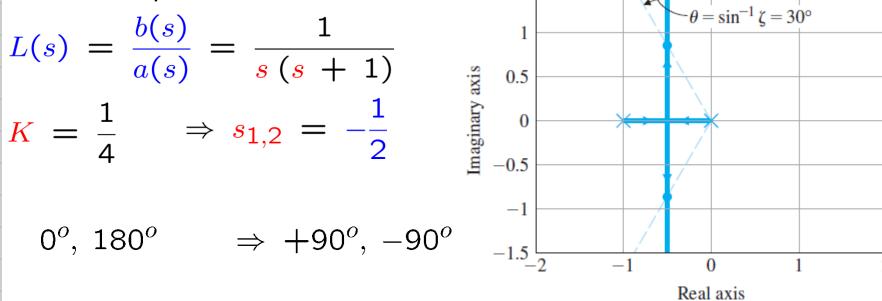


Rule 5:

separated by $180^{o} - 360^{o}(l-1)$

The locus can have multiple roots at points on the locus and the branches will approach a point of q roots at angles





Real axis

CS52-RLGuidelines - 16 Feng-Li Lian © 2020

Rules for Determining a Positive (180°) Root Locus

- Rule 5:
- Continuation Locus:

$$L(s) = \frac{b(s)}{a(s)} = \frac{1}{s(s+1)}$$

$$a(s)$$
 $s(s+1)$
 $K_1 = \frac{1}{4}$ $\Rightarrow K = K_1 + K_2 = \frac{1}{4} + K_2$

$$\Rightarrow s^2 + s + \frac{1}{4} + K_2 = 0$$

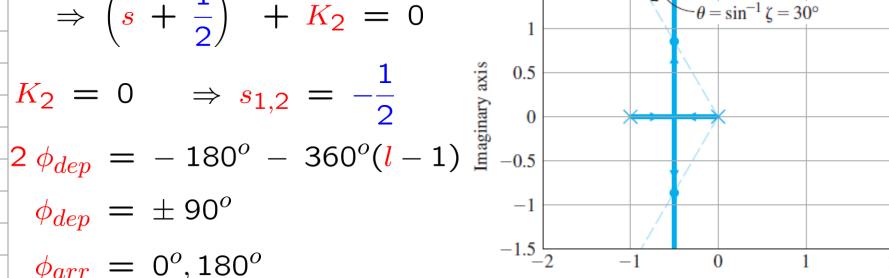
$$\Rightarrow \left(\frac{s}{2} + \frac{1}{2}\right)^2 + K_2 = 0$$

$$s_{1,2} = 0$$

$$s_{1,2} = 0$$

$$s_{1,2} = 0$$

$$s_{1,2} = 0$$

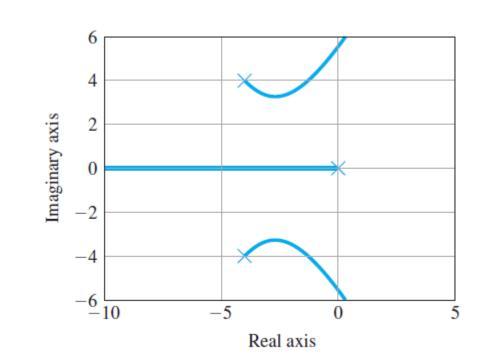


• The third-order example:

$$L(s) = \frac{1}{s(s^2 + 8s + 32)}$$

$$s = tf('s')$$

 $sysL = (1)/(s*(s^2+8*s+32));$
 $sysL = 1/(s*((s+4)^2+16));$
 $rlocus(sysL);$



- Rule 1:
- The n branches of the locus start at the poles of L(s) and
- m of these branches end on the zeros of L(s).
- The loci are on the real axis to the left of an odd number of poles and zeros.
- Rule 3:

Rule 2:

For large s and K, n-m branches of the loci are asymptotic to lines at angles φ radiating out

from the point $s = \alpha$ on the real axis, where

$$\phi_{l} = \frac{180^{o} + 360^{o} (l - 1)}{n - m}$$
 $\alpha = \frac{\sum p_{i} - p_{i}}{n - m}$
 $\alpha = \frac{n - m}{n - m}$

• Rule 4:

The angle of departure of a branch of the locus from repeated poles with multiplicity q is given by $q \phi_{l,dep} = \sum \psi_i - \sum \phi_i - 180^o - 360^o (l-1)$

$$i
eq l, dep$$
 $l=1,2,\cdots,q$

The angle of arrival of a branch at a zero with multiplicity q is given by

is given by
$$q~\psi_{l,arr}~=~\sum \phi_i - \sum_{i
eq l,arr} \psi_i + 180^o + 360^o (\emph{l}-1)$$

- Rule 5:
- The locus can have multiple roots at points on the locus and the branches will approach a point of q roots at angles $180^{o} - 360^{o}(l-1)$ separated by
 - And will depart at angles with same separation.

The positive root locus
 is a plot of all possible locations
 for roots to the equation 1 + K L(s) = 0
 for some real positive value of K.

The purpose of design
 is to select a particular value of K
 that will meet the specifications
 for static and dynamic response.

CS52-RLGuidelines - 21 Feng-Li Lian © 2020

$$\frac{L(s)}{s} = \frac{1}{s [(s + 4)^2 + 16)]}$$

$$L(s_0) = rac{1}{s_0 \left(s_0 - s_2
ight) \left(s_0 - s_3
ight)} = rac{1}{|L(s_0)|} = |s_0| \left|s_0 - s_2
ight| \left|s_0 - s_3
ight| = s_3 = 1$$
 $pprox 4.0 imes 2.1 imes 7.7$

 $s_0 - s_2$

 $\zeta = 0.5$

[K, p] = rlocfind(sysL);

rlocus(sysL);

 $sysL = (1)/(s*((s+4)^2+16));$

- Compute the error constant of the control system
- For example,
 the steady-state error in tracking a ramp input
 is giving by the velocity constant:

$$K_v = \lim_{s \to 0} s K L(s)$$

$$= \lim_{s \to 0} s K \frac{1}{s [(s + 4)^2 + 16)]}$$

$$= \frac{K}{32}$$

$$= \frac{65}{32} \approx 2 \sec^{-1}$$